

TOSHIBA BiCD Integrated Circuit Silicon Monolithic

TB67S261FTG

CLOCK-in controlled Bipolar Stepping Motor Driver

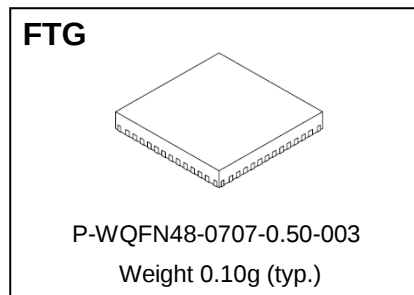
The TB67S261 is a two-phase bipolar stepping motor driver using a PWM chopper. An interface is PHASE in control. Fabricated with the BiCD process, rating is 50 V/2.0 A .

Features

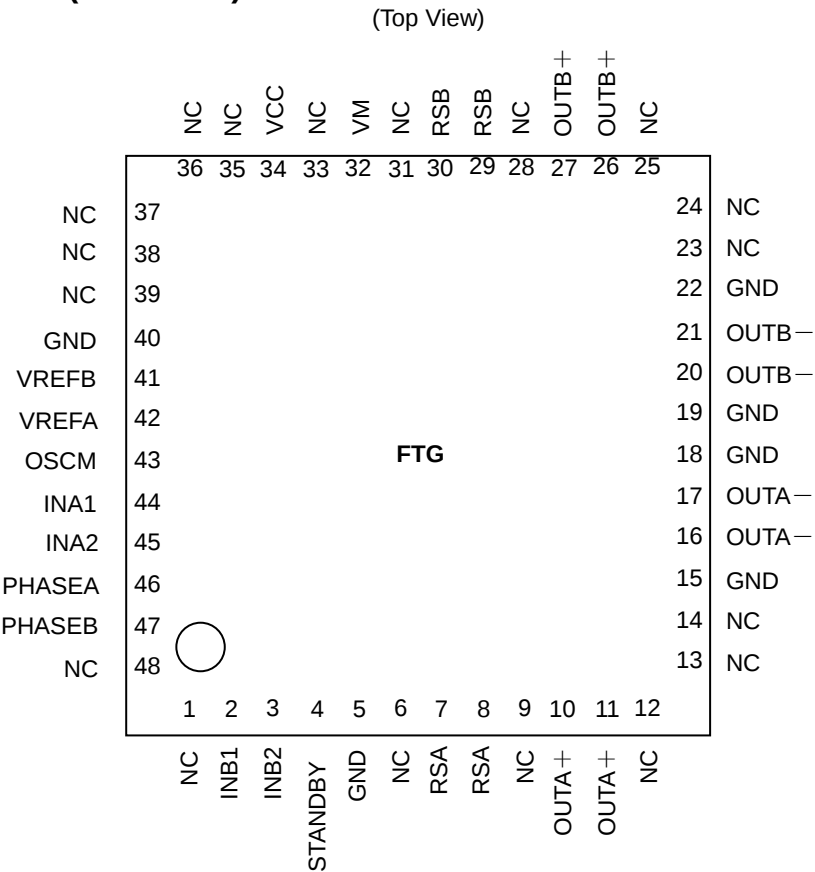
- BiCD process integrated monolithic IC.
- Capable of controlling 1 bipolar stepping motor.
- PWM controlled constant-current drive.
- Allows full, half, quarter step operation.
- Low on-resistance (High + Low side=0.8Ω(typ)) MOSFET output stage.
- High efficiency motor current control mechanism (Advanced Dynamic Mixed Decay)
- High voltage and current (For specification, please refer to absolute maximum ratings and operation ranges)
- Error detection (TSD/ISD) signal output function
- Built-in error detection circuits (Thermal shutdown (TSD), over-current shutdown (ISD), and power-on reset (POR))
- Built-in VCC regulator for internal circuit use.
- Chopping frequency of a motor can be customized by external resistance and capacitor.
- package

TB67S261FTG: P-WQFN48-0707-0.50-003

Note) Please be careful about thermal conditions during use.

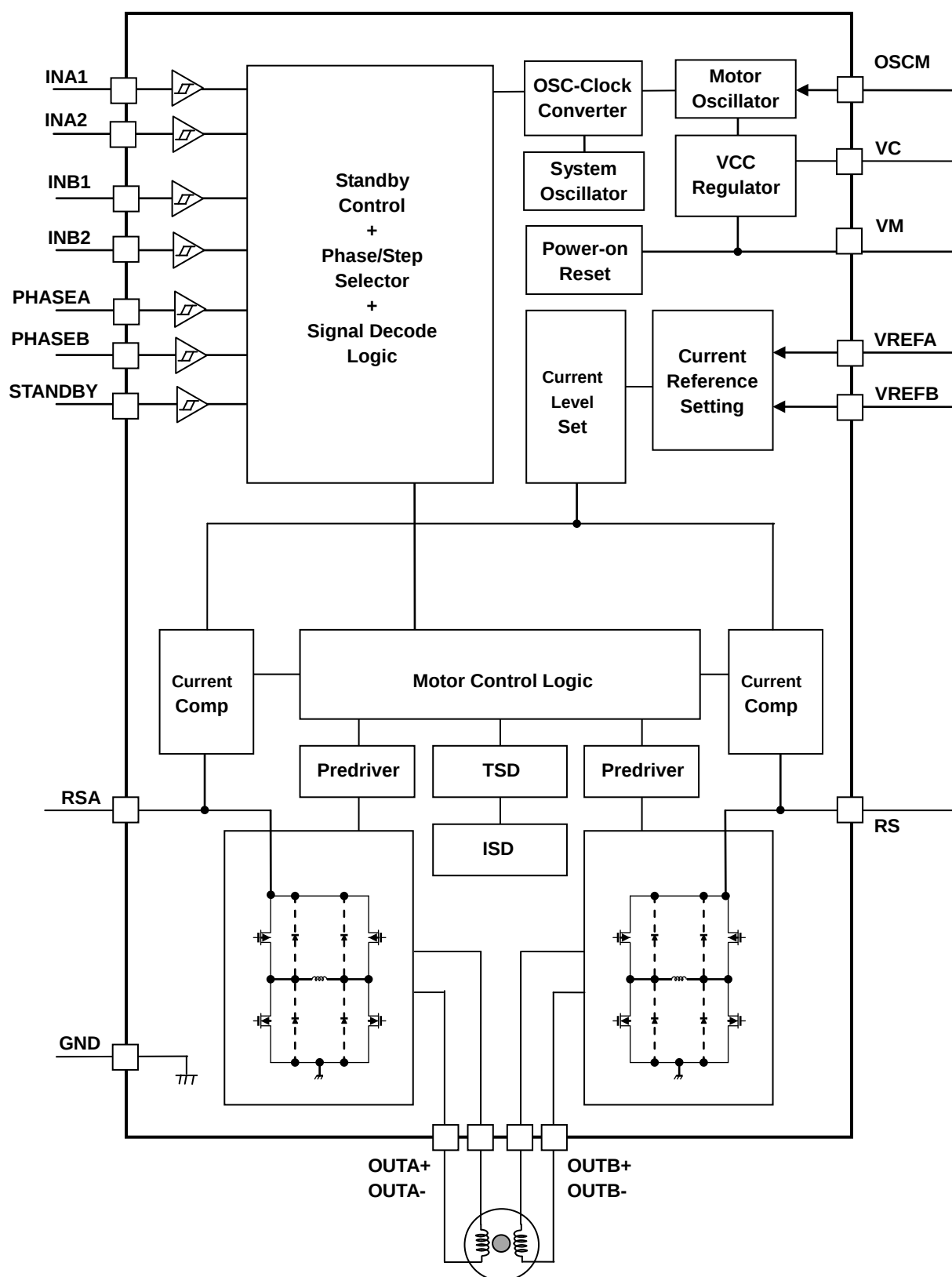


Pin assignment (TB67S261)



Please mount the four corner pins of the QFN package and the exposed pad to the GND area of the PCB.

TB67S261 Block diagram



Functional blocks/circuits/constants in the block chart etc. may be omitted or simplified for explanatory purposes.

Application Notes

All the grounding wires of the TB67S261 must run on the solder mask on the PCB and be externally terminated at only one point. Also, a grounding method should be considered for efficient heat dissipation.

Careful attention should be paid to the layout of the output, VDD(VM) and GND traces, to avoid short circuits across output pins or to the power supply or ground. If such a short circuit occurs, the device may be permanently damaged.

Also, the utmost care should be taken for pattern designing and implementation of the device since it has power supply pins (VM, RS, OUT, GND) through which a particularly large current may run. If these pins are wired incorrectly, an operation error may occur or the device may be destroyed.

The logic input pins must also be wired correctly. Otherwise, the device may be damaged owing to a current running through the IC that is larger than the specified current.

