

Double channel high-side driver with analog current sense for 24 V automotive applications



PowerSSO-24

Features

Description	Parameter	Value
Max. transient supply voltage	V_{CC}	58 V
Operating voltage range	V_{CC}	8 to 36 V
Typ. on-state resistance (per channel)	R_{ON}	35 m Ω
Current limitation (typ.)	I_{LIM}	42 A
Off-state supply current	I_S	2 μ A



- AEC-Q100 qualified
- General
 - Very low standby current
 - 3.0 V CMOS compatible input
 - Optimized electromagnetic emission
 - Very low electromagnetic susceptibility
 - Compliant with European directive 2002/95/EC
 - Fault reset standby pin (FR_Stby)
 - Optimized for LED application
- Diagnostic functions
 - Proportional load current sense
 - High current sense precision for wide range currents
 - Off-state openload detection
 - Output short to V_{CC} detection
 - Overload and short to ground latch-off
 - Thermal shutdown latch-off
 - Very low current sense leakage
- Protections
 - Undervoltage shutdown
 - Overvoltage clamp
 - Load current limitation
 - Self limiting of fast thermal transients
 - Protection against loss of ground and loss of V_{CC}
 - Thermal shutdown
 - Electrostatic discharge protection

Product status link	
VND5T035LAK-E	
Product summary	
Order code	VND5T035LAK-E
Package	PowerSSO-24
Packing	Tube
Order code	VND5T035LAKTR-E
Package	PowerSSO-24
Packing	Tape and reel

Applications

- All types of resistive, inductive and capacitive loads

Description

The VND5T035LAK-E is a monolithic device made using STMicroelectronics VIPower technology, intended for driving resistive or inductive loads with one side connected to the ground. Active V_{CC} pin voltage clamp protects the device against low energy spikes.

The device integrates an analog current sense, which delivers a current proportional to the load current.

Fault conditions such as overload, overtemperature, or short to V_{CC} are reported via the current sense pin.

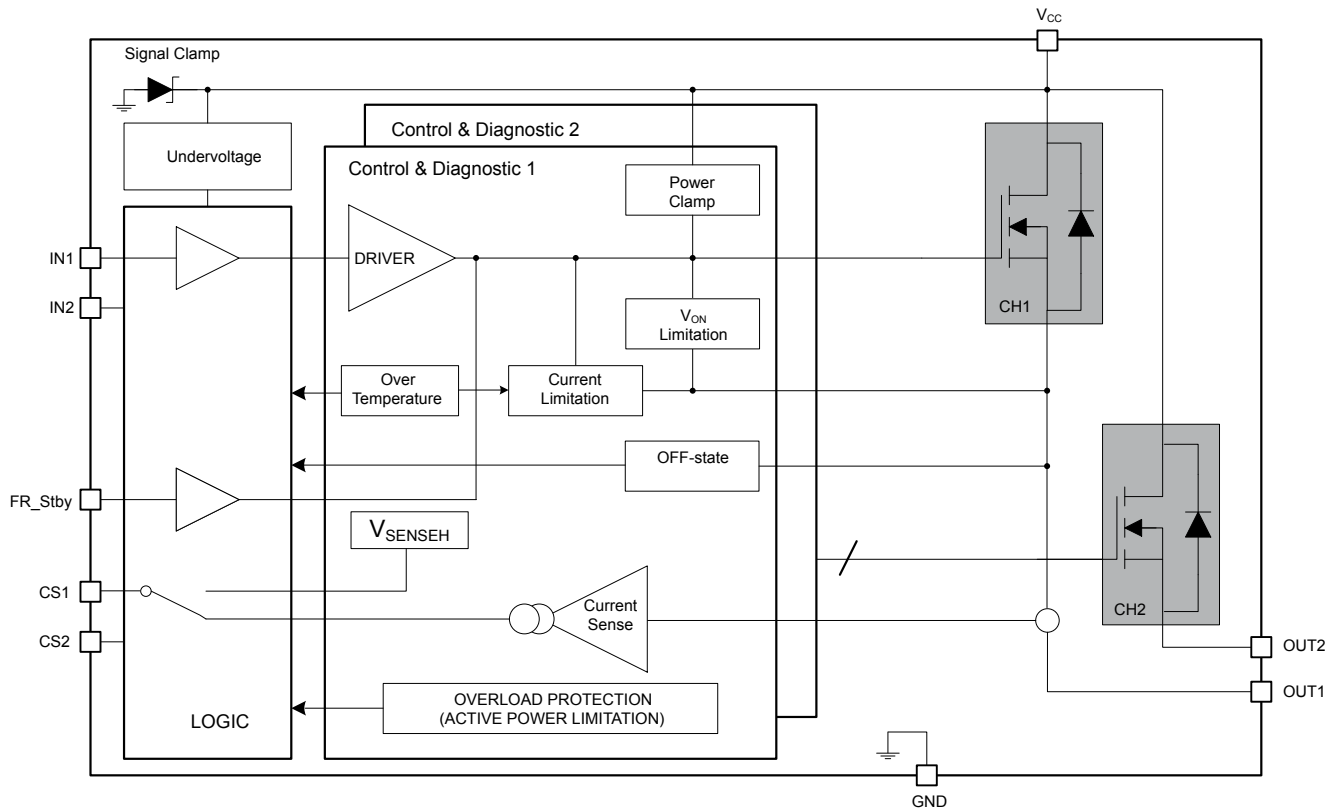
Output current limitation protects the device in overload conditions. The device latches off in case of overload or thermal shutdown.

