



INA2128

Dual, Low Power INSTRUMENTATION AMPLIFIER

FEATURES

● LOW OFFSET VOLTAGE: 50µV max

● LOW DRIFT: 0.5µV/°C max

● LOW INPUT BIAS CURRENT: 5nA max

• HIGH CMR: 120dB min

INPUTS PROTECTED TO ±40V

● WIDE SUPPLY RANGE: ±2.25V to ±18V

LOW QUIESCENT CURRENT: 700μA / IA

• 16-PIN PLASTIC DIP, SOL-16

APPLICATIONS

- SENSOR AMPLIFIER THERMOCOUPLE, RTD, BRIDGE
- MEDICAL INSTRUMENTATION
- MULTIPLE-CHANNEL SYSTEMS
- BATTERY OPERATED EQUIPMENT

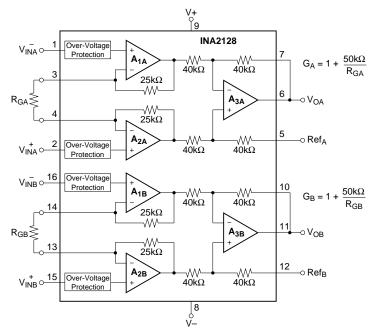
DESCRIPTION

The INA2128 is a dual, low power, general purpose instrumentation amplifier offering excellent accuracy. Its versatile 3-op amp design and small size make it ideal for a wide range of applications. Current-feedback input circuitry provides wide bandwidth even at high gain (200kHz at G = 100).

A single external resistor sets any gain from 1 to 10,000. Internal input protection can withstand up to $\pm 40V$ without damage.

The INA2128 is laser trimmed for very low offset voltage (50µV), drift (0.5µV/°C) and high common-mode rejection (120dB at G \geq 100). It operates with power supplies as low as $\pm 2.25 V$, and quiescent current is only 700µA per IA—ideal for battery operated and multiple-channel systems.

The INA2128 is available in 16-pin plastic DIP, and SOL-16 surface-mount packages, specified for the -40°C to +85°C temperature range.



International Airport Industrial Park • Mailing Address: PO Box 11400, Tucson, AZ 85734 • Street Address: 6730 S. Tucson Blvd., Tucson, AZ 85706 • Tel: (520) 746-1111 • Twx: 910-952-1111 Internet: http://www.burr-brown.com/ • FAXLine: (800) 548-6133 (US/Canada Only) • Cable: BBRCORP • Telex: 066-6491 • FAX: (520) 889-1510 • Immediate Product Info: (800) 548-6132

SPECIFICATIONS

At T_A = +25°C, V_S = ±15V, R_L = 10k Ω , unless otherwise noted.