Pin explanations

TB67S261FTG (QFN48)

Pin No.1 – 28

Pin No.	Pin Name	Function
1	NC	Non-connection pin
2	INB1	Motor Bch excitation control input 1
3	INB2	Motor Bch excitation control input 2
4	STANDBY	All-function-initializing and Low power dissipation mode
5	GND	Ground pin
6	NC	Non-connection pin
7	RSA(*)	Motor Ach current sense pin
8	RSA(*)	Motor Ach current sense pin
9	NC	Non-connection pin
10	OUTA+(*)	Motor Ach (+) output pin
11	OUTA+(*)	Motor Ach (+) output pin
12	NC	Non-connection pin
13	NC	Non-connection pin
14	NC	Non-connection pin
15	GND	Ground pin
16	OUTA-(*)	Motor Ach (-) output pin
17	OUTA-(*)	Motor Ach (-) output pin
18	GND	Ground pin
19	GND	Ground pin
20	OUTB-(*)	Motor Bch (-) output pin
21	OUTB-(*)	Motor Bch (-) output pin
22	GND	Ground pin
23	NC	Non-connection pin
24	NC	Non-connection pin
25	NC	Non-connection pin
26	OUTB+(*)	Motor Bch (+) output pin
27	OUTB+(*)	Motor Bch (+) output pin
28	NC	Non-connection pin

Pin No.29 – 48

Pin No.	Pin Name	Function
29	RSB(*)	Motor Bch current sense pin
30	RSB(*)	Motor Bch current sense pin
31	NC	Non-connection pin
32	VM	Motor power supply pin
33	NC	Non-connection pin
34	VCC	Internal VCC regulator monitor pin
35	NC	Non-connection pin
36	NC	Non-connection pin
37	NC	Non-connection pin
38	NC	Non-connection pin
39	NC	Non-connection pin
40	GND	Ground pin
41	VREFB	Motor Bch output set pin
42	VREFA	Motor Ach output set pin
43	OSCM	Oscillating circuit frequency for chopping set pin
44	INA1	Motor Ach excitation control input 1
45	INA2	Motor Ach excitation control input 2
46	PHASEA	Current direction signal input for motor Ach
47	PHASEB	Current direction signal input for motor Bch
48	NC	Non-connection pin

•Please do not run patterns under NC pins.

* : Please connect the pins with the same pin name, while using the TB67S261.

INPUT/OUTPUT equivalent circuit (TB67S261)

Pin name	IN/OUT signal	Equivalent circuit
INA1 INA2 PHASEA INB1 INB2 PHASEB STANDBY	Digital Input (VIH/VIL) VIH: 2.0V(min) to 5.5V(max) VIL : 0V(min) to 0.8V(max)	
VCC VREFA VREFB	VCC voltage range 4.75V(min) to 5.0V(typ.) to 5.25V(max) VREF voltage range 0V to 3.6V	
OSCM	OSCM frequency setting range 0.64MHz(min) to 1.12MHz(typ.) to 2.4MHz(max)	
OUT A+ OUT A- OUT B+ OUT B- RSA RSB	VM power supply voltage range 10V(min) to 47V(max) OUTPUT pin voltage 10V(min) to 47V(max)	

The equivalent circuit diagrams may be simplified or some parts of them may be omitted for explanatory purposes.

Function explanation (Stepping motor)

Motor output current (Iout) : The flow from OUT+ to OUT- is plus current. The flow from OUT- to OUT+ is minus current.

<Full step resolution>

Ach				Bch			
Input			Output	Input			Output
PHASEA	INA1	INA2	Iout(A)	PHASEB	INB1	INB2	Iout(B)
H	H	H	+100%	H	H	H	+100%
L	H	H	-100%	H	H	H	+100%
L	H	H	-100%	L	H	H	-100%
H	H	H	+100%	L	H	H	-100%

Please set INA1, INA2, INB1, and INB2 to Low until VM power supply reaches the proper operating range.

<Half step resolution>

Ach				Bch			
Input			Output	Input			Output
PHASEA	INA1	INA2	Iout(A)	PHASEB	INB1	INB2	Iout(B)
H	H	H	+100%	H	H	H	+100%
-	L	L	0%	H	H	H	+100%
L	H	H	-100%	H	H	H	+100%
L	H	H	-100%	-	L	L	0%
L	H	H	-100%	L	H	H	-100%
-	L	L	0%	L	H	H	-100%
H	H	H	+100%	L	H	H	-100%
H	H	H	+100%	-	L	L	0%

- : Don't care

<Quarter step resolution>

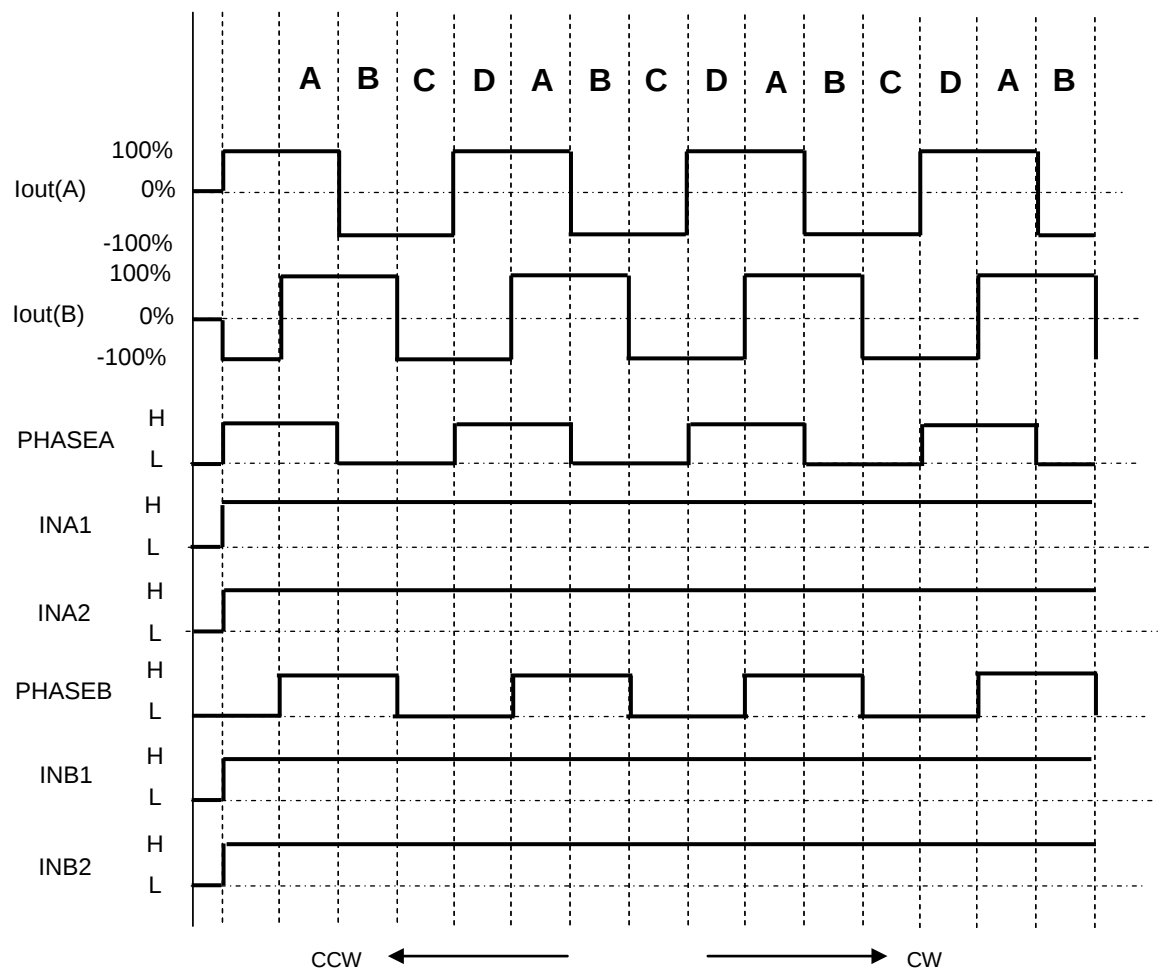
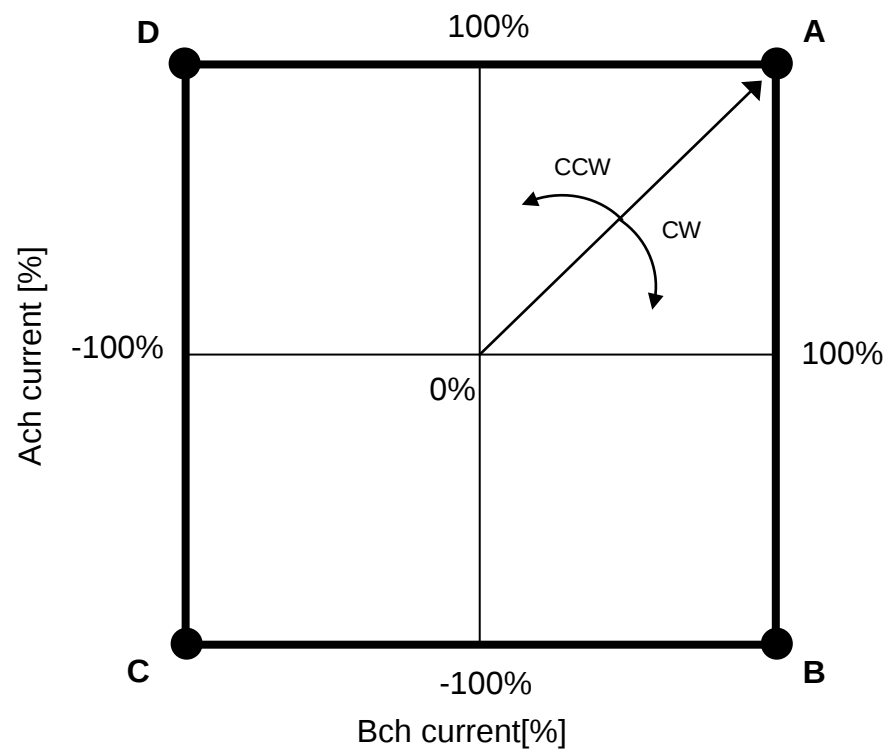
Ach				Bch			
Input			Output	Input			Output
PHASEA	INA1	INA2	Iout(A)	PHASEB	INB1	INB2	Iout(B)
H	H	L	+71%	H	H	L	+71%
H	L	H	+38%	H	H	H	+100%
X	L	L	0%	H	H	H	+100%
L	L	H	-38%	H	H	H	+100%
L	H	L	-71%	H	H	L	+71%
L	H	H	-100%	H	L	H	+38%
L	H	H	-100%	X	L	L	0%
L	H	H	-100%	L	L	H	-38%
L	H	L	-71%	L	H	L	-71%
L	L	H	-38%	L	H	H	-100%
X	L	L	0%	L	H	H	-100%
H	L	H	+38%	L	H	H	-100%
H	H	L	+71%	L	H	L	-71%
H	H	H	+100%	L	L	H	-38%
H	H	H	+100%	X	L	L	0%
H	H	H	+100%	H	L	H	+38%