The device is reset by a low level pass on the fault reset standby pin.

A permanent low level on the inputs and on the fault reset standby pins disables all outputs and sets the device in standby mode.

1 Block diagram and pin description

Figure 1. Block diagram

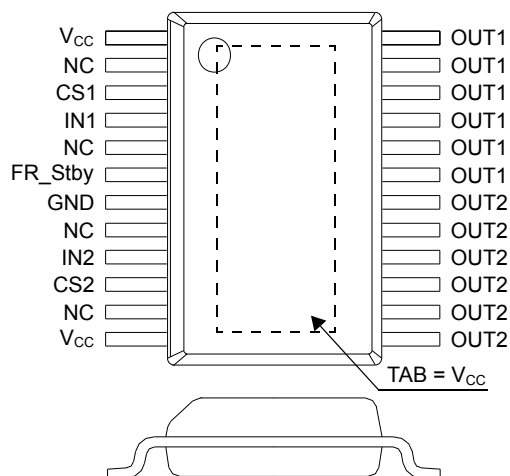


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Table 1. Pin function

Name	Function
V _{CC}	Battery connection.
OUT1, 2	Power outputs.
GND	Ground connection.
IN1, 2	Voltage controlled input pins with hysteresis, CMOS compatible. They control output switch state.
CS1, 2	Analog current sense pins, they deliver a current proportional to the load current.
FR_Stby	In case of latch-off for overtemperature/overcurrent condition, a low pulse on the FR_Stby pin is needed to reset the channel. The device enters in standby mode if all inputs and the FR_Stby pin are low.

Figure 2. Configuration diagram PowerSSO-24 (top view)



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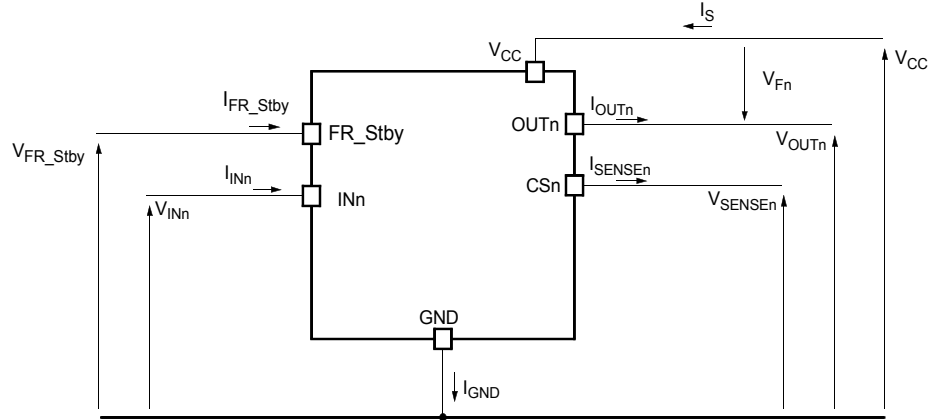
Table 2. Suggested connections for unused and not connected pins

Connection/pin	CurrentSense	NC	Output	Input	FR_Stby
Floating	Not allowed	X ⁽¹⁾	X	X	X
To ground	Through 10 kΩ resistor	X	Not allowed	Through 10 kΩ resistor	Through 10 kΩ resistor

1. X: do not care.

2 Electrical specification

Figure 3. Current and voltage conventions



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2.1 Absolute maximum ratings

Stressing the device above the ratings listed in the [Table 3](#) may cause permanent damage to the device. These are stress ratings only and operation of the device at these or any other conditions above those indicated in the operating sections of this specification is not implied. Exposure to the conditions reported in this section for extended periods may affect device reliability.