			INA2128P, U		INA2128PA, UA			
PARAMETER	CONDITIONS	MIN	TYP	MAX	MIN	TYP	MAX	UNITS
INPUT Offset Voltage, RTI Initial vs Temperature vs Power Supply Long-Term Stability Impedance, Differential	$T_A = +25^{\circ}C$ $T_A = T_{MIN} \text{ to } T_{MAX}$ $V_S = \pm 2.25 \text{ V to } \pm 18 \text{ V}$		±10 ±100/G ±0.2 ± 2/G ±0.2 ±20/G ±0.1 ±3/G 10 ¹⁰ 2	±50 ±500/G ±0.5 ± 20/G ±1 ±100/G		±25 ±100/G ±0.2 ± 5/G * *	±125 ±1000/G ±1 ± 20/G ±2 ±200/G	μV μV/°C μV/V μV/mo Ω pF
Common-Mode Common-Mode Voltage Range Safe Input Voltage	$V_0 = 0V$	(V+) - 2 (V-) + 2	10 ¹¹ 9 (V+) – 1.4 (V–) + 1.7	±40	* *	* * *	*	Ω pF V V V
Common-Mode Rejection	$V_{CM} = \pm 13 V, \Delta R_S = 1 k \Omega$ $G = 1$ $G = 10$ $G = 100$ $G = 100$	80 100 120 120	86 106 125 130		73 93 110 110	* * * *	·	dB dB dB dB
BIAS CURRENT vs Temperature Offset Current vs Temperature			±2 ±30 ±1 ±30	±5 ±5		* * *	±10 ±10	nA pA/°C nA pA/°C
NOISE VOLTAGE, RTI f = 10Hz f = 100Hz f = 10Hz f = 1kHz f _B = 0.1Hz to 10Hz Noise Current	$G = 1000, R_S = 0\Omega$		10 8 8 0.2			* * *		nV/√ <u>Hz</u> nV/√ <u>Hz</u> nV/√Hz μVp-p
f=10Hz f=1kHz f _B = 0.1Hz to 10Hz			0.9 0.3 30			* * *		pA/√ <u>Hz</u> pA/√Hz pAp-p
GAIN Gain Equation Range of Gain Gain Error Gain vs Temperature ⁽²⁾ 50kΩ Resistance ^(2, 3) Nonlinearity	$G=1$ $G=10$ $G=100$ $G=1000$ $G=1$ $V_0 = \pm 13.6V, G=1$	1	$\begin{array}{c} 1 + (50 k\Omega/R_G) \\ \pm 0.01 \\ \pm 0.02 \\ \pm 0.05 \\ \pm 0.5 \\ \pm 1 \\ \pm 25 \\ \pm 0.0001 \end{array}$	10000 ±0.024 ±0.4 ±0.5 ±1 ±10 ±100 ±0.001	*	* * * * * * * * * * * * * * * * * * * *	* ±0.1 ±0.5 ±0.7 ±2 * * ±0.002	V/V V/V % % % ppm/°C ppm/°C
	G=10 G=100 G=1000		±0.0003 ±0.0005 ±0.001	±0.002 ±0.002 (Note 4)		* * *	±0.004 ±0.004 *	% of FSR % of FSR % of FSR
OUTPUT Voltage: Positive Negative Load Capacitance Stability Short-Circuit Current	$R_{L} = 10k\Omega$ $R_{L} = 10k\Omega$	(V+) - 1.4 (V-) + 1.4	(V+) - 0.9 (V-) + 0.8 1000 +6/-15		* *	* * * *		V V pF mA
FREQUENCY RESPONSE Bandwidth, -3dB	G=1 G=10 G=100 G=1000		1.3 700 200 20			* * * *		MHz kHz kHz kHz
Slew Rate Settling Time, 0.01% Overload Recovery	$V_O = \pm 10V$, G=10 G=1 G=10 G=100 G=1000 50% Overdrive		4 7 7 9 80 4			* * * * * *		V/μs μs μs μs μs
POWER SUPPLY Voltage Range Current, Total	V _{IN} = 0V	±2.25	±15 ±1.4	±18 ±1.5	*	* * *	*	μs V mA
TEMPERATURE RANGE Specification Operating θ_{JA}		-40 -40	80	85 125	*	*	*	°C °C °C

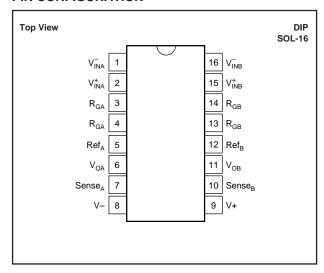
^{*} Specification same as INA2128P, U.

NOTE: (1) Input common-mode range varies with output voltage—see typical curves. (2) Guaranteed by wafer test. (3) Temperature coefficient of the " $50k\Omega$ " term in the gain equation. (4) Nonlinearity measurements in G = 1000 are dominated by noise. Typical nonlinearity is $\pm 0.001\%$.

The information provided herein is believed to be reliable; however, BURR-BROWN assumes no responsibility for inaccuracies or omissions. BURR-BROWN assumes no responsibility for the use of this information, and all use of such information shall be entirely at the user's own risk. Prices and specifications are subject to change without notice. No patent rights or licenses to any of the circuits described herein are implied or granted to any third party. BURR-BROWN does not authorize or warrant any BURR-BROWN product for use in life support devices and/or systems.