X : Don't care

Others

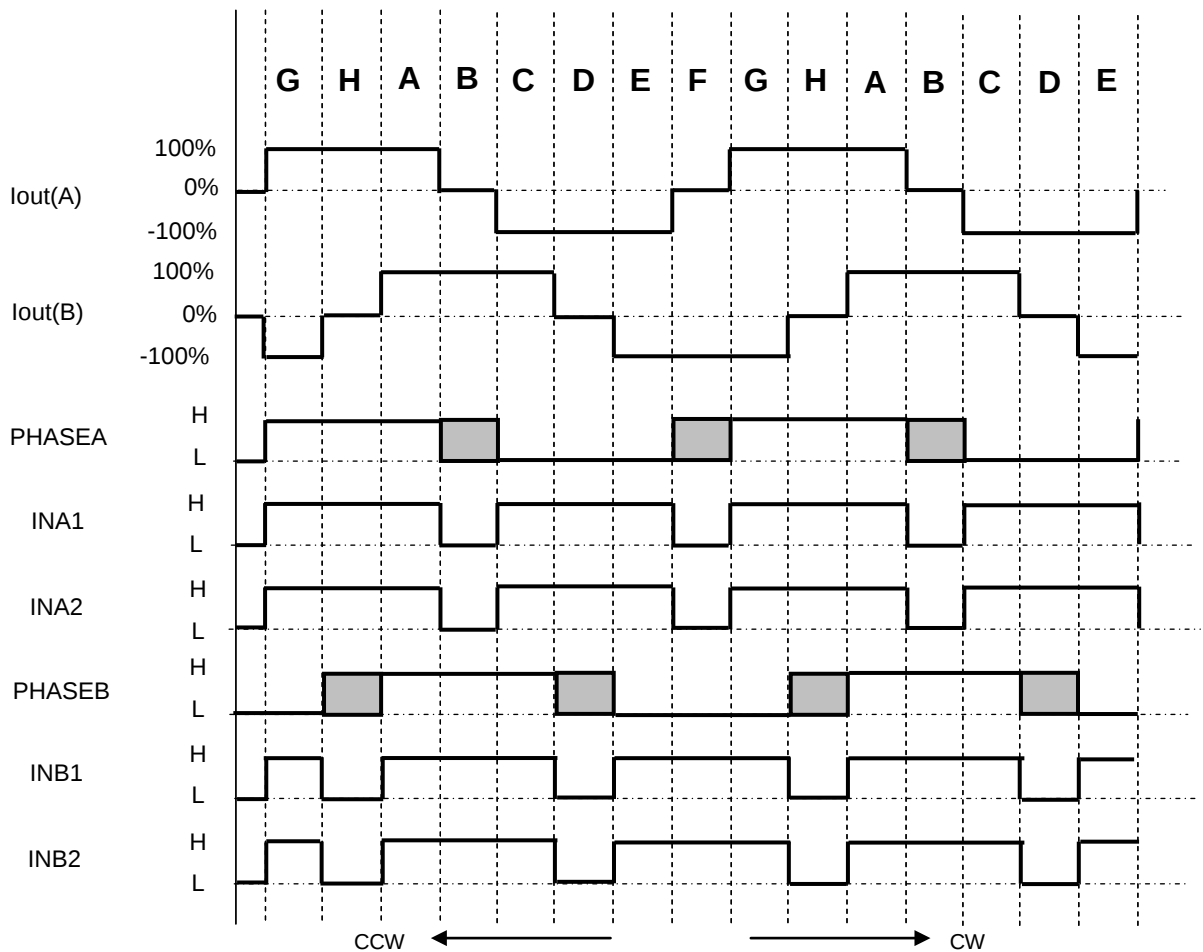
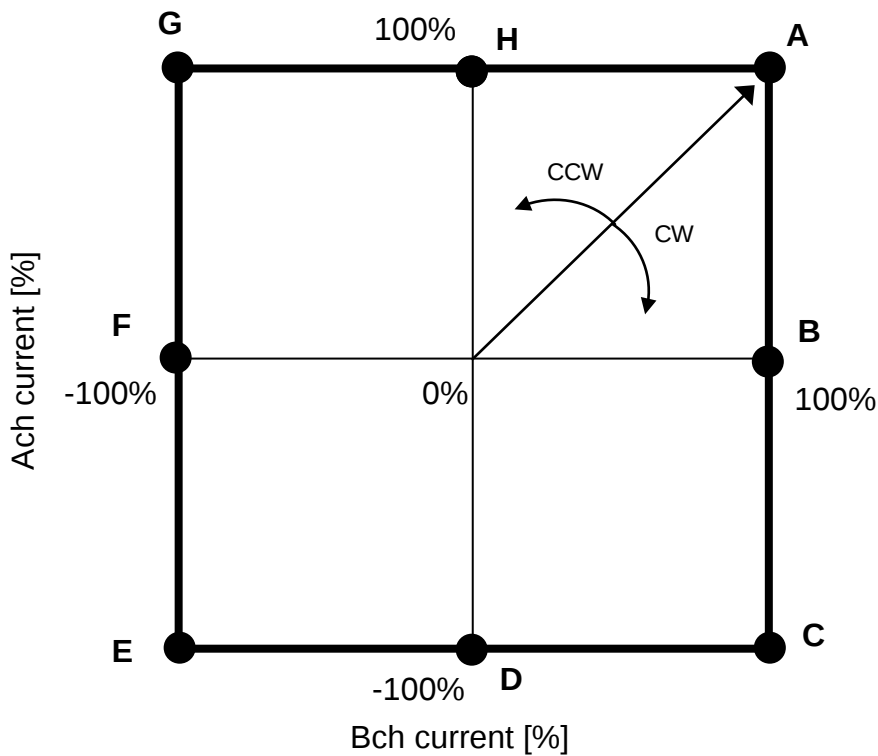
Pin Name	H	L	Notes
INA1, INA2 INB1, INB2	The current value of each ch is set up with 2 input 4 value.		Please refer to the above-mentioned current value setting table.
PHASEA PHASEB	OUT+: H OUT-: L	OUT+: L OUT-: H	In PHASE=H, Charge current flows in the direction of OUT- from OUT+.
STANDBY	Standby release	Standby mode	In STANDBY= L, an internal oscillating circuit and a motor output part are stopped. (The drive of a motor cannot be performed.)