Table 3. Absolute maximum ratings

Symbol	Parameter	Value	Unit
V_{CC}	DC supply voltage	58	V
$-V_{CC}$	Reverse DC supply voltage	0.3	V
$-I_{GND}$	DC reverse ground pin current	200	mA
I_{OUT}	DC output current	Internally limited	A
$-I_{OUT}$	Reverse DC output current	40	A
I_{IN}	DC input current	-1 to 10	mA
I_{FR_Stby}	Fault reset standby DC input current	-1 to 1.5	mA
$-I_{CSENSE}$	DC reverse CS pin current	200	mA
V_{CSENSE}	Current sense maximum voltage	$(V_{CC} - 58)$ to V_{CC}	V
E_{MAX}	Maximum switching energy ($L = 2.3$ mH; $V_{BAT} = 32$ V; $T_{Jstart} = 150$ °C; $I_{OUT} = I_{limL}$ (typ.))	250	mJ
L_{smax}	Maximum stray inductance in short circuit condition $R_L = 300$ mΩ, $V_{BAT} = 32$ V, $T_{Jstart} = 150$ °C, $I_{OUT} = I_{limH}$ (max.)	40	μH
V_{ESD}	Electrostatic discharge (human body model: $R = 1.5$ kΩ, $C = 100$ pF)	IN1, 2	4000
		CS1, 2	2000
		FR_Stby	4000
		OUT1, 2	5000
		V_{CC}	5000
V_{ESD}	Charge device model (CDM-AEC-Q100-011)	750	V

Symbol	Parameter	Value	Unit
T_J	Junction operating temperature	-40 to 150	°C
T_{stg}	Storage temperature	-55 to 150	°C

2.2 Thermal data

Table 4. Thermal data

Symbol	Parameter	Value	Unit
R_{thJC}	Thermal resistance, junction-to-case (with one channel ON)	2	°C/W
R_{thJA}	Thermal resistance, junction-to-ambient	See Figure 27	°C/W

2.3 Electrical characteristics

8 V < V_{CC} < 36 V, -40 °C < T_J < 150 °C, unless otherwise specified.

Table 5. Power section

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_{CC}	Operating supply voltage		8	24	36	V
V_{USD}	Undervoltage shutdown			3.5	5	V
$V_{USDhyst}$	Undervoltage shutdown hysteresis			0.5		V
R_{ON}	On-state resistance ⁽¹⁾	$I_{OUT} = 3\text{ A}$, $T_J = 25\text{ °C}$		35		mΩ
		$I_{OUT} = 3\text{ A}$, $T_J = 150\text{ °C}$			70	
V_{clamp}	Clamp voltage	$I_S = 20\text{ mA}$	58	64	70	V
I_S	Supply current	Off-state, $V_{CC} = 24\text{ V}$, $T_J = 25\text{ °C}$, $V_{IN} = V_{OUT} = V_{SENSE} = 0\text{ V}$, $V_{FR_Stby} = 0\text{ V}$		2 ⁽²⁾	5 ⁽²⁾	μA
		On-state, $V_{CC} = 24\text{ V}$, $V_{IN} = 5\text{ V}$, $I_{OUT} = 0\text{ A}$		4.2	6	mA
$I_{L(off)}$	Off-state output current	$V_{IN} = V_{OUT} = 0\text{ V}$, $V_{CC} = 24\text{ V}$, $T_J = 25\text{ °C}$	0	0.01	3	μA
		$V_{IN} = V_{OUT} = 0\text{ V}$, $V_{CC} = 24\text{ V}$, $T_J = 125\text{ °C}$	0		5	
V_F	Output - V_{CC} diode voltage	$-I_{OUT} = 3\text{ A}$, $T_J = 150\text{ °C}$			0.7	V

1. For each channel.

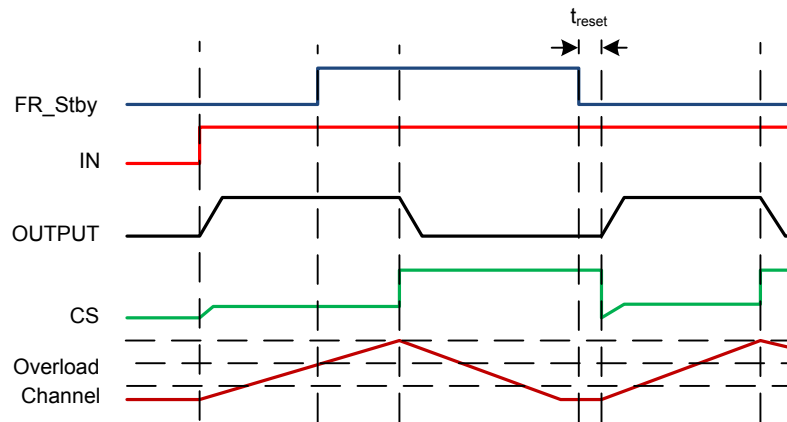
2. Power MOSFET leakage included.

