

SINGLE-CELL LI-ION CHARGE MANAGEMENT IC WITH TIMER-ENABLE FOR PDAs AND INTERNET APPLIANCES

FEATURES

- **Highly Integrated Solution With FET Pass Transistor and Reverse-Blocking Schottky and Thermal Protection**
- **Timer-Enable Function That Allows Host to Disable Charge Timer and Termination When Charge Current Is Shared With a Load or When Battery Is Absent**
- **Integrated Voltage and Current Regulation With Programmable Charge Current**
- **High-Accuracy Voltage Regulation ($\pm 1\%$)**
- **Ideal for Low-Dropout Linear Charger Designs for Single-Cell Li-Ion Packs With Coke or Graphite Anodes**
- **Up to 1.2-A Continuous Charge Current**
- **Safety-Charge Timer During Preconditioning and Fast Charge**
- **Integrated Cell Conditioning for Reviving Deeply Discharged Cells and Minimizing Heat Dissipation During Initial Stage of Charge**
- **Optional Temperature or Input-Power Monitoring Before and During Charge**
- **Various Charge-Status Output Options for Driving Single, Double, or Bicolor LEDs or Host-Processor Interface**
- **Charge Termination by Minimum Current and Time**
- **Low-Power Sleep Mode**
- **Packaging: 5 mm $\times$ 5 mm MLP (bq24007) or 20-Lead TSSOP PowerPAD™**

APPLICATIONS

- **PDAs**
- **Internet Appliances**
- **MP3 Players**
- **Digital Cameras**

DESCRIPTION

The bq2400x series ICs are advanced Li-Ion linear charge management devices for highly integrated and space-limited applications. They combine high-accuracy current and voltage regulation; FET pass-transistor and reverse-blocking Schottky; battery conditioning, temperature, or input-power monitoring; charge termination; charge-status indication; and charge timer in a small package.

The bq24007 and bq24008 provide the timer-enable function, allowing the host to disable the charge timer when charge current is shared with a load or when the battery is absent. This feature is ideal for applications such as cellular phones, PDAs, and internet appliances.

The bq2400x measures battery temperature using an external thermistor. For safety reasons, the bq2400x inhibits charge until the battery temperature is within the user-defined thresholds. Alternatively, the user can monitor the input voltage to qualify charge. The bq2400x series then charge the battery in three phases: preconditioning, constant current, and constant voltage. If the battery voltage is below the internal low-voltage threshold, the bq2400x uses low-current precharge to condition the battery. A preconditioning timer provides additional safety. Following preconditioning, the bq2400x applies a constant-charge current to the battery. An external sense-resistor sets the magnitude of the current. The constant-current phase is maintained until the battery reaches the charge-regulation voltage. The bq2400x then transitions to the constant voltage phase. The user can configure the device for cells with either coke or graphite anodes. The accuracy of the voltage regulation is better than $\pm 1\%$ over the operating junction temperature and supply voltage range.

Charge is terminated by maximum time or minimum taper current detection

The bq2400x automatically restarts the charge if the battery voltage falls below an internal recharge threshold.



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

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These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

ORDERING INFORMATION

T_J	PACKAGE		CHARGE STATUS CONFIGURATION
	20-LEAD HTTSOP PowerPAD™ (PWP)(1)	20-LEAD 5 mm × 5 mm MLP (RGW)(2)	
–40°C to 125°C	bq24007PWP	bq24007RGW	Single LED
	bq24008PWP	Not available	Single bicolor LED

(1) The PWP package is available taped and reeled. Add R suffix to device type (e.g. bq24007PWPR) to order. Quantities 2500 devices per reel.

(2) The RGW package is available taped and reeled. Add R suffix to device type (e.g. bq24007RGWR) to order. Quantities 3000 devices per reel.

PACKAGE DISSIPATION RATINGS

PACKAGE	Θ_{JA}	Θ_{JC}	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$
PWP(1)	30.88°C/W	1.19°C/W	3.238 W	0.0324W/°C
RGW(2)	31.41°C/W	1.25°C/W	3.183 W	0.0318W/°C

(1) This data is based on using the JEDEC high-K board and topside traces, top and bottom thermal pad (6,5 × 3,4 mm), internal 1 oz power and ground planes, 8 thermal via underneath the die connecting to ground plane.

(2) This data is based on using the JEDEC high-K board and topside traces, top and bottom thermal pad (3,25 × 3,25 mm), internal 1 oz power and ground planes, 9 thermal via underneath the die connecting to ground plane.

ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range unless otherwise noted(1)

	bq24007 bq24008
Supply voltage (V_{CC} with respect to GND)	13.5 V
Input voltage (IN, ISNS, TMR EN, APG/THERM/CR/STAT1/STAT2, VSENSE, TMR SEL, VSEL) (all with respect to GND)	13.5 V
Output current (OUT pins)	2 A
Output sink/source current (STAT1 and STAT2)	10 mA
Operating free-air temperature range, T_A	–40°C to 70°C
Storage temperature range, T_{stg}	–65°C to 150°C
Junction temperature range, T_J	–40°C to 125°C
Lead temperature (Soldering, 10 s)	300°C

(1) Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “recommended operating conditions” is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

RECOMMENDED OPERATING CONDITIONS

	MIN	MAX	UNIT
Supply voltage, V_{CC}	4.5	10	V
Input voltage, V_{IN}	4.5	10	V
Continuous output current		1.2	A
Operating junction temperature range, T_J	–40	125	°C

ELECTRICAL CHARACTERISTICS

over recommended operating junction temperature supply and input voltages, and $V_I (V_{CC}) \geq V_I (IN)$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{CC} current	$V_{CC} > V_{CC_UVLO}$			1	mA
Standby current (sum of currents into OUT and VSENSE pins)	$V_{CC} < V_{CC_UVLO}$, $V_{O(OUT)} = 4.3\text{ V}$, $V_{SENSE} = 4.3\text{ V}$		2	4	μA

VOLTAGE REGULATION, $0^\circ\text{C} \leq T_J \leq 125^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Output voltage	$V_{SEL} = V_{SS}$, $0 < I_O \leq 1.2\text{ A}$	4.059	4.10	4.141	V
	$V_{SEL} = V_{CC}$, $0 < I_O \leq 1.2\text{ A}$	4.158	4.20	4.242	V
Load regulation	$1\text{ mA} \leq I_O \leq 1.2\text{ A}$, $V_{I(IN)} = 5\text{ V}$, $V_{CC} = 5\text{ V}$, $T_J = 25^\circ\text{C}$		1		mV
Line regulation	$V_{O(OUT)} + V_{DO} + V_{ilim(MAX)} < V_I(V_{CC}) < 10\text{ V}$, $T_J = 25^\circ\text{C}$		0.01		%/V
Dropout voltage = $V_I(IN) - V_{O(OUT)}$	$I_O = 1.0\text{ A}$, $4.9\text{ V} < V_I(V_{CC}) < 10\text{ V}$			0.7	V
	$I_O = 1.2\text{ A}$, $V_{O(OUT)} + V_{DO} + V_{ilim(MAX)} < V_I(V_{CC}) < 10\text{ V}$			0.8	V

CURRENT REGULATION, $0^\circ\text{C} \leq T_J \leq 125^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Current regulation threshold	$V_{SENSE} < V_{O(V_{SEL-LOW/HIGH})}$	0.095	0.1	0.105	V
Delay time	V_{SENSE} pulsed above V_{LOWV} to $I_O = 10\%$ of regulated value ⁽¹⁾			1	ms
Rise time	I_O increasing from 10% to 90% of regulated value. $R_{(SNS)} \geq 0.2\ \Omega$, ⁽¹⁾	0.1		1	ms

(1) Specified by design, not production tested.

CURRENT SENSE RESISTOR, $0^\circ\text{C} \leq T_J \leq 125^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
External current sense resistor range ($R_{(SNS)}$)	$100\text{ mA} \leq i_{lim} \leq 1.2\text{ A}$	0.083		1	Ω

PRECHARGE CURRENT REGULATION, $0^\circ\text{C} \leq T_J \leq 125^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Precharge current regulation	$V_{SENSE} < V_{LOWV}$, $0.083 \leq R_{(SNS)} \leq 1.0\ \Omega$	40	60	80	mA

V_{CC} UVLO COMPARATOR, $0^\circ\text{C} \leq T_J \leq 125^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Start threshold		4.35	4.43	4.50	V
Stop threshold		4.25	4.33	4.40	V
Hysteresis		50			mV

APG/THERM COMPARATOR, $0^\circ\text{C} \leq T_J \leq 125^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Upper trip threshold		1.480	1.498	1.515	V
Lower trip threshold		0.545	0.558	0.570	V
Input bias current				1	μA

LOWV COMPARATOR, $0^\circ\text{C} \leq T_J \leq 125^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Start threshold		2.80	2.90	3.00	V
Stop threshold		3.00	3.10	3.20	V
Hysteresis		100			mV

ELECTRICAL CHARACTERISTICS CONTINUEDover recommended operating junction temperature supply and input voltages, and $V_I (V_{CC}) \geq V_I (IN)$ (unless otherwise noted)

HIGHV (RECHARGE) COMPARATOR, $0^{\circ}\text{C} \leq T_J \leq 125^{\circ}\text{C}$						
PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
Start threshold		3.80	3.90	4.00	V	

OVERV COMPARATOR, $0^{\circ}\text{C} \leq T_J \leq 125^{\circ}\text{C}$						
PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
Start threshold		4.35	4.45	4.55	V	
Stop threshold		4.25	4.30	4.35	V	
Hysteresis		50			mV	

TAPERDET COMPARATOR, $0^{\circ}\text{C} \leq T_J \leq 125^{\circ}\text{C}$						
PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
Trip threshold		12	18.5	25	mV	

TMR EN LOGIC INPUT, $0^{\circ}\text{C} \leq T_J \leq 125^{\circ}\text{C}$						
PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
High-level input voltage		2.25			V	
Low-level input voltage				0.8	V	
Input pulldown resistance		100		200	k Ω	

VSEL LOGIC INPUT, $0^{\circ}\text{C} \leq T_J \leq 125^{\circ}\text{C}$						
PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
High-level input voltage		2.25			V	
Low-level input voltage				0.8	V	
Input pulldown resistance		100		200	k Ω	

TMR SEL INPUT $0^{\circ}\text{C} \leq T_J \leq 125^{\circ}\text{C}$						
PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
High-level input voltage		2.7			V	
Low-level input voltage				0.6	V	
Input bias current	$V_I(\text{TMR SEL}) \leq 5\text{V}$			15	μA	

STAT1, STAT2 (bq24008), $0^{\circ}\text{C} \leq T_J \leq 125^{\circ}\text{C}$						
PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
Output (low) saturation voltage	$I_O = 10\text{ mA}$			1.5	V	
Output (low) saturation voltage	$I_O = 4\text{ mA}$			0.6	V	
Output (high) saturation voltage	$I_O = -10\text{ mA}$	$V_{CC}-1.5$			V	
Output (high) saturation voltage	$I_O = -4\text{ mA}$	$V_{CC}-0.5$			V	
Output turn on/off time	$I_O = \pm 10\text{ mA}, C = 100\text{ p}^{(1)}$			100	μs	

(1) Specified by design, not production tested.