Current phasor (Full step resolution)



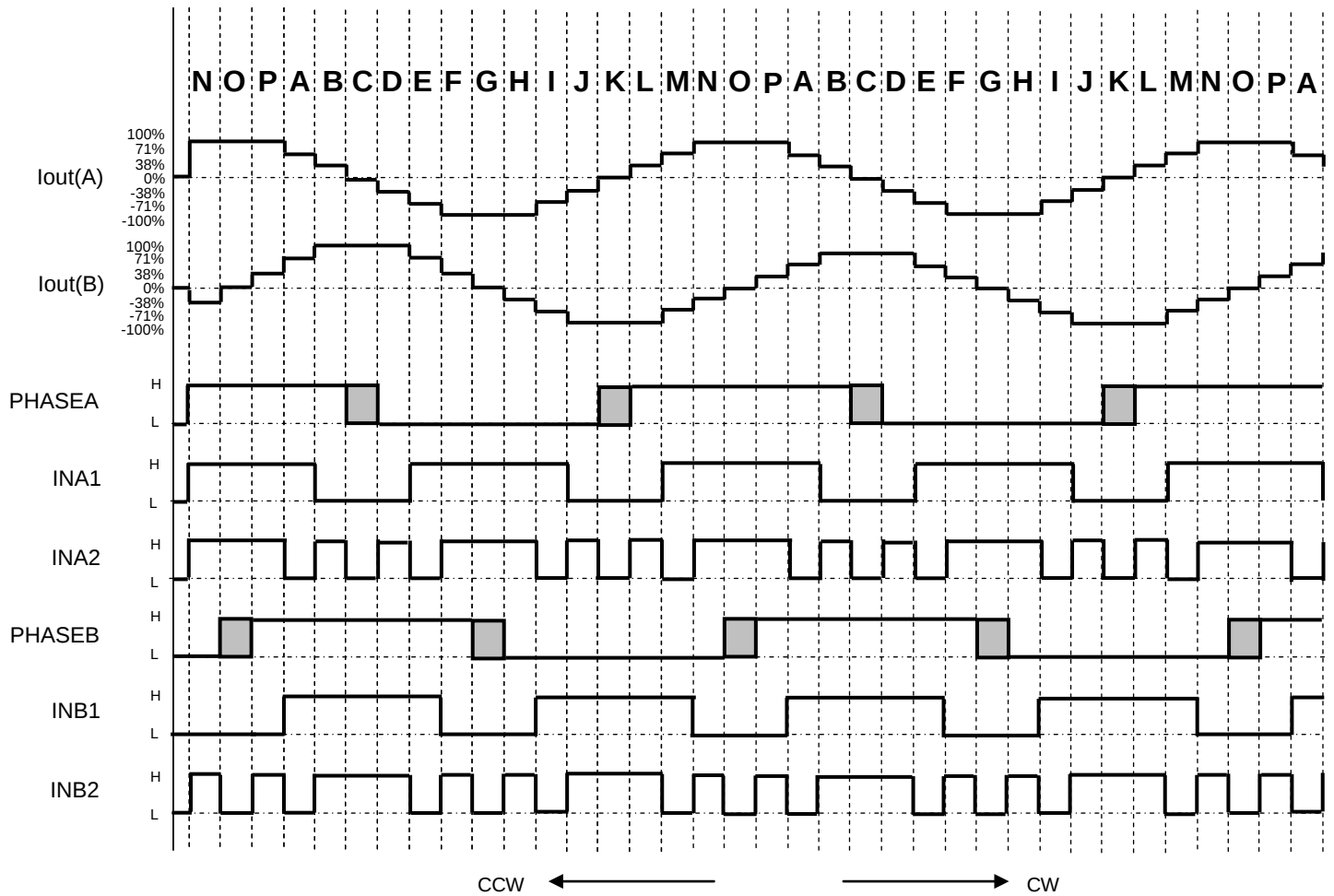
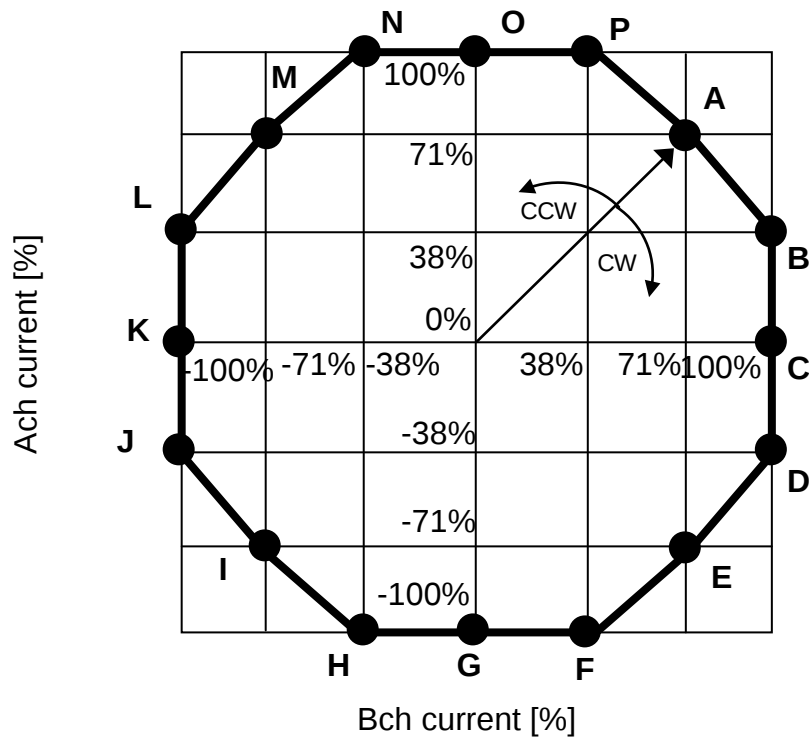
Timing charts may be simplified for explanatory purpose.
Please set INA1, INA2, INB1, and INB2 to Low until VM power supply reaches the proper operating range.

Current phasor (Half step resolution)



Timing charts may be simplified for explanatory purpose.
Please set INA1, INA2, INB1, and INB2 to Low until VM power supply reaches the proper operating range.

Current phasor (Quarter step resolution)

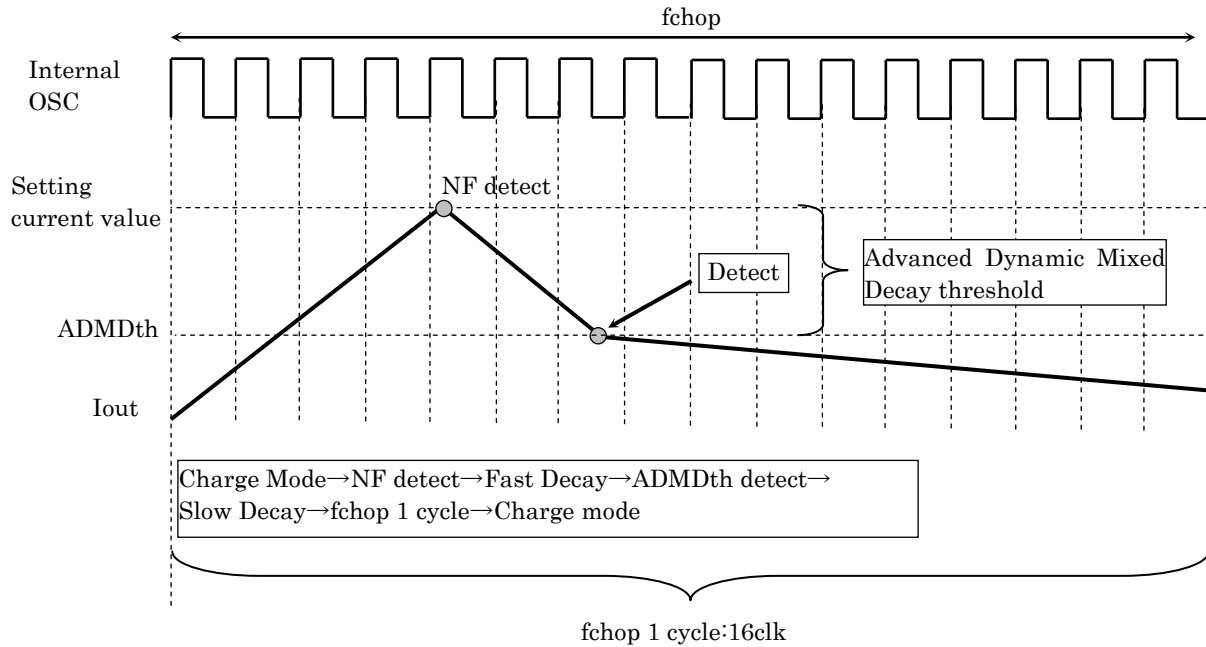


Timing charts may be simplified for explanatory purpose.
Please set INA1, INA2, INB1, and INB2 to Low until VM power supply reaches the proper operating range.

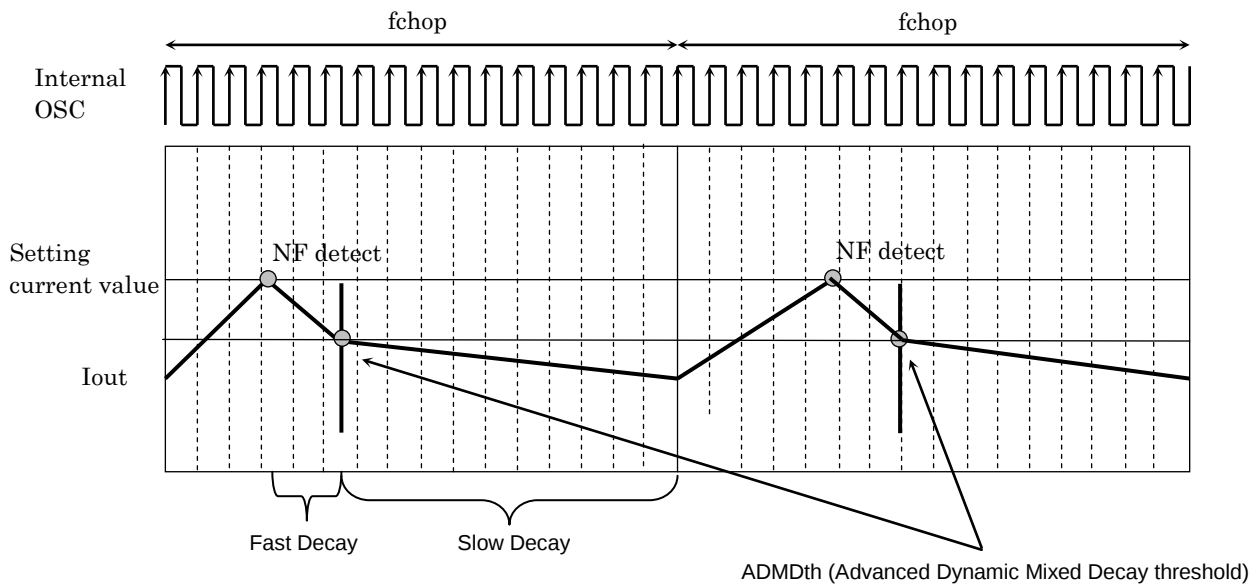
Decay function

ADMD(Advanced Dynamic Mixed Decay) constant current control

The Advanced Dynamic Mixed Decay threshold, which determines the current ripple level during current feedback control, is a unique value.