Table 6. Switching ($V_{CC} = 24\text{ V}$, $T_J = 25\text{ °C}$)

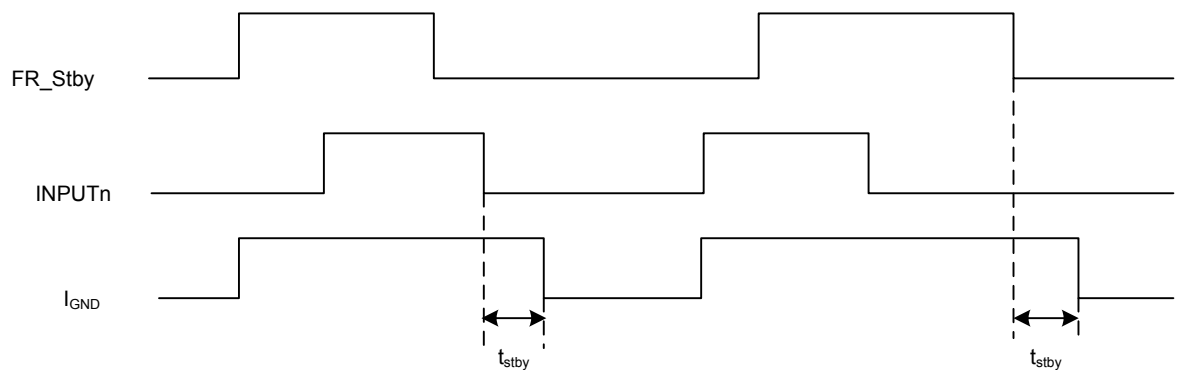
Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$t_{d(on)}$	Turn-on delay time	$R_L = 8\ \Omega$		46		μs
$t_{d(off)}$	Turn-off delay time	$R_L = 8\ \Omega$		54		μs
$(dV_{OUT}/dt)_{(on)}$	Turn-on voltage slope	$R_L = 8\ \Omega$		0.55		$\text{V}/\mu\text{s}$
$(dV_{OUT}/dt)_{(off)}$	Turn-off voltage slope	$R_L = 8\ \Omega$		0.46		$\text{V}/\mu\text{s}$
W_{ON}	Switching energy losses during t_{won}	$R_L = 8\ \Omega$		1		mJ
W_{OFF}	Switching energy losses during t_{woff}	$R_L = 8\ \Omega$		0.65		mJ

Table 7. Logic inputs

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_{IL}	Input low level voltage				0.9	V
I_{IL}	Low level input current	$V_{IN} = 0.9\text{ V}$	1			μA
V_{IH}	Input high level voltage		2.1			V
I_{IH}	High level input current	$V_{IN} = 2.1\text{ V}$			10	μA
$V_{I(hyst)}$	Input hysteresis voltage		0.25			V
V_{ICL}	Input clamp voltage	$I_{IN} = 1\text{ mA}$	5.5		7	V
		$I_{IN} = -1\text{ mA}$		-0.7		
$V_{FR_Stby_L}$	Fault_reset_standby low level voltage				0.9	V
$I_{FR_Stby_L}$	Low level fault_reset_standby current	$V_{FR_Stby} = 0.9\text{ V}$	1			μA
$V_{FR_Stby_H}$	Fault_reset_standby high level voltage		2.1			V
$I_{FR_Stby_H}$	High level fault_reset_standby current	$V_{FR_Stby} = 2.1\text{ V}$			10	μA
$V_{FR_Stby(hyst)}$	Fault_reset_standby hysteresis voltage		0.25			V
$V_{FR_Stby_CL}$	Fault_reset_standby clamp voltage	$I_{FR_Stby} = 15\text{ mA}$ ($t < 10\text{ ms}$)	11		15	V
		$I_{FR_Stby} = -1\text{ mA}$		-0.7		V
t_{reset}	Overload latch-off reset time	See Figure 4	2		24	μs
t_{stby}	Standby delay	See Figure 5	120		1200	μs

Figure 4. t_{reset} definition


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Figure 5. t_{stby} definition


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Table 8. Protections and diagnostics