POWER-ON RESET (POR), $0^{\circ}\text{C} \leq T_J \leq 125^{\circ}\text{C}$						
PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
POR delay	See Note 1	1.2		3	ms	
POR falling-edge deglitch	See Note 1	25		75	μs	

(1) Specified by design, not production tested.

ELECTRICAL CHARACTERISTICS CONTINUED

over recommended operating junction temperature supply and input voltages, and $V_I (V_{CC}) \geq V_I (IN)$ (unless otherwise noted)

APG/THERM DELAY, $0^\circ\text{C} \leq T_J \leq 125^\circ\text{C}$					
PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
APG/THERM falling-edge deglitch	See Note 1	25		75	μs

(1) Specified by design, not production tested.

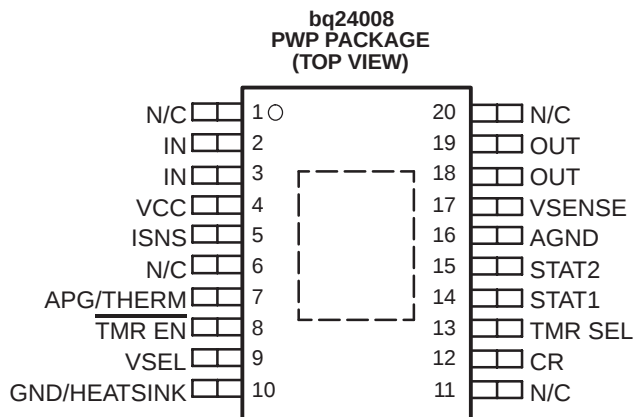
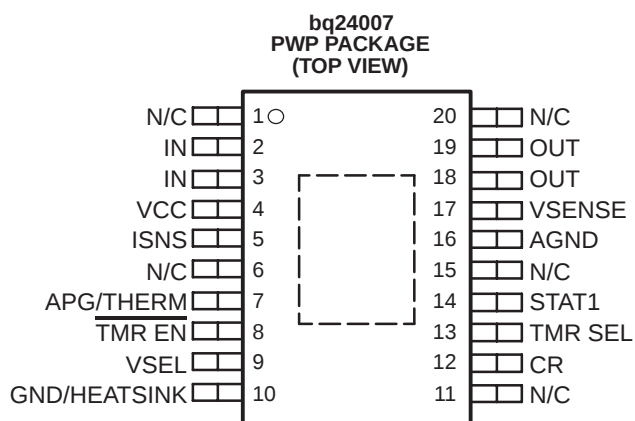
TIMERS, $0^\circ\text{C} \leq T_J \leq 125^\circ\text{C}$					
PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
User-selectable timer accuracy	$T_A = 25^\circ\text{C}$	-15%		15%	
		-20%		20%	
Precharge and taper timer			22.5		minute

THERMAL SHUTDOWN, $0^\circ\text{C} \leq T_J \leq 125^\circ\text{C}$					
PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Thermal trip	See Note 1		165		$^\circ\text{C}$
Thermal hysteresis	See Note 1		10		$^\circ\text{C}$

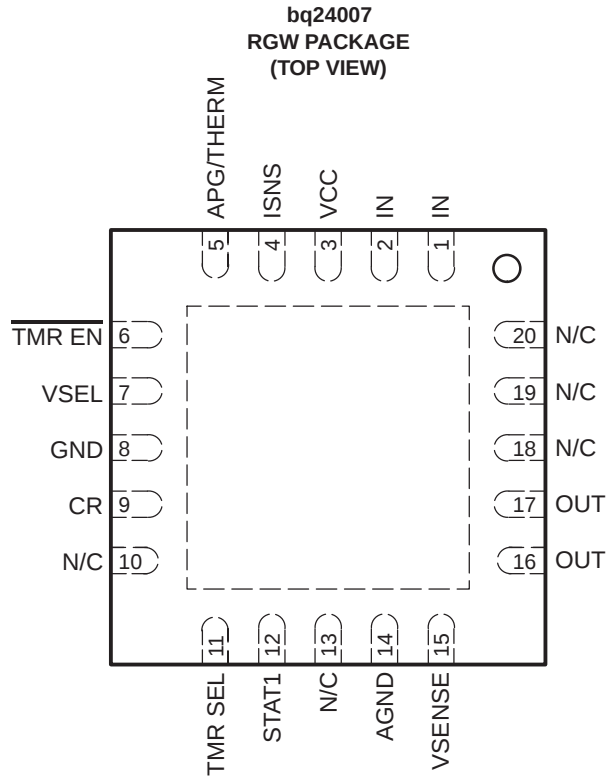
(1) Specified by design, not production tested.

CR PIN, $0^\circ\text{C} \leq T_J \leq 125^\circ\text{C}$					
PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Output voltage	$0 < I_{O(CR)} < 100 \mu\text{A}$	2.81	2.85	2.88	V