PIN CONFIGURATION



ABSOLUTE MAXIMUM RATINGS

Supply Voltage	±18V
Analog Input Voltage Range	
Output Short-Circuit (to ground)	Continuous
Operating Temperature	40°C to +125°C
Storage Temperature	40°C to +125°C
Junction Temperature	+150°C
Lead Temperature (soldering, 10s)	+300°C



This integrated circuit can be damaged by ESD. Burr-Brown recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

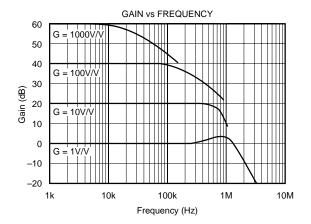
ORDERING INFORMATION

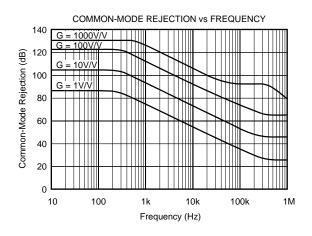
PRODUCT	PACKAGE	PACKAGE DRAWING NUMBER ⁽¹⁾	TEMPERATURE RANGE
INA2128PA	16-Pin Plastic DIP	180	-40°C to +85°C
INA2128P	16-Pin Plastic DIP	180	-40°C to +85°C
INA2128UA	SOL-16 Surface-Mount	211	-40°C to +85°C
INA2128U	SOL-16 Surface-Mount	211	-40°C to +85°C

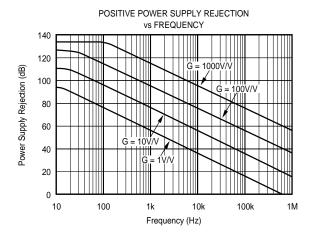
NOTE: (1) For detailed drawing and dimension table, please see end of data sheet, or Appendix C of Burr-Brown IC Data Book.

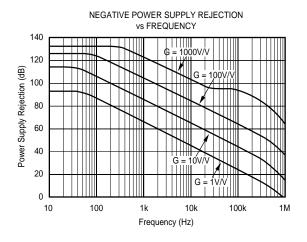
TYPICAL PERFORMANCE CURVES

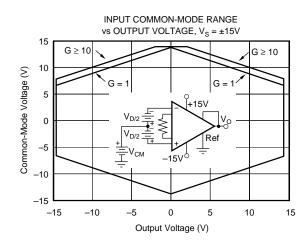
At $T_A = +25$ °C, $V_S = \pm 15$ V, unless otherwise noted.

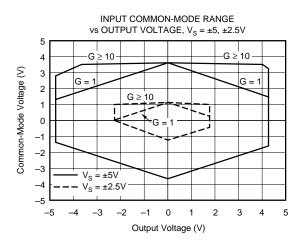








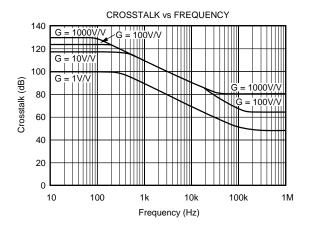


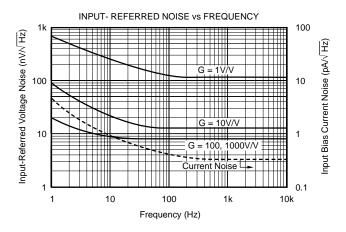


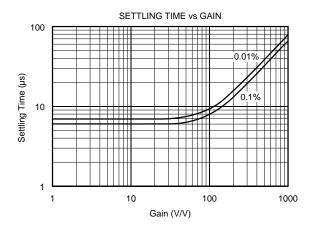


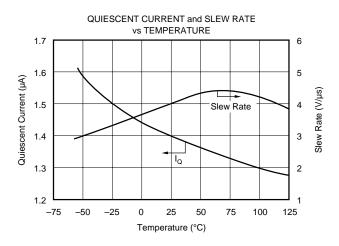
TYPICAL PERFORMANCE CURVES (CONT)

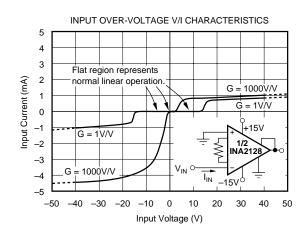
At $T_A = +25^{\circ}C$, $V_S = \pm 15V$, unless otherwise noted.