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{limH}	DC short circuit current	$V_{\text{CC}} = 24 \text{ V}$	30	42	55	A
		$5 \text{ V} < V_{\text{CC}} < 36 \text{ V}$			55	
I_{limL}	Short circuit current during thermal cycling	$V_{\text{CC}} = 24 \text{ V}, T_{\text{R}} < T_{\text{J}} < T_{\text{TSD}}$		10.5		A
T_{TSD}	Shutdown temperature		150	175	200	°C
T_{R}	Reset temperature		$T_{\text{RS}} + 1$	$T_{\text{RS}} + 5$		°C
T_{RS}	Thermal reset of status		135			°C
T_{HYST}	Thermal hysteresis ($T_{\text{TSD}} - T_{\text{R}}$)			7		°C
V_{DEMG}	Turn-off output voltage clamp	$I_{\text{OUT}} = 3 \text{ A}, V_{\text{IN}} = 0 \text{ V}, L = 6 \text{ mH}$	$V_{\text{CC}} - 58$	$V_{\text{CC}} - 64$	$V_{\text{CC}} - 70$	V
V_{ON}	Output voltage drop limitation	$I_{\text{OUT}} = 150 \text{ mA}, T_{\text{J}} = -40 \text{ °C to } 150 \text{ °C}$		25		mV

Table 9. Current sense (8 V < V_{CC} < 36 V)

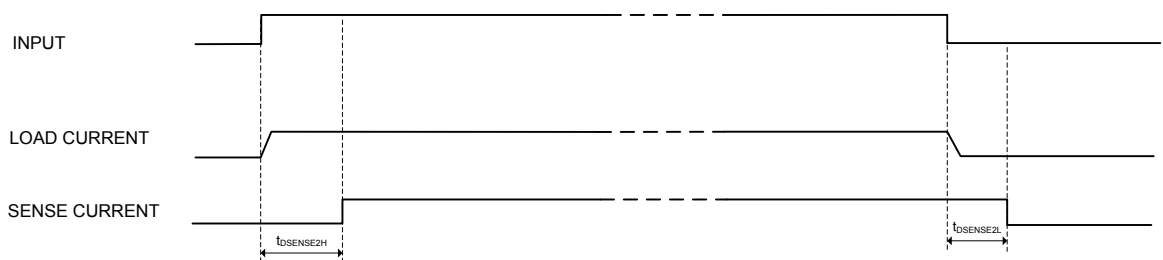
Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
K _{OL}	I _{OUT} /I _{SENSE}	I _{OUT} = 12 mA, V _{SENSE} = 0.5 V, T _J = -40 °C to 150 °C	820			
		I _{OUT} = 12 mA, V _{SENSE} = 0.5 V, T _J = 25 °C to 150 °C	1357			
K _{LED}	I _{OUT} /I _{SENSE}	I _{OUT} = 60 mA, V _{SENSE} = 0.5 V, T _J = -40 °C to 150 °C	1690	4160	7210	
		I _{OUT} = 60 mA, V _{SENSE} = 0.5 V, T _J = 25 °C to 150 °C	2040		6460	
dK _{LED} /K _{LEDTOT} ⁽¹⁾	Current sense ratio drift	I _{OUT} = 12 mA to 60 mA, I _{CAL} = 35 mA, V _{SENSE} = 0.5 V, T _J = -40 °C to 150 °C	-40		40	%
K ₀	I _{OUT} /I _{SENSE}	I _{OUT} = 100 mA, V _{SENSE} = 0.5 V, T _J = -40 °C to 150 °C	2215	4030	6305	
dK ₀ /K ₀ ⁽¹⁾	Current sense ratio drift	I _{OUT} = 100 mA, V _{SENSE} = 0.5 V, T _J = -40 °C to 150 °C	-20		20	%
K ₁	I _{OUT} /I _{SENSE}	I _{OUT} = 0.8 A, V _{SENSE} = 1 V, T _J = -40 °C to 150 °C	2210	3400	4590	
dK ₁ /K ₁ ⁽¹⁾	Current sense ratio drift	I _{OUT} = 0.8 A, V _{SENSE} = 1 V, T _J = -40 °C to 150 °C	-18		18	%
K ₂	I _{OUT} /I _{SENSE}	I _{OUT} = 1.6 A, V _{SENSE} = 2 V, T _J = -40 °C to 150 °C	2541	3385	4059	
dK ₂ /K ₂ ⁽¹⁾	Current sense ratio drift	I _{OUT} = 1.6 A, V _{SENSE} = 2 V, T _J = -40 °C to 150 °C	-14		14	%
K ₃	I _{OUT} /I _{SENSE}	I _{OUT} = 2.4 A, V _{SENSE} = 4 V, T _J = -40 °C to 150 °C	2706	3360	3894	
dK ₃ /K ₃ ⁽¹⁾	Current sense ratio drift	I _{OUT} = 2.4 A, V _{SENSE} = 4 V, T _J = -40 °C to 150 °C	-10		10	%
K ₄	I _{OUT} /I _{SENSE}	I _{OUT} = 12 A, V _{SENSE} = 4 V, T _J = -40 °C to 150 °C	3055	3313	3445	
dK ₄ /K ₄ ⁽¹⁾	Current sense ratio drift	I _{OUT} = 12 A, V _{SENSE} = 4 V, T _J = -40 °C to 150 °C	-4		4	%
dK _{BULB} /K _{BULBTOT} ⁽¹⁾	Current sense ratio drift	I _{OUT} = 1.3 A to 3.2 A, I _{OUTCAL} = 2.2 A, V _{SENSE} = 2 V, T _J = -40 °C to 150 °C	-18		18	%
I _{SENSE0}	Analog sense leakage current	I _{OUT} = 0 A, V _{SENSE} = 0 V, V _{IN} = 0 V, T _J = -40 °C to 150 °C	0		1	μA
		I _{OUT} = 0 A, V _{SENSE} = 0 V, V _{IN} = 5 V, T _J = -40 °C to 150 °C	0		2	
V _{SENSE}	Max analog sense output voltage	I _{OUT} = 12 A, R _{SENSE} = 3.9 kΩ	5			V
V _{SENSEH}	Analog sense output voltage in fault condition ⁽²⁾	V _{CC} = 24 V, R _{SENSE} = 3.9 kΩ	7.5	8.5	9.5	V
I _{SENSEH}	Analog sense output current in fault condition ⁽²⁾	V _{CC} = 24 V, V _{SENSE} = 5 V	4.9	9	12	mA
t _{DSSENSE2H}	Delay response time from rising edge of INPUT pins	V _{SENSE} < 4 V, 0.2 A < I _{OUT} < 12 A, I _{SENSE} = 90% of I _{SENSEMAX} , (see Figure 6)		200	400	μs