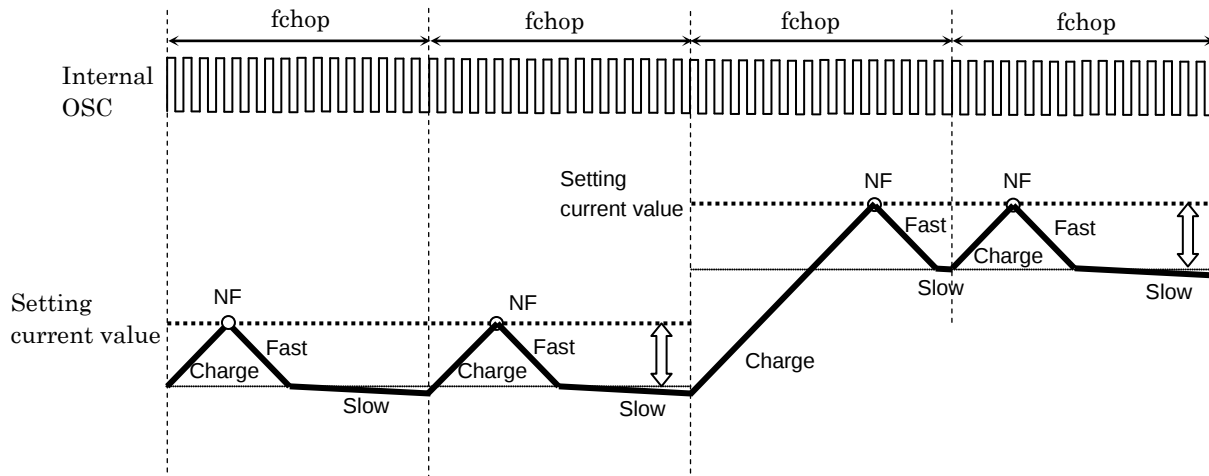
Auto Decay Mode current waveform



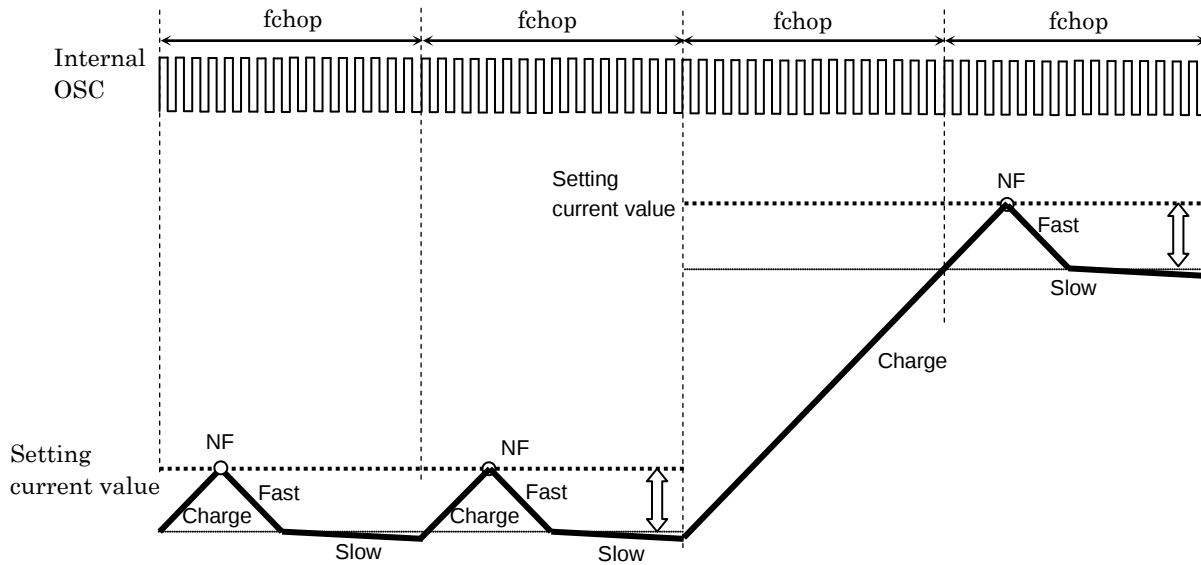
Timing charts may be simplified for explanatory purpose.

ADMD current waveform

•When the next current step is higher :



•When Charge period is more than 1 fchop cycle :



When the Charge period is longer than fchop cycle, the Charge period will be extended until the motor current reaches the NF threshold. Once the current reaches the next current step, then the sequence will go on to ADMD control.

Internal OSC

Setting current value

NF

Charge

Fast

Slow

Charge

NF

Fast

Slow

Charge

Fast

Slow

Setting current value

f_{chop}

f_{chop}

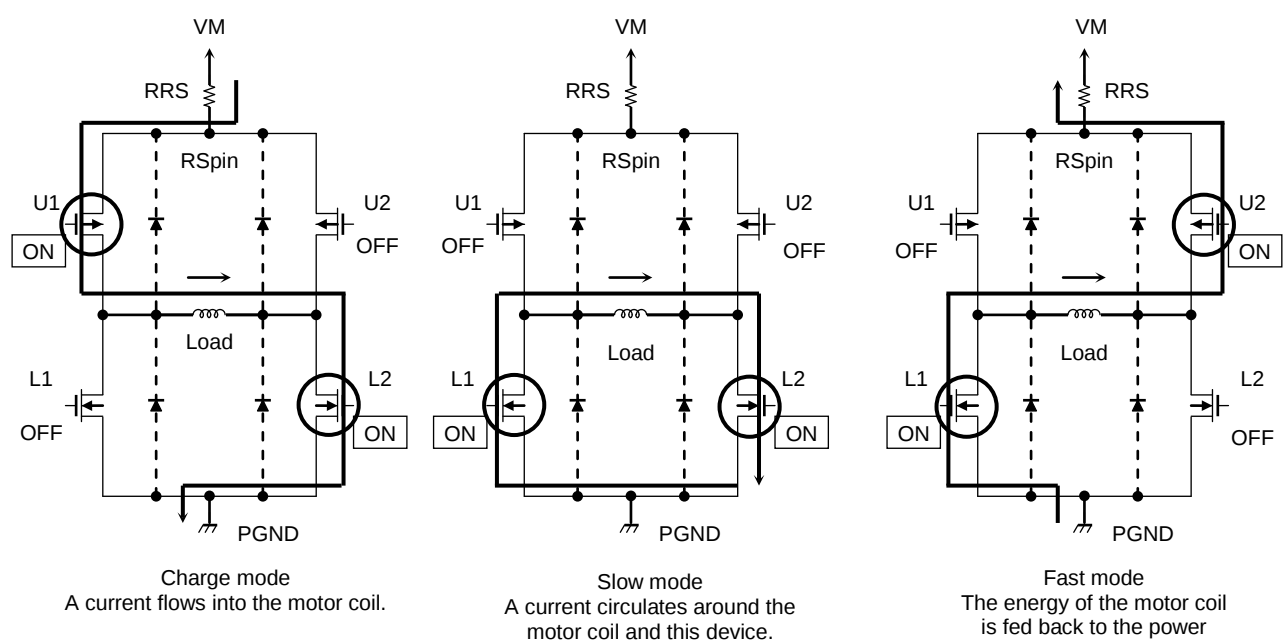
f_{chop}

f_{chop}

The operation mode will be switched to 'Charge' to monitor the motor current with the RS comparator; then will be switched to 'Fast' because the motor current is above the threshold.

If the motor current is still above the ADMD threshold after reaching 1 f_{chop} cycle, the output stage function will stay 'Fast' until the current reaches the ADMDth.

Output transistor function mode



Output transistor function

MODE	U1	U2	L1	L2
CHARGE	ON	OFF	OFF	ON
SLOW	OFF	OFF	ON	ON
FAST	OFF	ON	ON	OFF

Note: This table shows an example of when the current flows as indicated by the arrows in the figures shown above. If the current flows in the opposite direction, refer to the following table.