PIN ASSIGNMENTS



N/C – Do not connect

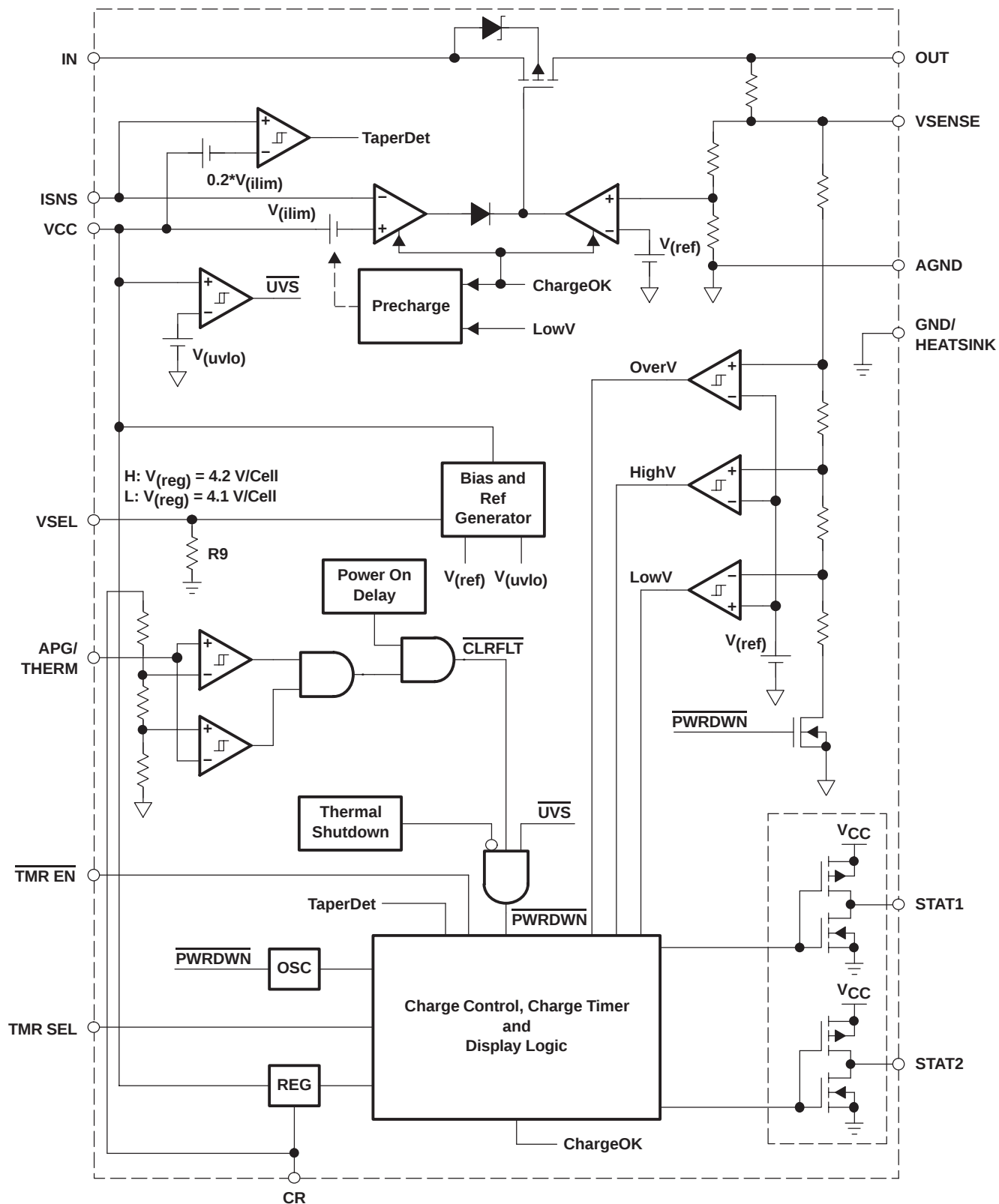


N/C – Do Not Connect

Terminal Functions

TERMINAL			I/O	DESCRIPTION
NAME	NO.			
	PWP	RGW		
AGND	16	14		Ground pin; connect close to the negative battery terminal.
APG/THERM	7	5	I	Adapter power good input/thermistor sense input
CR	12	9	I	Internal regulator bypass capacitor
TMR EN	8	6	I	Charge-enable input. Active-high enable input with internal pulldown. Low-current stand-by mode active when EN is low.
GND/HEATSINK	10	8		Ground pin; connect to PowerPAD heat-sink layout pattern.
IN	2, 3	1, 2	I	Input voltage. This input provides the charging voltage for the battery.
ISNS	5	4	I	Current sense input
N/C	1, 6, 11, 15, 20	10, 13, 18–20		No connect. These pins must be left floating. Pin 15 is N/C on bq24007PWP only. Pin 13 is N/C on bq24007RGW only.
OUT	18, 19	16, 17	O	Charge current output
STAT1	14	12	O	Status display output 1
STAT2	15	–	O	Status display output 2 (for bq24008 only)
TMR SEL	13	11	I	Charge timer selection input
VCC	4	3	I	Supply voltage
VSEL	9	7	I	4.1-V or 4.2-V charge regulation selection input
VSENSE	17	15	I	Battery voltage sense input