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$\Delta t_{\text{DSENSE2H}}$	Delay response time between rising edge of output current and rising edge of current sense	$V_{\text{SENSE}} < 4 \text{ V}$, $I_{\text{SENSE}} = 90 \% \text{ of } I_{\text{SENSEMAX}}$, $I_{\text{OUTMAX}} = 3 \text{ A}$ (see Figure 10)			250	μs
t_{DSENSE2L}	Delay response time from falling edge of INPUT pins	$V_{\text{SENSE}} < 4 \text{ V}$, $0.2 \text{ A} < I_{\text{OUT}} < 12 \text{ A}$, $I_{\text{SENSE}} = 10 \% \text{ of } I_{\text{SENSEMAX}}$, $I_{\text{OUT}} = 90 \% \text{ of } I_{\text{OUTMAX}}$ (see Figure 6)		5	20	μs

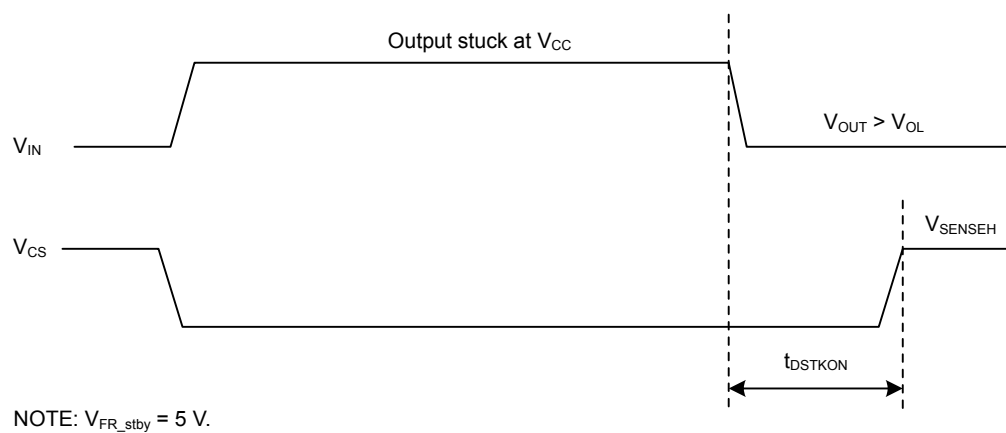
1. Specified by design, not tested in production.
2. Fault condition includes: power limitation, overtemperature and openload in off-state condition.