MODE	U1	U2	L1	L2
CHARGE	OFF	ON	ON	OFF
SLOW	OFF	OFF	ON	ON
FAST	ON	OFF	OFF	ON

This IC controls the motor current to be constant by 3 modes listed above.

The equivalent circuit diagrams may be simplified or some parts of them may be omitted for explanatory purposes.

Calculation of the Predefined Output Current

For PWM constant-current control, this IC uses a clock generated by the OSCM oscillator.

The peak output current (Setting current value) can be set via the current-sensing resistor (RS) and the reference voltage (Vref), as follows:

$$I_{out(max)} = V_{ref(gain)} \times \frac{V_{ref(V)}}{R_{RS}(\Omega)}$$

Vref(gain) : the Vref decay rate is 1/ 5.0 (typ.)

For example : In the case of a 100% setup

when Vref = 3.0 V, Torque=100%, RS=0.51Ω, the motor constant current (Setting current value) will be calculated as:

$$I_{out} = 3.0V / 5.0 / 0.51\Omega = 1.18 \text{ A}$$

Calculation of the OSCM oscillation frequency (chopper reference frequency)

An approximation of the OSCM oscillation frequency (fOSCM) and chopper frequency (fchop) can be calculated by the following expressions.

$$f_{OSCM} = 1 / [0.56 \times \{C_x(R1 + 500)\}]$$

.....C, R1: External components for OSCM (C=270pF, R1=5.1kΩ => fOSCM= 1.12MHz(Typ.))

$$f_{chop} = f_{OSCM} / 16$$

.....fOSCM=1.12MHz => fchop =About 70kHz

If chopping frequency is raised, Rippl of current will become small and wave-like reproducibility will improve. However, the gate loss inside IC goes up and generation of heat becomes large.

By lowering chopping frequency, reduction in generation of heat is expectable. However, Rippl of current may become large. It is a standard about about 70 kHz. A setup in the range of 50 to 100 kHz is recommended.

Absolute Maximum Ratings (Ta = 25°C)

Characteristics	Symbol	Rating	Unit	Remarks
Motor power supply	VM	50	V	-
Motor output voltage	Vout	50	V	-
Motor output current	Iout	2.0	A	(Note 1)
Internal Logic power supply	VCC	6.0	V	When externally applied.
Logic input voltage	VIN(H)	6.0	V	-
	VIN(L)	-0.4	V	-
Vref reference voltage	Vref	5.0	V	-
Power dissipation	WQFN48	PD	1.3	W
				(Note 2)
Operating temperature	TOPR	-20 to 85	°C	-
Storage temperature	TSTG	-55 to 150	°C	-
Junction temperature	Tj(max)	150	°C	-

Note 1: Usually, the maximum current value at the time should use 70% or less of the absolute maximum ratings for a standard on thermal rating. The maximum output current may be further limited in view of thermal considerations, depending on ambient temperature and board conditions.

Note 2: Device alone (Ta =25°C)

When Ta exceeds 25°C, it is necessary to do the derating with 10.4mW/°C.

Ta: Ambient temperature

Topr: Ambient temperature while the IC is active

Tj: Junction temperature while the IC is active. The maximum junction temperature is limited by the thermal shutdown (TSD) circuitry. It is advisable to keep the maximum current below a certain level so that the maximum junction temperature, Tj (MAX), will not exceed 120°C.

Caution)Absolute maximum ratings

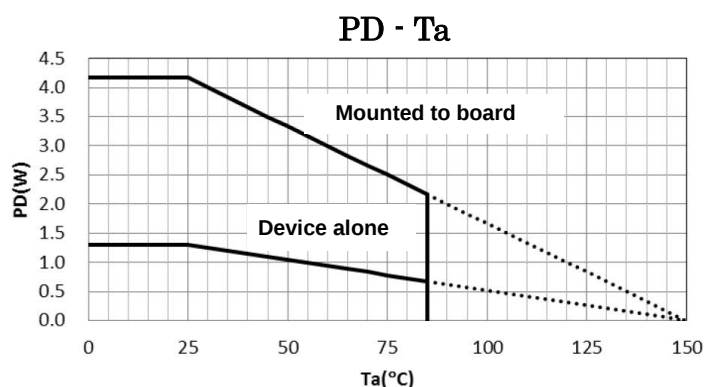
The absolute maximum ratings of a semiconductor device are a set of ratings that must not be exceeded, even for a moment. Do not exceed any of these ratings.

Exceeding the rating (s) may cause device breakdown, damage or deterioration, and may result in injury by explosion or combustion.

The value of even one parameter of the absolute maximum ratings should not be exceeded under any circumstances. The TB67S261FTG does not have overvoltage detection circuit. Therefore, the device is damaged if a voltage exceeding its rated maximum is applied.

All voltage ratings, including supply voltages, must always be followed. The other notes and considerations described later should also be referred to.

(For reference) PD-Ta graph



Board condition

4 layer glass epoxy board

Cu thickness:1 layer and 4 layer:55μm,2 layer and 3 layer:35μm

Board size:100mm×110mm×1.6mm

Operation Ranges (Ta=-20 to 85°C)

Characteristics	Symbol	Min	Typ.	Max	Unit	Remarks
Motor power supply	VM	10	24	47	V	-
Motor output current	Iout	-	1.4	2.0	A	(Note 1)
Logic input voltage	VIN(H)	2.0	-	5.5	V	Logic input High Level
	VIN(L)	0	-	0.8	V	Logic input Low Level
Phase input frequency	fPHASE	-	-	400	kHz	-
Chopper frequency	fchop(range)	40	70	150	kHz	-
Vref input voltage	Vref	GND	2.0	3.6	V	-

Note 1: Maximum current for actual usage may be limited by the operating circumstances such as operating conditions (exciting mode, operating time, and so on), ambient temperature, and heat conditions (board condition and so on).

Electrical Specifications 1 (Ta = 25°C, VM = 24 V, unless specified otherwise)

Characteristics		Symbol	Test condition	Min	Typ.	Max	Unit
Logic input voltage	HIGH	VIN(H)	Logic input (Note)	2.0	-	5.5	V
	LOW	VIN(L)	Logic input (Note)	0	-	0.8	V
Logic input hysteresis voltage		VIN(HYS)	Logic input (Note)	100	-	300	mV
Logic input current	HIGH	IIN(H)	VIN(H)=3.3V	-	33	-	μA
	LOW	IIN(L)	VIN(L)=0V	-	-	1	μA
Power consumption		IM1	Output pins=open STANDBY=L	-	2	3.5	mA
		IM2	Output pins=open STANDBY=H	-	3.5	5.5	mA
		IM3	Output pins=open Full step resolution	-	5.5	7	mA
Output leakage current	High-side	IOH	VRS=VM=50V, Vout=0V	-	-	1	μA
	Low-side	IOL	VRS=VM=Vout=50V	1	-	-	μA
Motor current channel differential		ΔIout1	Current differential between Ch	-5	0	5	%
Motor current setting accuracy		ΔIout2	Iout=1.5A	-5	0	5	%
RS pin current		IRS	VRS=VM=24V	0	-	10	μA
Motor output ON-resistance (High-side+Low-side)		Ron(H+L)	Tj=25°C, Forward direction (High-side+Low-side)	—	0.8	0.9	Ω

Note: VIN (H) is defined as the VIN voltage that causes the outputs (OUTA, OUTB) to change when a pin under test is gradually raised from 0 V. VIN (L) is defined as the VIN voltage that causes the outputs (OUTA, OUTB) to change when the pin is then gradually lowered from 5V. The difference between VIN (H) and VIN (L) is defined as the VIN (HYS).

Note: When the logic signal is applied to the device whilst the VM power supply is not asserted; the device is designed not to function, but for safe usage, please apply the logic signal after the VM power supply is asserted and the VM voltage reaches the proper operating range.

Electrical Specifications 2 (Ta =25°C, VM = 24 V, unless specified otherwise)

Characteristics	Symbol	Test condition	Min	Typ.	Max	Unit
Vref input current	Iref	VREF=2.0V	-	0	1	μA
VCC voltage	VCC	ICC=5.0mA	4.75	5.0	5.25	V
VCC current	ICC	VCC=5.0V	-	2.5	5	mA
Vref gain rate	Vref(gain)	Vref=2.0V	1/5.2	1/5.0	1/4.8	—
Thermal shutdown(TSD) threshold (Note1)	TjTSD	—	145	160	175	°C
VM recovery voltage	VMR	—	7.0	8.0	9.0	V
Over-current detection (ISD) threshold (Note2)	ISD	—	2.1	3.0	4.0	A

Note1: About TSD

When the junction temperature of the device reached the TSD threshold, the TSD circuit is triggered; the internal reset circuit then turns off the output transistors. Noise rejection blanking time is built-in to avoid misdetection. Once the TSD circuit is triggered, the device will be set to standby mode, and can be cleared by reasserting the VM power source, or setting the DMODE pins to standby mode. The TSD circuit is a backup function to detect a thermal error, therefore is not recommended to be used aggressively.