FUNCTIONAL BLOCK DIAGRAM



TYPICAL CHARACTERISTICS

OUTPUT VOLTAGE
vs
OUTPUT CURRENT

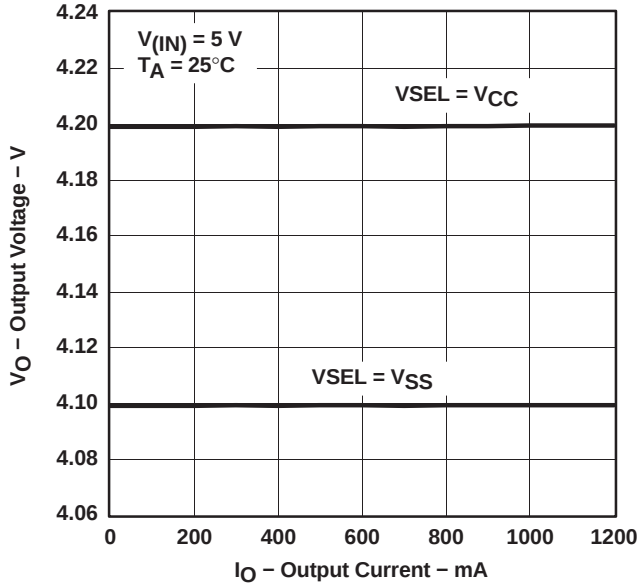


Figure 1

OUTPUT VOLTAGE
vs
JUNCTION TEMPERATURE

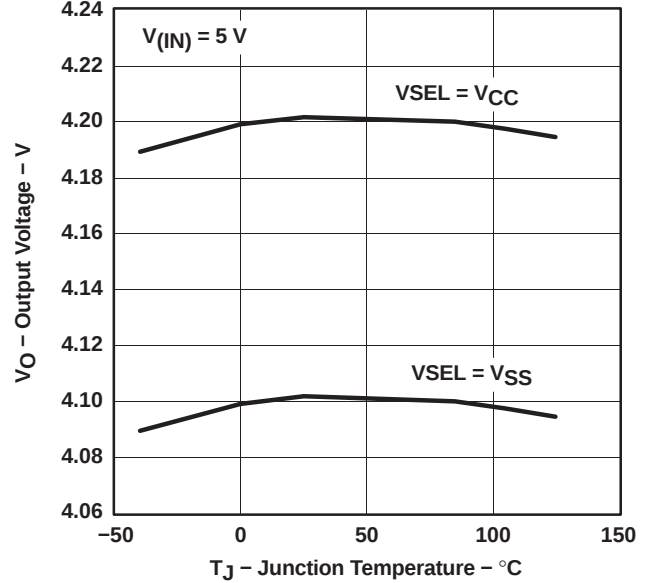


Figure 2

OUTPUT VOLTAGE
vs
INPUT VOLTAGE

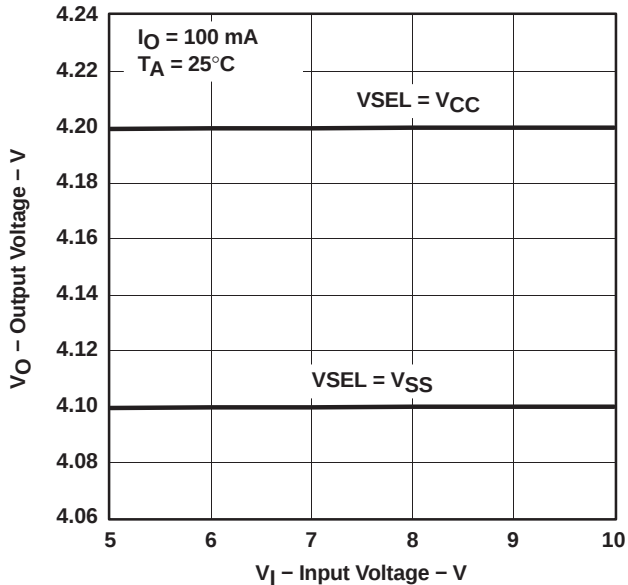


Figure 3

CURRENT SENSE VOLTAGE
vs
INPUT VOLTAGE

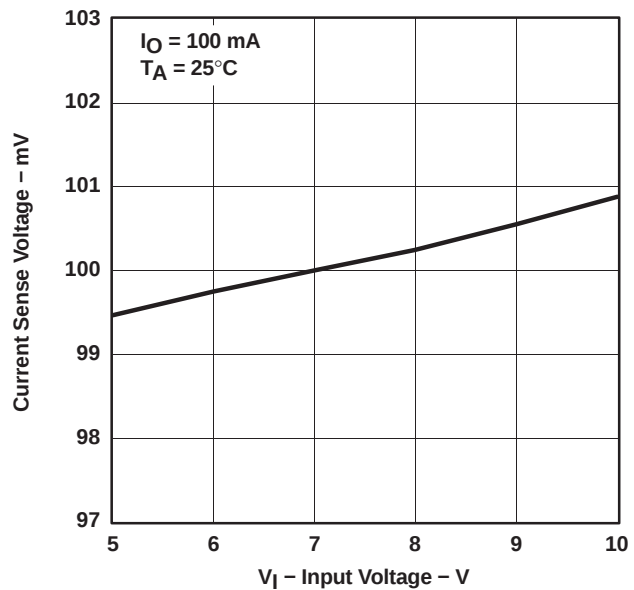


Figure 4

TYPICAL CHARACTERISTICS

CURRENT SENSE VOLTAGE
vs
JUNCTION TEMPERATURE

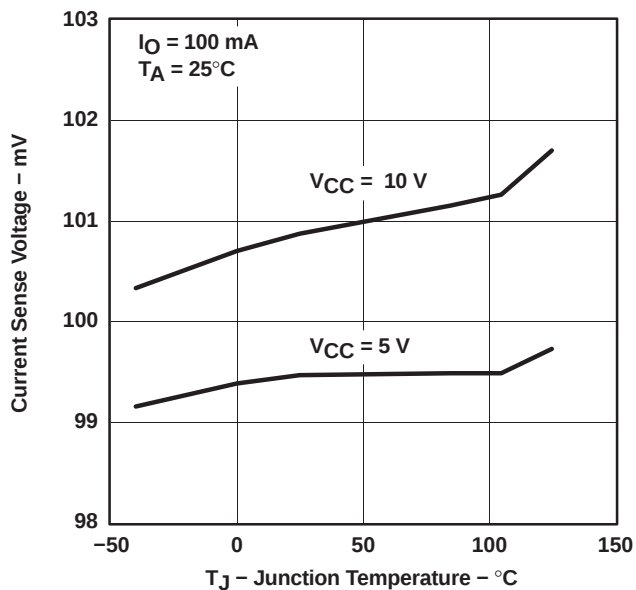


Figure 5

QUIESCENT CURRENT
vs
INPUT VOLTAGE

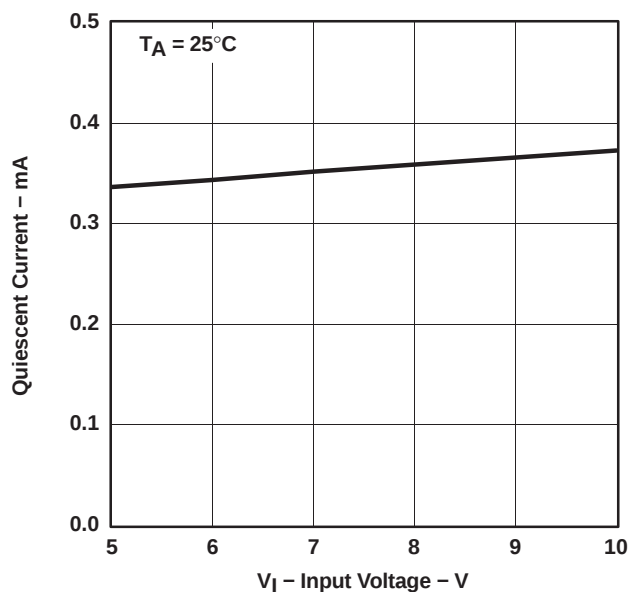


Figure 6

DROPOUT VOLTAGE
vs
INPUT VOLTAGE

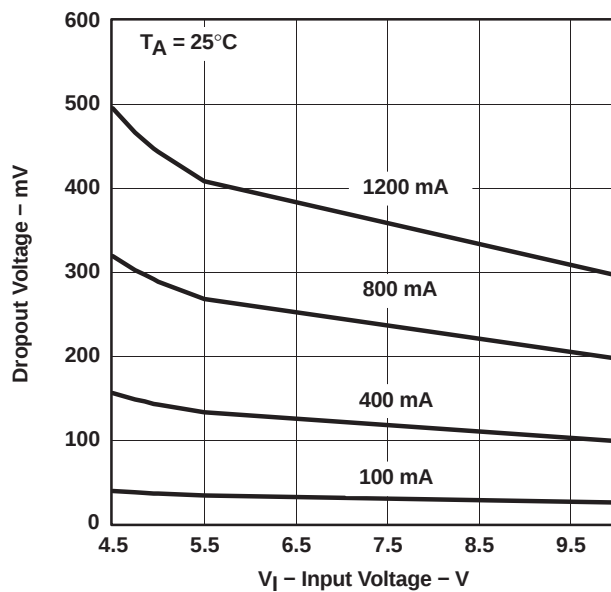


Figure 7

TYPICAL CHARACTERISTICS

DROPOUT VOLTAGE
vs
OUTPUT CURRENT

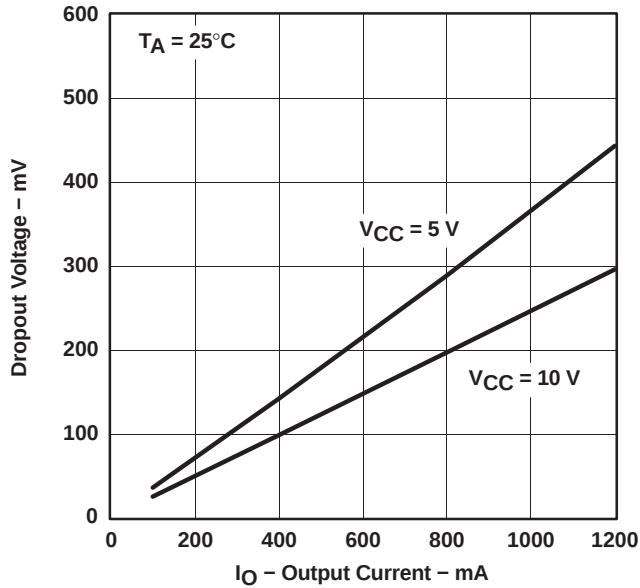


Figure 8

DROPOUT VOLTAGE
vs
JUNCTION TEMPERATURE

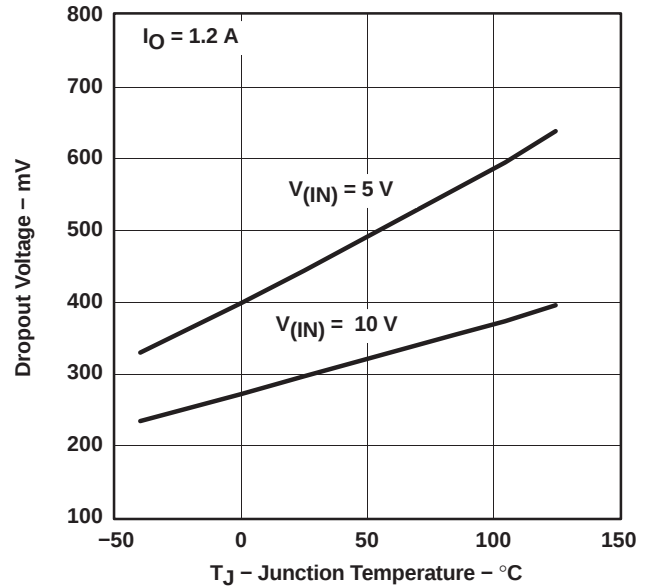


Figure 9

REVERSE CURRENT
vs
JUNCTION TEMPERATURE

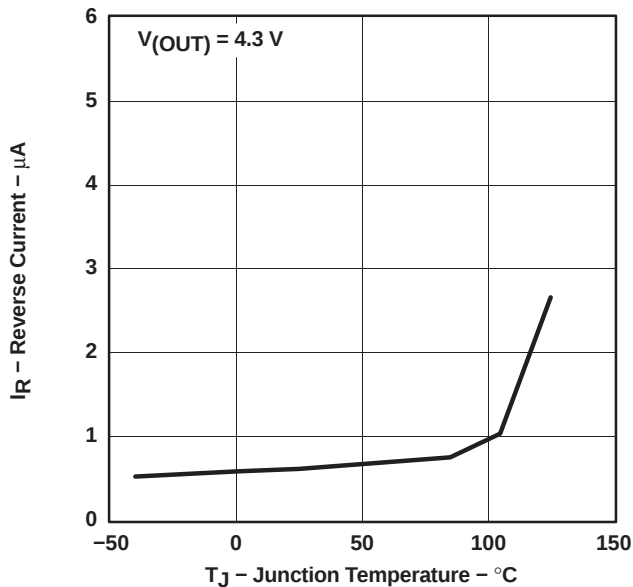


Figure 10

REVERSE CURRENT LEAKAGE
vs
VOLTAGE ON OUT PIN

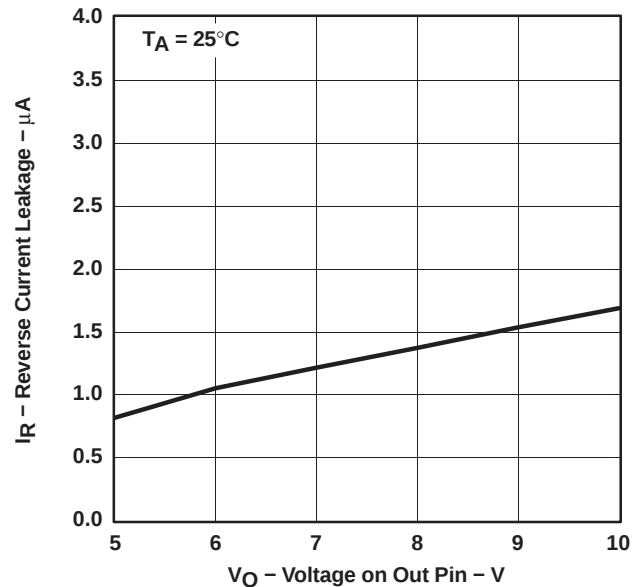


Figure 11

APPLICATION INFORMATION

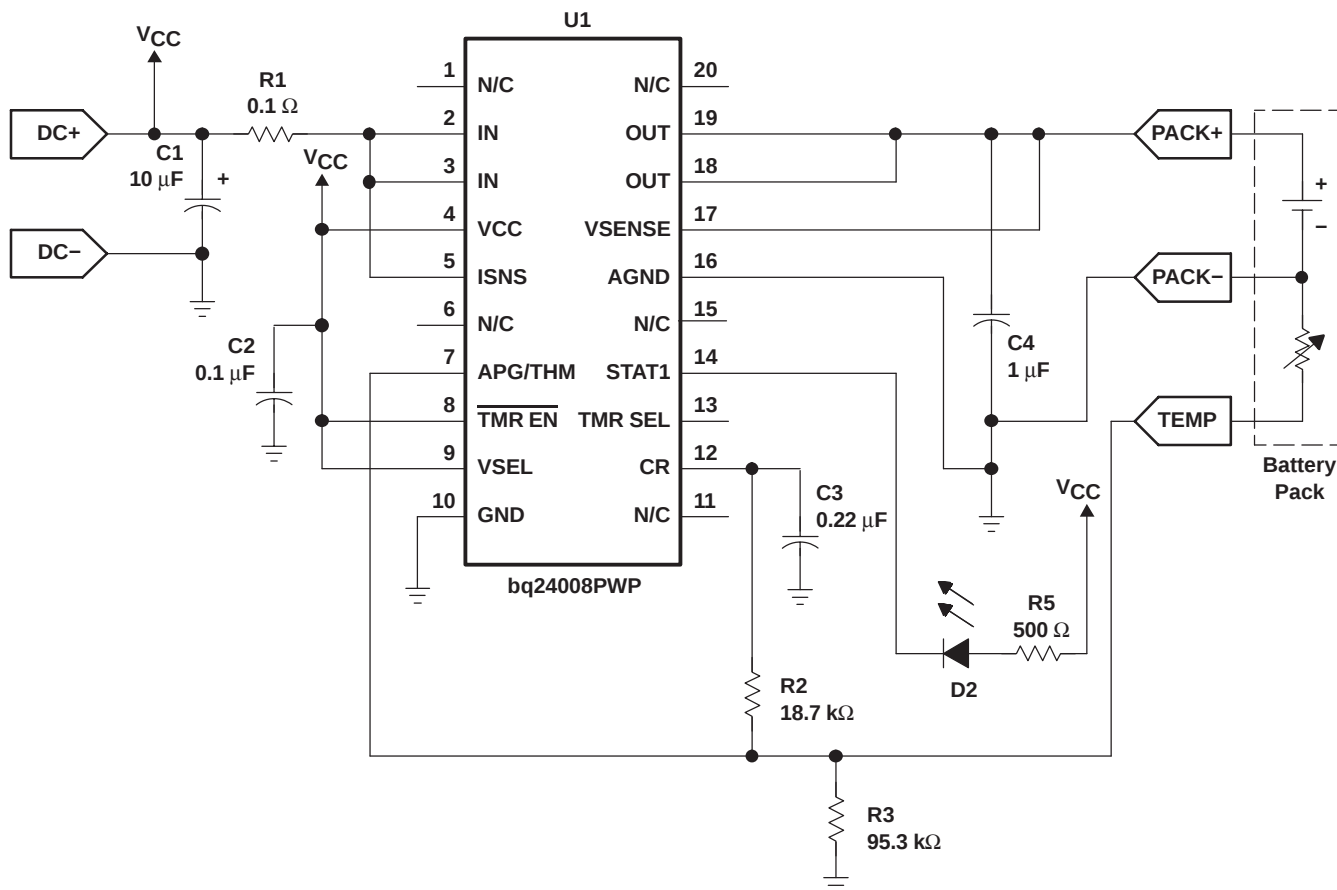


Figure 12. Single-Cell Li-Ion/Li-Pol Charger With Safety Timer Disabled

- If the TMR SEL pin is left floating (3 HR time), a 10-pF capacitor should be installed between TMR SEL and CR.
- If a micro process is monitoring the STAT pins, it may be necessary to add some hysteresis into the feedback to prevent the STAT pins from cycling while crossing the taper detect threshold (usually less than one half second). See SLUU083 EVM or SLUU113 EVM for additional resistors used for the STAT pins.