Table 10. Openload detection ($V_{\text{FR_Stby}} = 5 \text{ V}$)

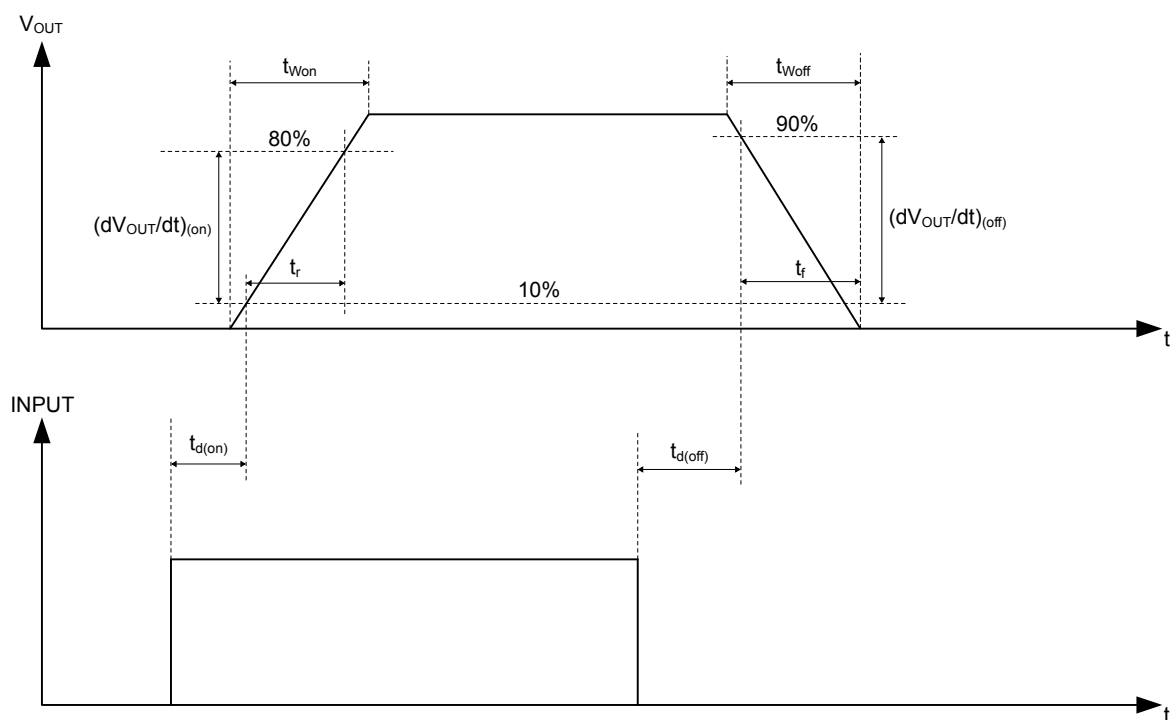
Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_{OL}	Openload off-state voltage detection threshold	$V_{\text{IN}} = 0 \text{ V}$, $8 \text{ V} < V_{\text{CC}} < 36 \text{ V}$	2	-	4	V
t_{DSTKON}	Output short circuit to VCC detection delay at turn off	See Figure 7	180	-	1800	μs
$I_{\text{L(off2)}}$	Off-state output current at $V_{\text{OUT}} = 4 \text{ V}$	$V_{\text{IN}} = 0 \text{ V}$, $V_{\text{SENSE}} = 0 \text{ V}$, V_{OUT} rising from 0 V to 4 V	-120	-	0	μA
$t_{\text{d_vol}}$	Delay response from output rising edge to V_{SENSE} rising edge in openload	$V_{\text{OUT}} = 4 \text{ V}$, $V_{\text{IN}} = 0 \text{ V}$, $V_{\text{SENSE}} = 90 \% \text{ of } V_{\text{SENSEH}}$, $R_{\text{SENSE}} = 3.9 \text{ k}\Omega$		-	20	μs
$t_{\text{DFRSTK_ON}}$	Output short circuit to V_{CC} detection delay at FR_Stby activation	Input1, 2 = low (see Figure 9)		-	50	μs

Figure 6. Current sense delay characteristics


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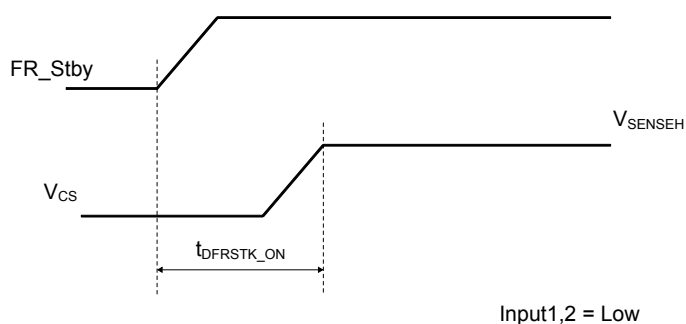
Figure 7. Openload off-state delay timing