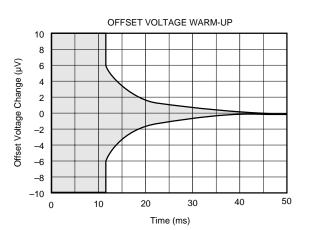






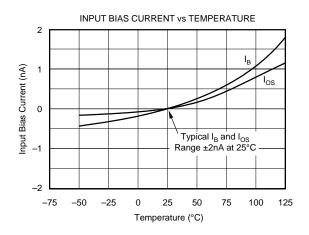


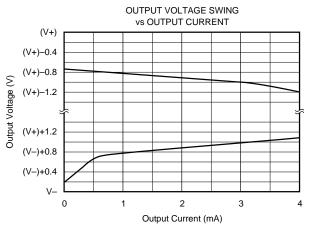


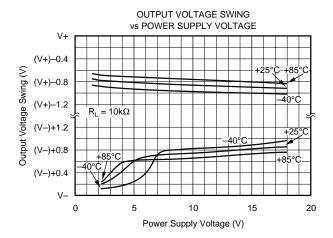


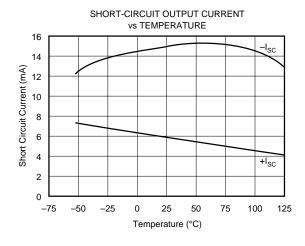
TYPICAL PERFORMANCE CURVES (CONT)

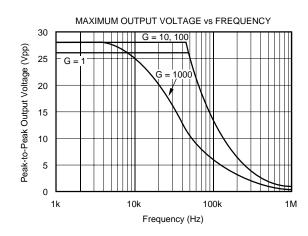
At $T_A = +25$ °C, $V_S = \pm 15$ V, unless otherwise noted.

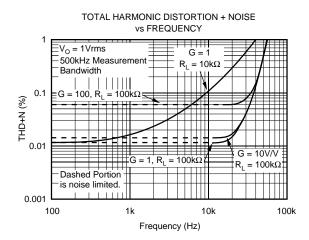








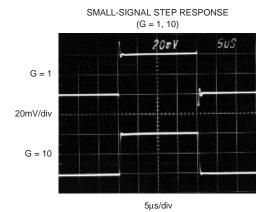


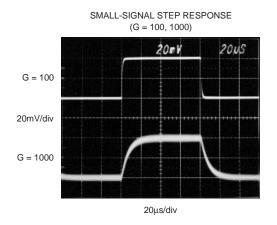


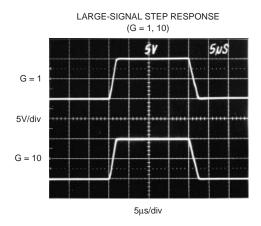


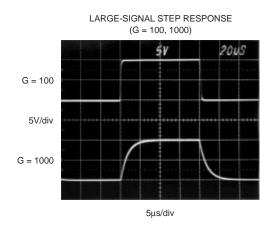
TYPICAL PERFORMANCE CURVES (CONT)

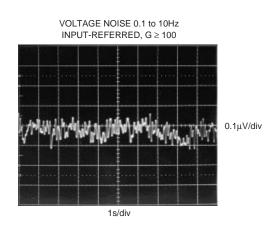
At $T_A = +25^{\circ}C$, $V_S = \pm 15V$, unless otherwise noted.













APPLICATION INFORMATION

Figure 1 shows the basic connections required for operation of the INA2128. Applications with noisy or high impedance power supplies may require decoupling capacitors close to the device pins as shown.

The output is referred to the output reference (Ref) terminals (Ref_A and Ref_B) which are normally grounded. These must be low-impedance connections to assure good common-mode rejection. A resistance of 8Ω in series with a Ref pin will cause a typical device to degrade to approximately 80dB CMR (G=1).

The INA2128 has a separate output sense feedback connections Sense_A and Sense_B. These must be connected to their respective output terminals for proper operation. The output sense connection can be used to sense the output voltage directly at the load for best accuracy.