APPLICATION INFORMATION

FUNCTIONAL DESCRIPTION

The bq2400x supports a precision current- and voltage-regulated Li-Ion charging system suitable for cells with either coke or graphite anodes. See Figure 13 for a typical charge profile and Figures 14 and 15 for an operational flowchart.

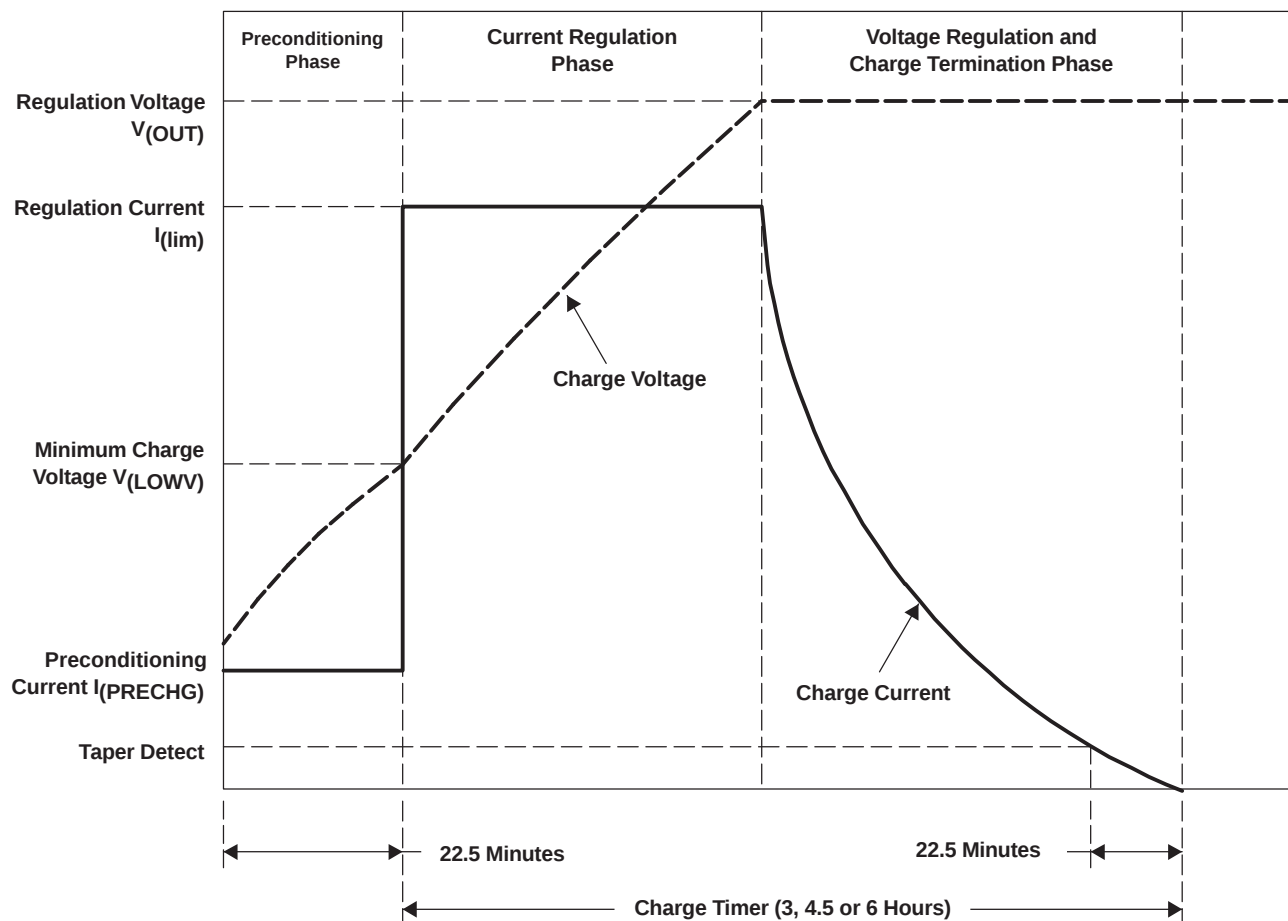


Figure 13. Typical Charge Profile

13

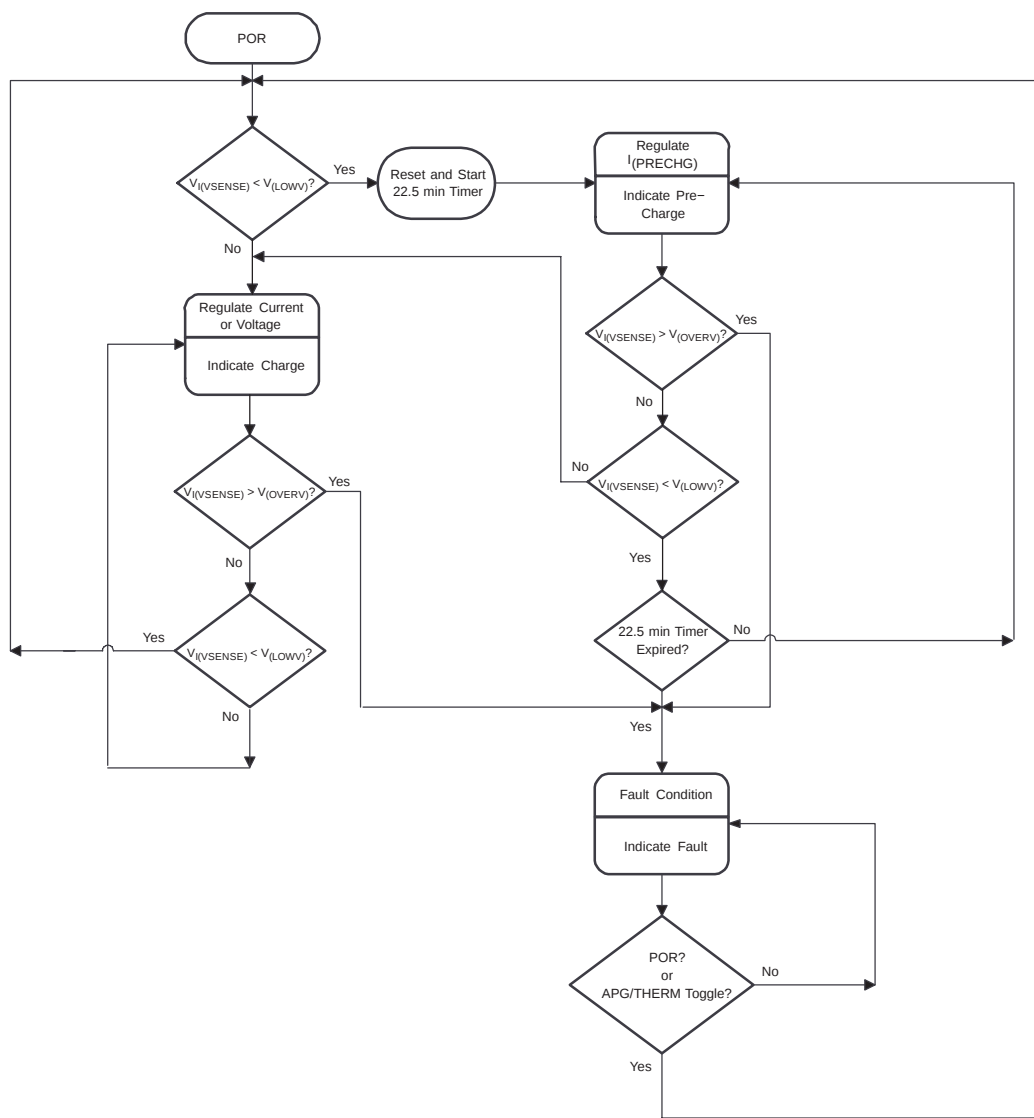


Figure 15. Operational Flow Chart With the Charge Timers Disabled ($\overline{\text{TMR EN}} = 1$)

Charge Qualification and Preconditioning

The bq2400x starts a charge cycle when power is applied while a battery is present. Charge qualification is based on battery voltage and the APG/THERM input.

As shown in the block diagram, the internal LowV comparator output prevents fast-charging a deeply depleted battery. When set, charging current is provided by a dedicated precharge current source. The precharge timer limits the precharge duration. The precharge current also minimizes heat dissipation in the pass element during the initial stage of charge.

The APG/THERM input can also be configured to monitor

either the adapter power or the battery temperature using a thermistor. The bq2400x suspends charge if this input is outside the limits set by the user. Please refer to the APG/THERM input section for additional details.

Timer Enable Function

The logic $\overline{\text{TMR EN}}$ enables or disables the charge safety timer. A low-level signal on this pin enables the timer (charge and taper). A high-level input disables the timers and allows the charger to operate continuously. No charge termination is provided when this input is high. Note that the preconditioning timer remains active regardless of the status of this input.

The $\overline{\text{TMR EN}}$ input can be changed any time during the charge cycle. A low-to-high transition on this pin resets the timers.