Note2: About ISD

When the output current reaches the threshold, the ISD circuit is triggered; the internal reset circuit then turns off the output transistors. Once the ISD circuit is triggered, the device keeps the output off until power-on reset (POR), is reasserted or the device is set to standby mode by DMODE pins. For fail-safe, please insert a fuse to avoid secondary trouble.

Back-EMF

While a motor is rotating, there is a timing at which power is fed back to the power supply. At that timing, the motor current recirculates back to the power supply due to the effect of the motor back-EMF.

If the power supply does not have enough sink capability, the power supply and output pins of the device might rise above the rated voltages. The magnitude of the motor back-EMF varies with usage conditions and motor characteristics.

Cautions on Overcurrent Shutdown (ISD) and Thermal Shutdown (TSD)

The ISD and TSD circuits are only intended to provide temporary protection against irregular conditions such as an output short-circuit; they do not necessarily guarantee the complete IC safety.

If the device is used beyond the specified operating ranges, these circuits may not operate properly; then the device may be damaged due to an output short-circuit.

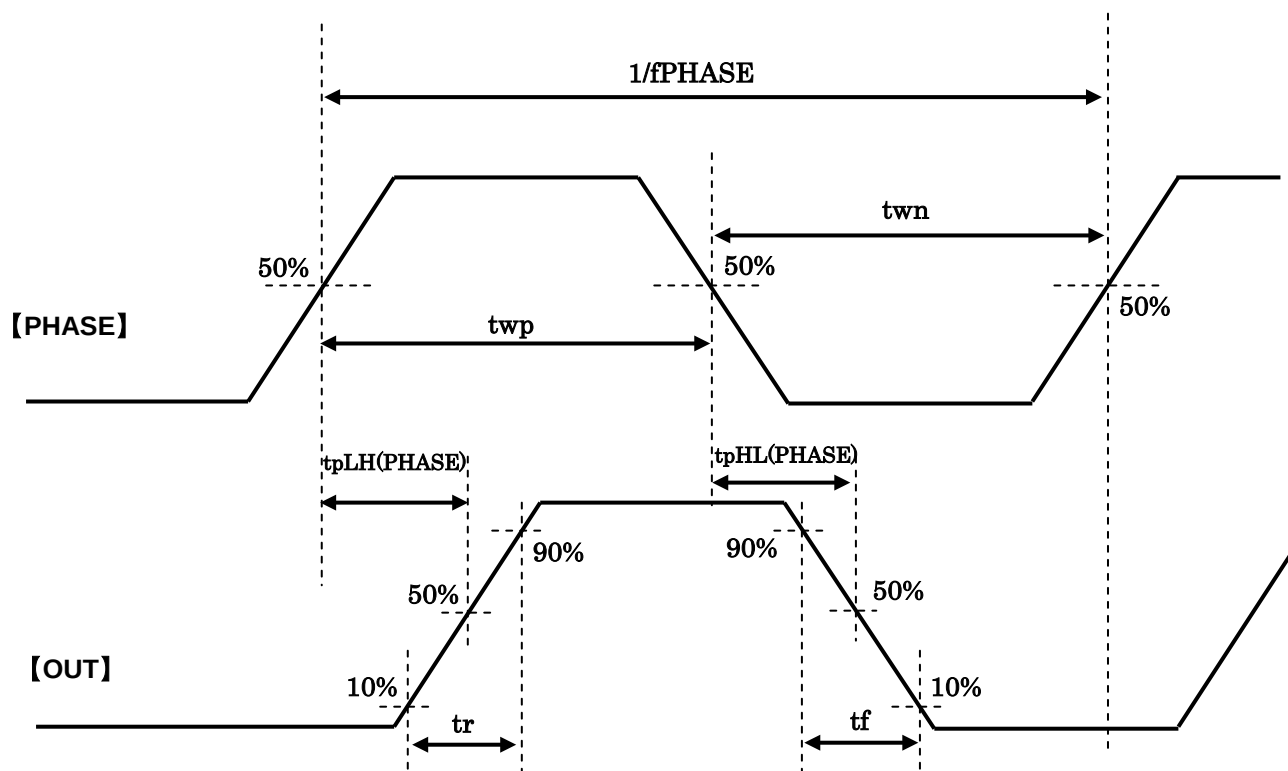
The ISD circuit is only intended to provide a temporary protection against an output short-circuit. If such a condition persists for a long time, the device may be damaged due to overstress. Overcurrent conditions must be removed immediately by external hardware.

IC Mounting

Do not insert devices incorrectly or in the wrong orientation. Otherwise, it may cause breakdown, damage and/or deterioration of the device.

AC Electrical Specification (Ta = 25°C, VM = 24 V, 6.8 mH/5.7 Ω)

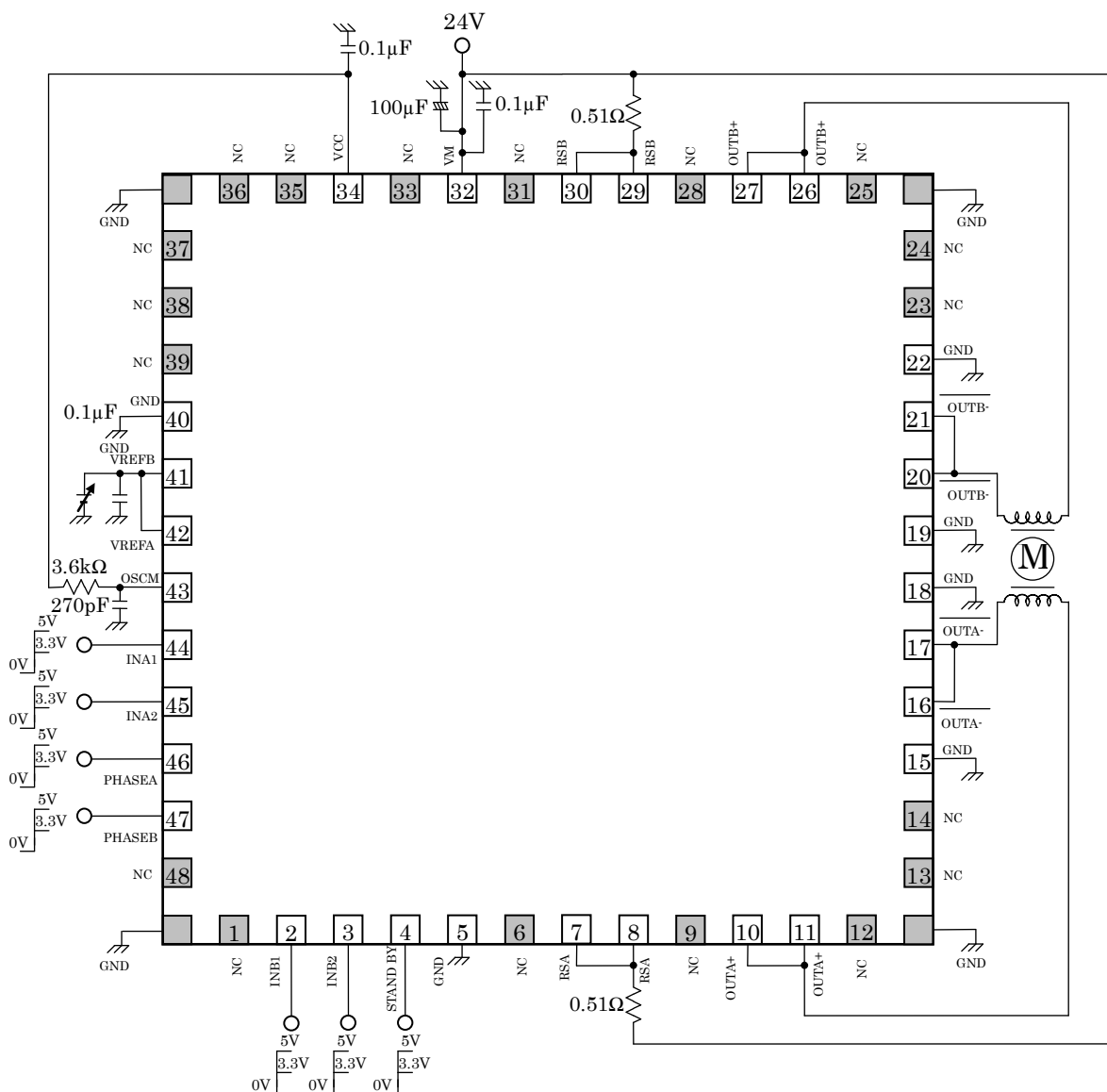
Characteristics	Symbol	Test condition	Min	Typ.	Max	Unit
Minimum PHASE pulse width	fPHASE(min)	—	100	-	-	ns
	twp	—	50	-	-	
	twl	—	50	-	-	
Output transistor switching specific	tr	—	30	80	130	ns
	tf	—	40	90	140	
	tpLH(PHASE)	PHASE - Output	250	-	1200	
	tpHL(PHASE)	PHASE - Output	250	-	1200	
Analog noise blanking time	AtBLK	VM=24V, Iout=1.5A Analog tblank	250	400	550	ns
Oscillator frequency accuracy	Δf_{OSCM}	COSC=270pF, ROSC=5.1kΩ	-15	-	+15	%
Oscillator reference frequency	fOSCM	COSC=270pF, ROSC=5.1kΩ	952	1120	1288	kHz
Chopping frequency	fchop	Output:Active(IOUT =1.5 A), fOSCM = 1120 kHz	-	70	-	kHz

AC Electrical Specification Timing chart


Timing charts may be simplified for explanatory purpose.