SETTING THE GAIN

Gain of the INA2128 is set by connecting a single external resistor, R_G , connected as shown:

$$G = 1 + \frac{50k\Omega}{R_G}$$
 (1)

Commonly used gains and resistor values are shown in Figure 1.

The $50k\Omega$ term in equation 1 comes from the sum of the two

internal feedback resistors of A_1 and A_2 . These on-chip metal film resistors are laser trimmed to accurate absolute values. The accuracy and temperature coefficient of these resistors are included in the gain accuracy and drift specifications of the INA2128.

The stability and temperature drift of the external gain setting resistor, R_G, also affects gain. R_G's contribution to gain accuracy and drift can be directly inferred from the gain equation (1). Low resistor values required for high gain can make wiring resistance important. Sockets add to the wiring resistance which will contribute additional gain error in gains of approximately 100 or greater.

DYNAMIC PERFORMANCE

The typical performance curve "Gain vs Frequency" shows that despite its low quiescent current, the INA2128 achieves wide bandwidth, even at high gain. This is due to its current-feedback topology. Settling time also remains excellent at high gain—see "Settling Time vs Gain."

NOISE PERFORMANCE

The INA2128 provides very low noise in most applications. Low frequency noise is approximately $0.2\mu\text{Vp-p}$ measured from 0.1 to 10Hz (G \geq 100). This provides dramatically improved noise when compared to state-of-the-art chopper-stabilized amplifiers.

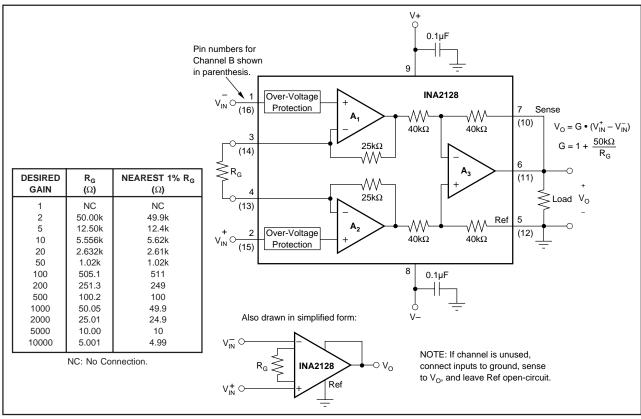


FIGURE 1. Basic Connections.



OFFSET TRIMMING

The INA2128 is laser trimmed for low offset voltage and offset voltage drift. Most applications require no external offset adjustment. Figure 2 shows an optional circuit for trimming the output offset voltage. The voltage applied to Ref terminal is summed with the output. The op amp buffer provides low impedance at the Ref terminal to preserve good common-mode rejection.

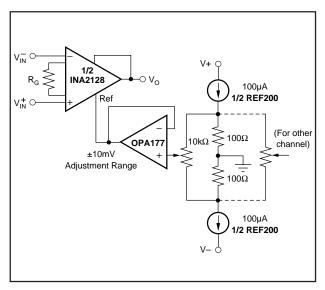


FIGURE 2. Optional Trimming of Output Offset Voltage.

INPUT BIAS CURRENT RETURN PATH

The input impedance of the INA2128 is extremely high—approximately $10^{10}\Omega$. However, a path must be provided for the input bias current of both inputs. This input bias current is approximately $\pm 2nA$. High input impedance means that this input bias current changes very little with varying input voltage.

Input circuitry must provide a path for this input bias current for proper operation. Figure 3 shows various provisions for an input bias current path. Without a bias current path, the inputs will float to a potential which exceeds the common-mode range of the INA2128 and the input amplifiers will saturate.

If the differential source resistance is low, the bias current return path can be connected to one input (see the thermocouple example in Figure 3). With higher source impedance, using two equal resistors provides a balanced input with possible advantages of lower input offset voltage due to bias current and better high-frequency common-mode rejection.