APG/THERM Input

The bq400x continuously monitors temperature or system input voltage by measuring the voltage between the APG/THERM (adapter power good/thermistor) and GND. For temperature, a negative- or a positive- temperature

coefficient thermistor (NTC, PTC) and an external voltage divider typically develop this voltage. (See Figure 16.) The bq2400x compares this voltage against its internal V_{TP1} and V_{TP2} thresholds to determine if charging is allowed. (See Figure 17.)

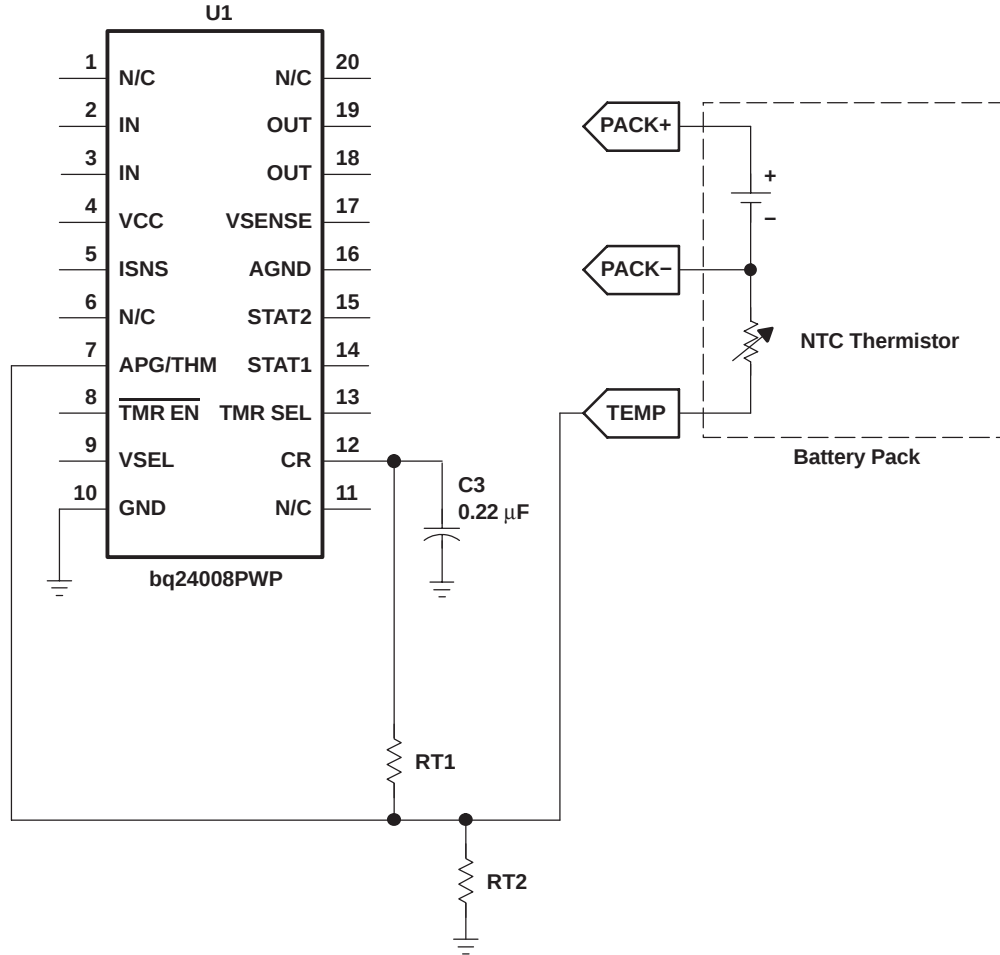


Figure 16. Temperature Sensing Circuit

If the charger designs incorporate a thermistor, the resistor divider RT1 and RT2 is calculated by using the following two equations.

First, calculate RT2.

$$RT2 = \frac{V_B R_H R_C \left[\frac{1}{V_C} - \frac{1}{V_H} \right]}{R_H \left(\frac{V_B}{V_H} - 1 \right) - R_C \left(\frac{V_B}{V_C} - 1 \right)}$$

then use the resistor value to find RT1.

$$RT1 = \frac{\frac{V_B}{V_C} - 1}{\frac{1}{RT2} + \frac{1}{R_C}}$$

Where:

$V_B = V_{CR}$ (bias voltage)

R_H = Resistance of the thermistor at the desired hot trip threshold

R_C = Resistance of the thermistor at the desired cold trip threshold

V_H = VP2 or the lower APG trip threshold

V_C = VP2 or the upper APG trip threshold

RT1 = Top resistor in the divider string

RT2 = Bottom resistor in the divider string

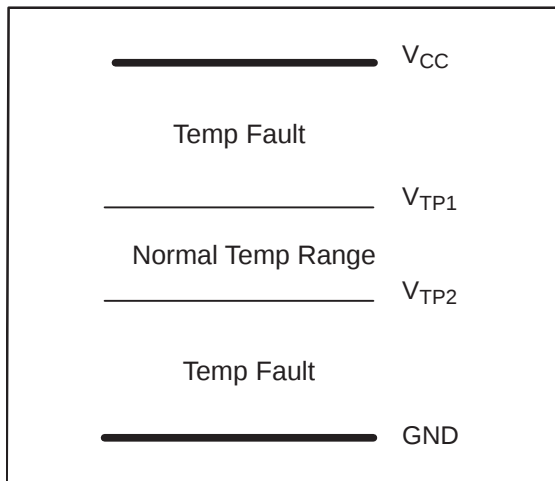


Figure 17. Temperature Threshold

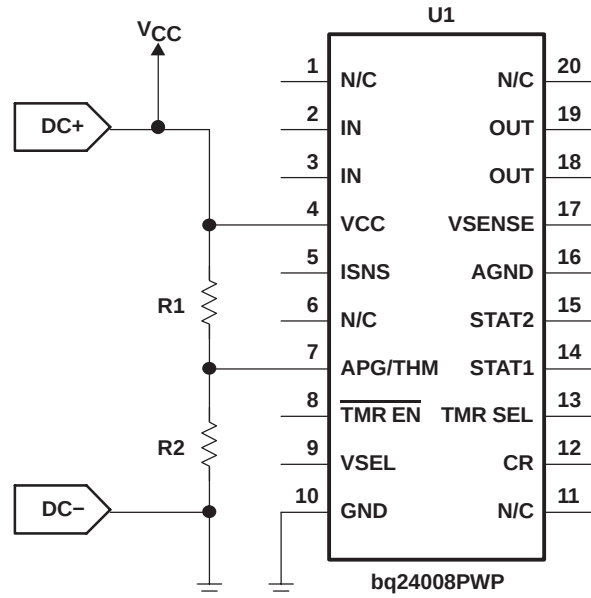


Figure 18. APG Sensing Circuit

Values of resistors R1 and R2 can be calculated using the following equation:

$$V_{(APG)} = V_{CC} \frac{R2}{(R1 + R2)}$$

where $V_{(APG)}$ is the voltage at the APG/THM pin.

Current Regulation

The bq2400x provides current regulation while the battery-pack voltage is less than the regulation voltage. The current regulation loop effectively amplifies the error between a reference signal, V_{lim} , and the drop across the external sense resistor, R_{SNS} .

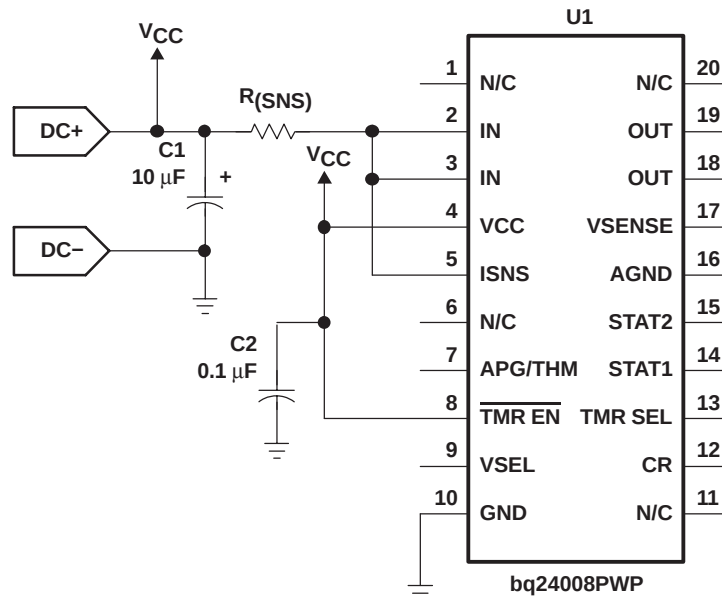


Figure 19. Current Sensing Circuit

Charge current feedback, applied through pin ISNS, maintains regulation around a threshold of V_{lim} . The following formula calculates the value of the sense resistor:

$$R_{(SNS)} = \frac{V_{(lim)}}{I_{(REG)}}$$

where I_{REG} is the desired charging current.

Voltage Monitoring and Regulation

Voltage regulation feedback is through pin VSENSE. This input is tied directly to the positive side of the battery pack. The bq2400x supports cells with either coke (4.1 V) or graphite (4.2 V) anode. Pin VSEL selects the charge regulation voltage.

VSEL State (see Note)	CHARGE REGULATION VOLTAGE
Low	4.1 V
High	4.2 V

NOTE: VSEL should not be left floating.

Charge Termination

The bq2400x continues with the charge cycle until termination by one of the two possible termination conditions:

Maximum Charge Time: The bq2400x sets the maximum charge time through pin TMRSEL. The TMR SEL pin allows the user to select between three different total charge-time timers (3, 4, 5, or 6 hours). The charge timer is initiated after the preconditioning phase of the charge and is reset at the beginning of a new charge cycle. Note that in the case of a fault condition, such as an out-of-range signal on the APG/THERM input or a thermal shutdown, the bq2400x suspends the timer. Charge timers are active only when $\overline{TMR\ EN}$ input is low.