Example Application Circuits

The values shown in the following figure are typical values. For input conditions, see the Operating Ranges.



Note: I will recommend the addition of a capacitor if necessary. The GND wiring must become one point as much as possible-earth.

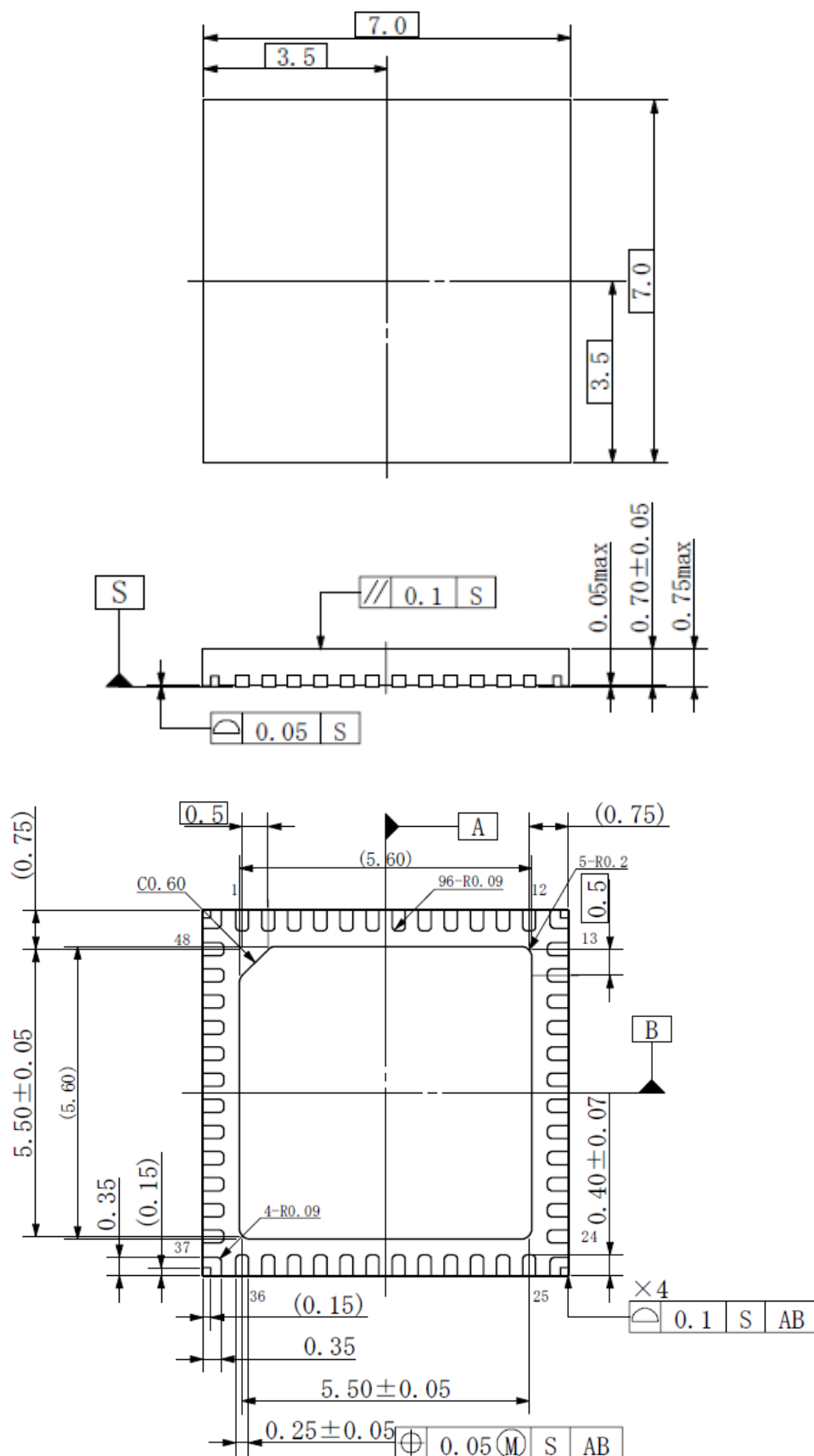
The example of an applied circuit is for reference, and enough evaluation should be done before the mass-production design.

Moreover, it is not the one to permit the use of the industrial property.

Package Dimensions

(unit :mm)

P-WQFN48-0707-0.50-003



Weight 0.10g (typ.)

Notes on Contents**Block Diagrams**

Some of the functional blocks, circuits, or constants in the block diagram may be omitted or simplified for explanatory purposes.

Equivalent Circuits

The equivalent circuit diagrams may be simplified or some parts of them may be omitted for explanatory purposes.

Timing Charts

Timing charts may be simplified for explanatory purposes.

Application Circuits

The application circuits shown in this document are provided for reference purposes only. Thorough evaluation is required, especially at the mass-production design stage.

Toshiba does not grant any license to any industrial property rights by providing these examples of application circuits.

Test Circuits

Components in the test circuits are used only to obtain and confirm the device characteristics. These components and circuits are not guaranteed to prevent malfunction or failure from occurring in the application equipment.

IC Usage Considerations**Notes on handling of ICs**

- (1) The absolute maximum ratings of a semiconductor device are a set of ratings that must not be exceeded, even for a moment. Do not exceed any of these ratings. Exceeding the rating(s) may cause device breakdown, damage or deterioration, and may result in injury by explosion or combustion.
- (2) Do not insert devices in the wrong orientation or incorrectly. Make sure that the positive and negative terminals of power supplies are connected properly.
Otherwise, the current or power consumption may exceed the absolute maximum rating, and exceeding the rating(s) may cause device breakdown, damage or deterioration, and may result in injury by explosion or combustion.
In addition, do not use any device inserted in the wrong orientation or incorrectly to which current is applied even just once.
- (3) Use an appropriate power supply fuse to ensure that a large current does not continuously flow in the case of overcurrent and/or IC failure. The IC will fully break down when used under conditions that exceed its absolute maximum ratings, when the wiring is routed improperly or when an abnormal pulse noise occurs from the wiring or load, causing a large current to continuously flow and the breakdown can lead to smoke or ignition. To minimize the effects of the flow of a large current in the case of breakdown, appropriate settings, such as fuse capacity, fusing time and insertion circuit location, are required.
- (4) If your design includes an inductive load such as a motor coil, incorporate a protection circuit into the design to prevent device malfunction or breakdown caused by the current resulting from the inrush current at power ON or the negative current resulting from the back electromotive force at power OFF. IC breakdown may cause injury, smoke or ignition. Use a stable power supply with ICs with built-in protection functions. If the power supply is unstable, the protection function may not operate, causing IC breakdown. IC breakdown may cause injury, smoke or ignition.
- (5) Carefully select external components (such as inputs and negative feedback capacitors) and load components (such as speakers), for example, power amp and regulator.
If there is a large amount of leakage current such as from input or negative feedback condenser, the IC output DC voltage will increase. If this output voltage is connected to a speaker with low input withstand voltage, overcurrent or IC failure may cause smoke or ignition. (The overcurrent may cause smoke or ignition from the IC itself.) In particular, please pay attention when using a Bridge Tied Load (BTL) connection-type IC that inputs output DC voltage to a speaker directly.

Points to remember on handling of ICs**Overcurrent detection Circuit**

Overcurrent detection circuits (referred to as current limiter circuits) do not necessarily protect ICs under all circumstances. If the overcurrent detection circuits operate against the overcurrent, clear the overcurrent status immediately.

Depending on the method of use and usage conditions, exceeding absolute maximum ratings may cause the overcurrent detection circuit to operate improperly or IC breakdown may occur before operation. In addition, depending on the method of use and usage conditions, if overcurrent continues to flow for a long time after operation, the IC may generate heat resulting in breakdown.

Thermal Shutdown Circuit

Thermal shutdown circuits do not necessarily protect ICs under all circumstances. If the thermal shutdown circuits operate against the over-temperature, clear the heat generation status immediately.

Depending on the method of use and usage conditions, exceeding absolute maximum ratings may cause the thermal shutdown circuit to operate improperly or IC breakdown to occur before operation.

Heat Radiation Design

When using an IC with large current flow such as power amp, regulator or driver, design the device so that heat is appropriately radiated, in order not to exceed the specified junction temperature (T_J) at any time or under any condition. These ICs generate heat even during normal use. An inadequate IC heat radiation design can lead to decrease in IC life, deterioration of IC characteristics or IC breakdown. In addition, when designing the device, take into consideration the effect of IC heat radiation with peripheral components.

Back-EMF

When a motor rotates in the reverse direction, stops or slows abruptly, current flows back to the motor's power supply owing to the effect of back-EMF. If the current sink capability of the power supply is small, the device's motor power supply and output pins might be exposed to conditions beyond the absolute maximum ratings. To avoid this problem, take the effect of back-EMF into consideration in system design.

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