INPUT COMMON-MODE RANGE

The linear input voltage range of the input circuitry of the INA2128 is from approximately 1.4V below the positive supply voltage to 1.7V above the negative supply. As a differential input voltage causes the output voltage increase, however, the linear input range will be limited by the output

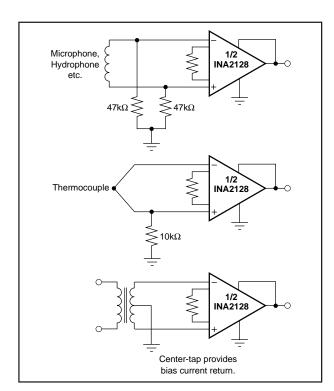


FIGURE 3. Providing an Input Common-Mode Current Path.

voltage swing of amplifiers A_1 and A_2 . So the linear common-mode input range is related to the output voltage of the complete amplifier. This behavior also depends on supply voltage—see performance curves "Input Common-Mode Range vs Output Voltage."

Input-overload can produce an output voltage that appears normal. For example, if an input overload condition drives both input amplifiers to their positive output swing limit, the difference voltage measured by the output amplifier will be near zero. The output of the INA2128 will be near 0V even though both inputs are overloaded.

LOW VOLTAGE OPERATION

The INA2128 can be operated on power supplies as low as ± 2.25 V. Performance remains excellent with power supplies ranging from ± 2.25 V to ± 18 V. Most parameters vary only slightly throughout this supply voltage range—see typical performance curves. Operation at very low supply voltage requires careful attention to assure that the input voltages remain within their linear range. Voltage swing requirements of internal nodes limit the input common-mode range with low power supply voltage. Typical performance curves, "Input Common-Mode Range vs Output Voltage" show the range of linear operation for ± 15 V, ± 5 V, and ± 2.5 V supplies.



INPUT PROTECTION

The inputs of the INA2128 are individually protected for voltages up to ± 40 V. For example, a condition of -40V on one input and +40V on the other input will not cause damage. Internal circuitry on each input provides low series impedance under normal signal conditions. To provide equivalent protection, series input resistors would contribute excessive noise. If the input is overloaded, the protection circuitry limits the input current to a safe value of approximately 1.5 to 5mA. The typical performance curve "Input Bias Current vs Common-Mode Input Voltage" shows this input current limit behavior. The inputs are protected even if the power supplies are disconnected or turned off.

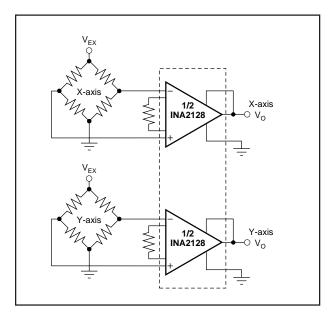


FIGURE 4. Two-Axis Bridge Amplifier.

CHANNEL CROSSTALK

The two channels of the INA2128 are completely independent, including all bias circuitry. At DC and low frequency there is virtually no signal coupling between channels. Crosstalk increases with frequency and is dependent on circuit gain, source impedance and signal characteristics.

As source impedance increases, careful circuit layout will help achieve lowest channel crosstalk. Most crossstalk is produced by capacitive coupling of signals from one channel to the input section of the other channel. To minimize coupling, separate the input traces as far as practical from any signals associated with the opposite channel. A grounded guard trace surrounding the inputs helps reduce stray coupling between channels. Run the differential inputs of each channel parallel to each other or directly adjacent on top and bottom side of a circuit board. Stray coupling then tends to produce a common-mode signal which is rejected by the IA's input.

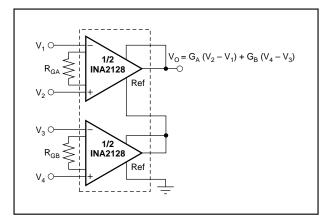


FIGURE 5. Sum of Differences Amplifier.