TMRSEL STATE	CHARGE TIME
Floating ⁽¹⁾	3 hours
Low	6 hours
High	4.5 hours

(1) To improve noise immunity, it is recommended that a minimum of 10 pF capacitor be tied to Vss on a floating pin.

Minimum Current: The bq2400x monitors the charging current during the voltage regulation phase. The bq2400x initiates a 22-minute timer once the current falls below the taperdet trip threshold. Fast charge is terminated once the 22-minute timer expires.

Charge Status Display

The two available options allow the user to configure the charge status display for single LED (bq24007) or a bicolor LED (bq24008). The output stage is totem pole for the bq24007 and bq24008. The following tables summarize the operation of the three options:

Table 1. bq24007 (Single LED)

CHARGE STATE	STAT1
Precharge	ON (LOW)
Fast charge	ON (LOW)
FAULT	Flashing (1 Hz, 50% duty cycle)
Done (>90%)	OFF (HIGH)
Sleep-mode	OFF (HIGH)
APG/Therm invalid	OFF (HIGH)
Thermal shutdown	OFF (HIGH)
Battery absent	OFF (HIGH)

Table 2. bq24008 (Single Bicolor LED)

CHARGE STATE	LED1 (RED)	LED2 (GREEN)	APPARENT COLOR
Precharge	ON (LOW)	OFF (HIGH)	RED
Fast charge	ON (LOW)	OFF (HIGH)	RED
FAULT	ON (LOW)	ON (LOW)	YELLOW
Done (>90%)	OFF (HIGH)	ON (LOW)	GREEN
Sleep-mode	OFF (HIGH)	OFF (HIGH)	OFF
APG/Therm invalid	OFF (HIGH)	OFF (HIGH)	OFF
Thermal shutdown	OFF (HIGH)	OFF (HIGH)	OFF
Battery absent	OFF (HIGH)	OFF (HIGH) ⁽¹⁾	OFF ⁽¹⁾

(1) If thermistor is used, then the green LED is off.

Thermal Shutdown

The bq2400x monitors the junction temperature T_J of the DIE and suspends charging if T_J exceeds 165°C. Charging resumes when T_J falls below 155°C.

DETAILED DESCRIPTION

POWER FET

The integrated transistor is a P-channel MOSFET. The power FET features a reverse-blocking Schottky diode, which prevents current flow from OUT to IN.

An internal thermal-sense circuit shuts off the power FET when the junction temperature rises to approximately 165°C. Hysteresis is built into the thermal sense circuit. After the device has cooled approximately 10°C, the power FET turns back on. The power FET continues to cycle off and on until the fault is removed.

CURRENT SENSE

The bq2400x regulates current by sensing, on the ISNS pin, the voltage drop developed across an external sense resistor. The sense resistor must be placed between the supply voltage (Vcc) and the input of the IC (IN pins).

VOLTAGE SENSE

To achieve maximum voltage regulation accuracy, the bq2400x uses the feedback on the VSENSE pin. Externally, this pin should be connected as close to the battery cell terminals as possible. For additional safety, a 10-k Ω internal pullup resistor is connected between the VSENSE and OUT pins.

TIMER ENABLE

The logic $\overline{\text{TMR EN}}$ enables or disables the charge safety timer. A low-level signal on this pin enables the timers. A high-level input disables the timers and allows the charge to operate continuously. No charge termination is provided when this input is high. Note that the preconditioning timer remains active regardless of the status of this input.

THERMAL INFORMATION

THERMALLY ENHANCED TSSOP-20

The thermally enhanced PWP package is based on the 20-pin TSSOP, but includes a thermal pad (see Figure 20) to provide an effective thermal contact between the IC and the PWB.

Traditionally, surface mount and power have been mutually exclusive terms. A variety of scaled-down TO220-type packages have leads formed as gull wings to make them applicable for surface-mount applications. These packages, however, suffer from several shortcomings: they do not address the very low profile requirements (<2 mm) of many of today's advanced systems, and they do not offer a pin-count high enough to accommodate increasing integration. On the other hand, traditional low-power surface-mount packages require power-dissipation derating that severely limits the usable range of many high-performance analog circuits.

The PWP package (thermally enhanced TSSOP) combines fine-pitch surface-mount technology with thermal performance comparable to much larger power packages.

The PWP package is designed to optimize the heat transfer to the PWB. Because of the very small size and limited mass of a TSSOP package, thermal enhancement is achieved by improving the thermal conduction paths that remove heat from the component. The thermal pad is formed using a lead-frame design (patent pending) and manufacturing technique to provide the user with direct connection to the heat-generating IC. When this pad is soldered or otherwise coupled to an external heat dissipator, high power dissipation in the ultrathin, fine-pitch, surface-mount package can be reliably achieved.

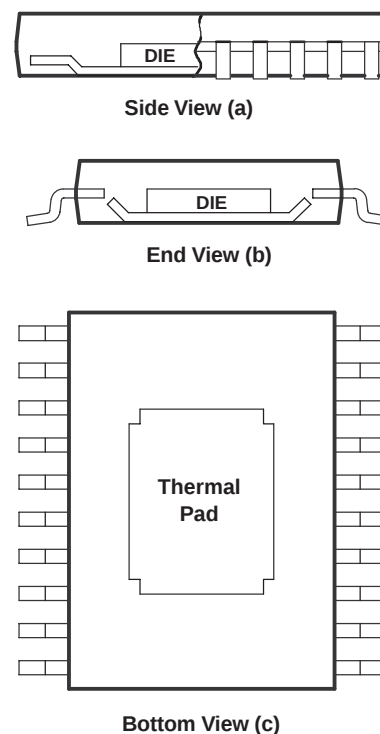


Figure 20. Views of Thermally Enhanced PWP Package

Because the conduction path has been enhanced, power-dissipation capability is determined by the thermal considerations in the PWB design. For example, simply adding a localized copper plane (heat-sink surface), which is coupled to the thermal pad, enables the PWP package to dissipate 2.5 W in free air. (Reference Figure 22(a), 8 cm² of copper heat sink and natural convection.) Increasing the heat-sink size increases the power dissipation range for the component. The power dissipation limit can be further improved by adding airflow to a PWB/IC assembly. (See Figure 22(b) and 22(c).) The line drawn at 0.3 cm² in Figures 21 and 22 indicates performance at the minimum recommended heat-sink size.

THERMAL INFORMATION

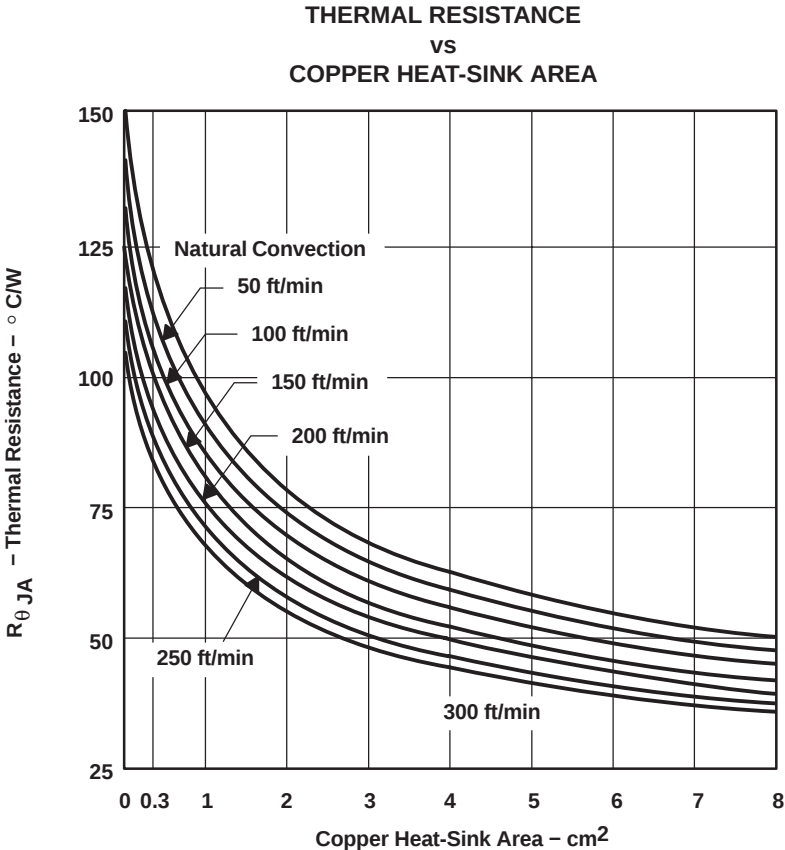
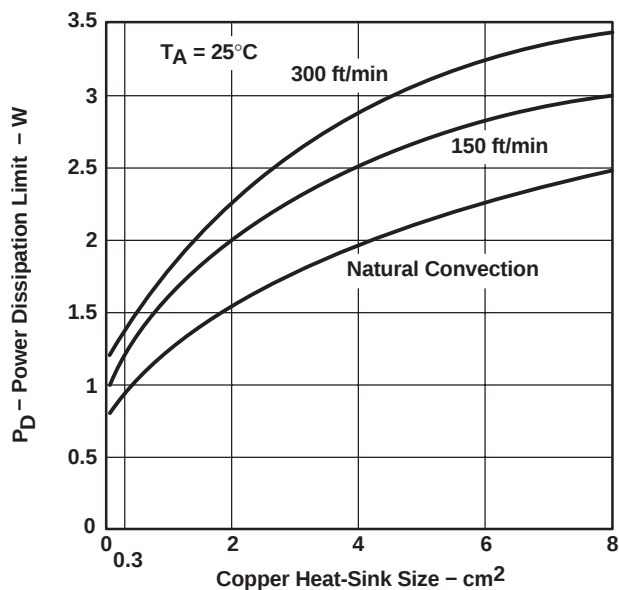
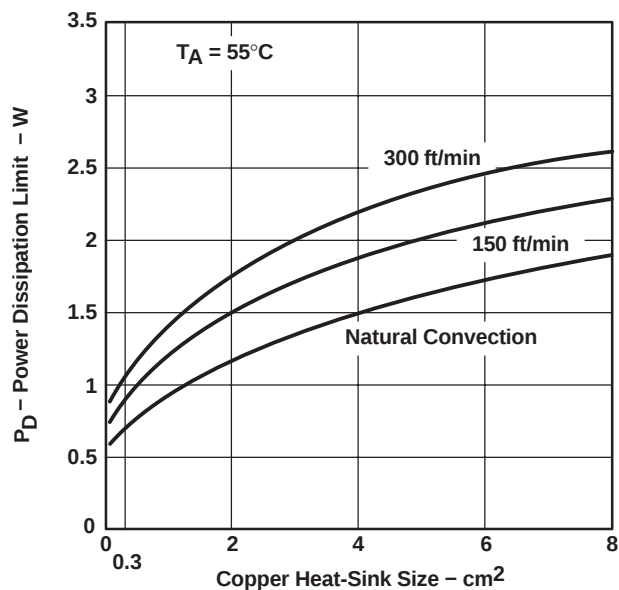