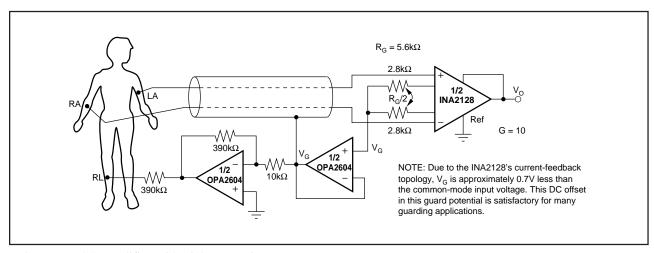


FIGURE 6. ECG Amplifier With Right-Leg Drive.







ti.com 12-Jan-2007

PACKAGING INFORMATION

Orderable Device	Status ⁽¹⁾	Package Type	Package Drawing	Pins	Package Qty	e Eco Plan ⁽²⁾	Lead/Ball Finish	MSL Peak Temp ⁽³⁾
INA2128U	ACTIVE	SOIC	DW	16	48	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR
INA2128U/1K	ACTIVE	SOIC	DW	16	1000	Pb-Free (RoHS)	CU NIPDAU	Level-3-260C-168 HR
INA2128U/1KE4	ACTIVE	SOIC	DW	16	1000	Pb-Free (RoHS)	CU NIPDAU	Level-3-260C-168 HR
INA2128UA	ACTIVE	SOIC	DW	16	48	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR
INA2128UA/1K	ACTIVE	SOIC	DW	16	1000	Pb-Free (RoHS)	CU NIPDAU	Level-3-260C-168 HR
INA2128UAG4	ACTIVE	SOIC	DW	16	48	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR
INA2128UG4	ACTIVE	SOIC	DW	16	48	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

Important Information and Disclaimer: The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

IMPORTANT NOTICE

Texas Instruments Incorporated and its subsidiaries (TI) reserve the right to make corrections, modifications, enhancements, improvements, and other changes to its products and services at any time and to discontinue any product or service without notice. Customers should obtain the latest relevant information before placing orders and should verify that such information is current and complete. All products are sold subject to TI's terms and conditions of sale supplied at the time of order acknowledgment.

TI warrants performance of its hardware products to the specifications applicable at the time of sale in accordance with TI's standard warranty. Testing and other quality control techniques are used to the extent TI deems necessary to support this warranty. Except where mandated by government requirements, testing of all parameters of each product is not necessarily performed.

TI assumes no liability for applications assistance or customer product design. Customers are responsible for their products and applications using TI components. To minimize the risks associated with customer products and applications, customers should provide adequate design and operating safeguards.

TI does not warrant or represent that any license, either express or implied, is granted under any TI patent right, copyright, mask work right, or other TI intellectual property right relating to any combination, machine, or process in which TI products or services are used. Information published by TI regarding third-party products or services does not constitute a license from TI to use such products or services or a warranty or endorsement thereof. Use of such information may require a license from a third party under the patents or other intellectual property of the third party, or a license from TI under the patents or other intellectual property of TI.

Reproduction of information in TI data books or data sheets is permissible only if reproduction is without alteration and is accompanied by all associated warranties, conditions, limitations, and notices. Reproduction of this information with alteration is an unfair and deceptive business practice. TI is not responsible or liable for such altered documentation.

Resale of TI products or services with statements different from or beyond the parameters stated by TI for that product or service voids all express and any implied warranties for the associated TI product or service and is an unfair and deceptive business practice. TI is not responsible or liable for any such statements.

Following are URLs where you can obtain information on other Texas Instruments products and application solutions:

Products		Applications	
Amplifiers	amplifier.ti.com	Audio	www.ti.com/audio
Data Converters	dataconverter.ti.com	Automotive	www.ti.com/automotive
DSP	dsp.ti.com	Broadband	www.ti.com/broadband
Interface	interface.ti.com	Digital Control	www.ti.com/digitalcontrol
Logic	logic.ti.com	Military	www.ti.com/military
Power Mgmt	power.ti.com	Optical Networking	www.ti.com/opticalnetwork
Microcontrollers	microcontroller.ti.com	Security	www.ti.com/security
Low Power Wireless	www.ti.com/lpw	Telephony	www.ti.com/telephony
		Video & Imaging	www.ti.com/video
		Wireless	www.ti.com/wireless

Mailing Address: Texas Instruments

Post Office Box 655303 Dallas, Texas 75265