Figure 21

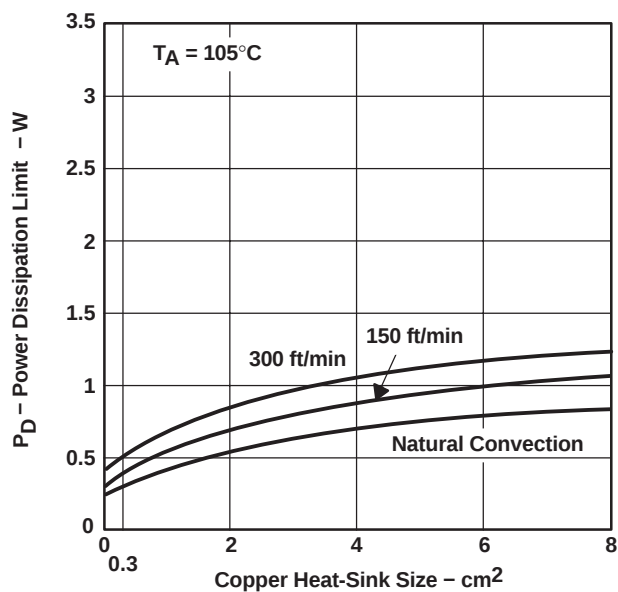
THERMAL INFORMATION



(a)



(b)



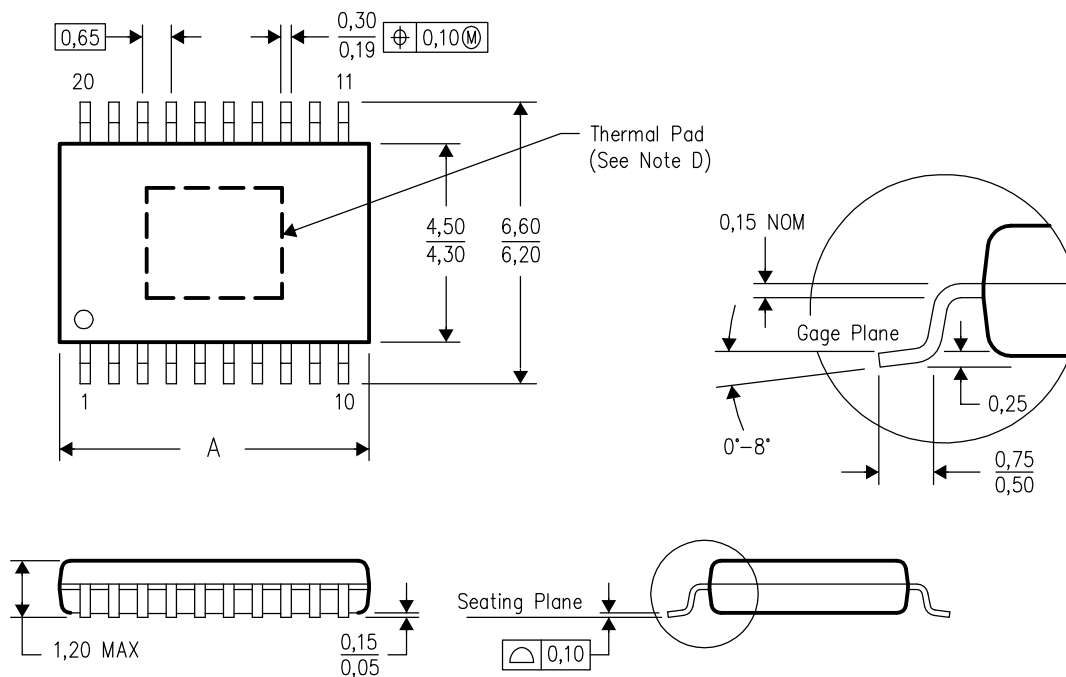
(c)

Figure 22. Power Ratings of the PWP Package at Ambient Temperatures of 25°C , 55°C , and 105°C

PWP (R-PDSO-G**)

PowerPAD™ PLASTIC SMALL-OUTLINE PACKAGE

20 PIN SHOWN



PINS **	14	16	20	24	28
DIM					
A MAX	5,10	5,10	6,60	7,90	9,80
A MIN	4,90	4,90	6,40	7,70	9,60

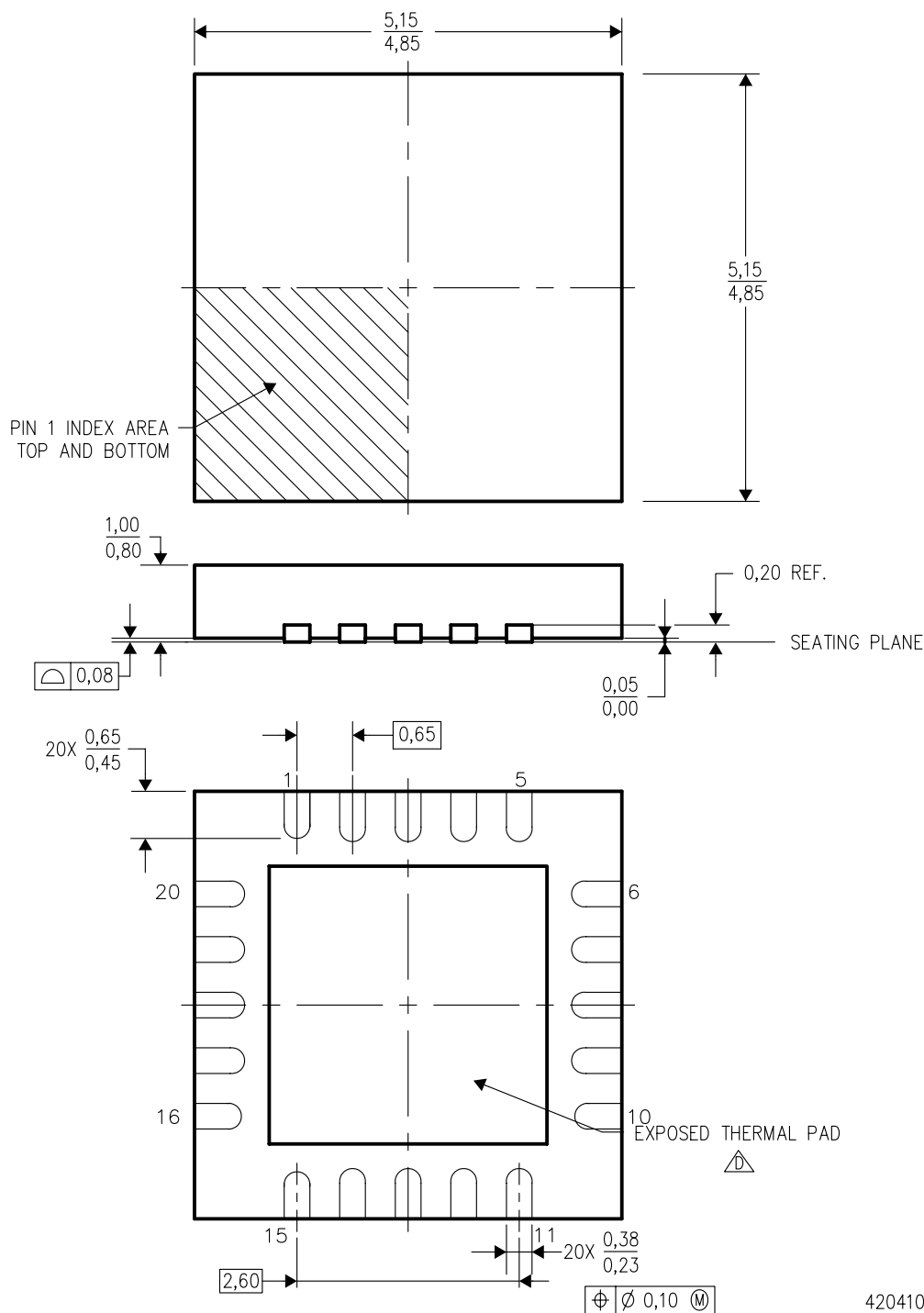
4073225/G 08/03

- NOTES:
- All linear dimensions are in millimeters.
 - This drawing is subject to change without notice.
 - Body dimensions do not include mold flash or protrusions.
 - This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 for information regarding recommended board layout. This document is available at www.ti.com <<http://www.ti.com>>.
 - Falls within JEDEC MO-153

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RGW (S-PQFP-N20)

PLASTIC QUAD FLATPACK



4204100/B 08/04

- NOTES:
- A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5-1994.
 - B. This drawing is subject to change without notice.
 - C. Quad Flat pack, No-leads (QFN) package configuration.
 - D. The package thermal pad must be soldered to the board for thermal and mechanical performance.. See the Product Data Sheet for details regarding the exposed thermal pad dimensions.
 - E. Falls within JEDEC MO-220.

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