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Figure 8. Switching characteristics


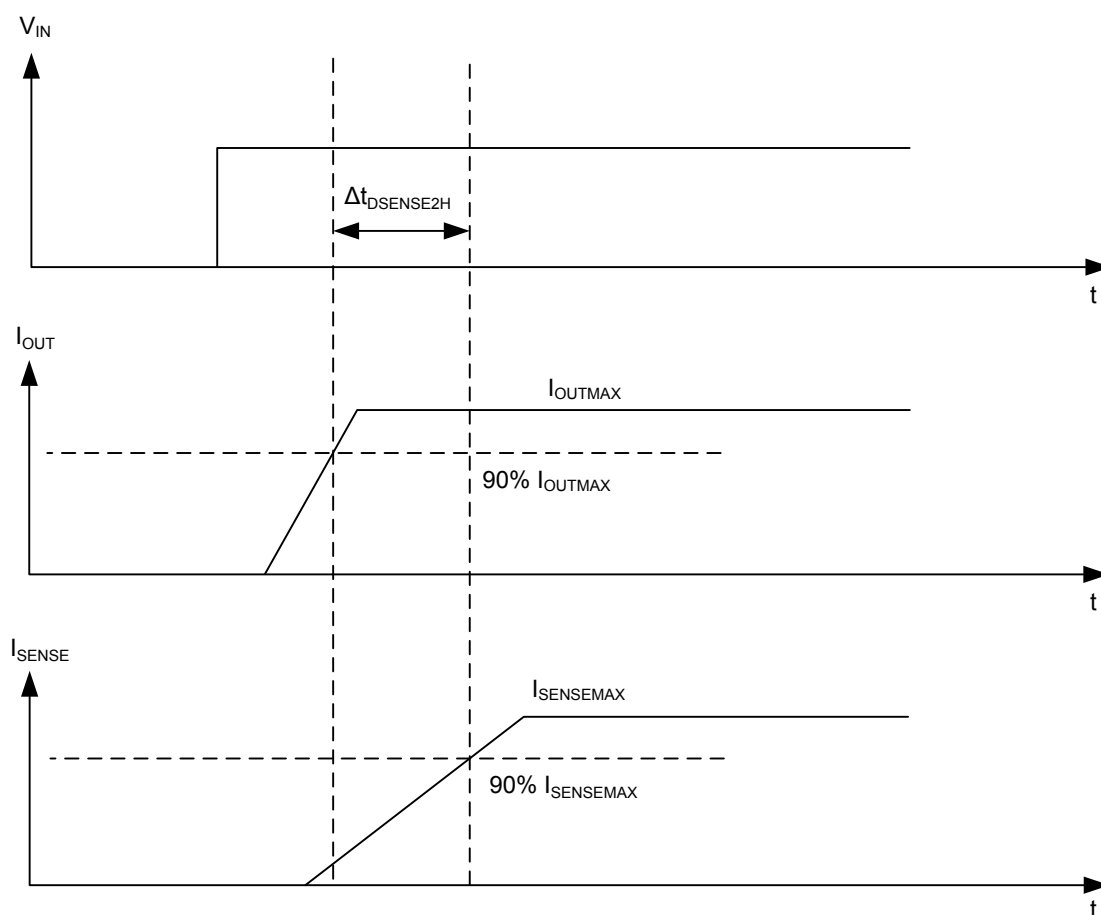
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Figure 9. Output stuck to V_{CC} detection delay time at FR_Stby activation

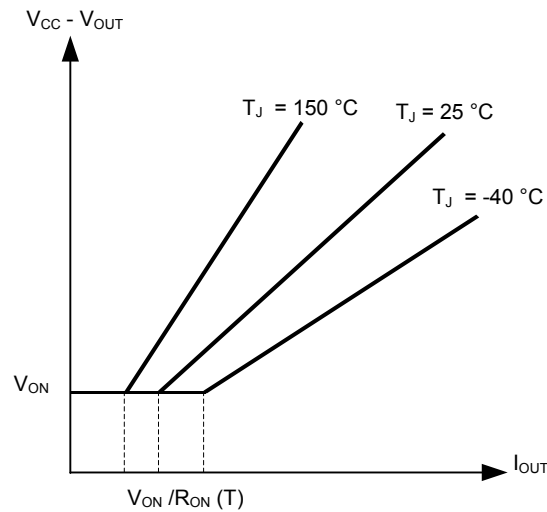


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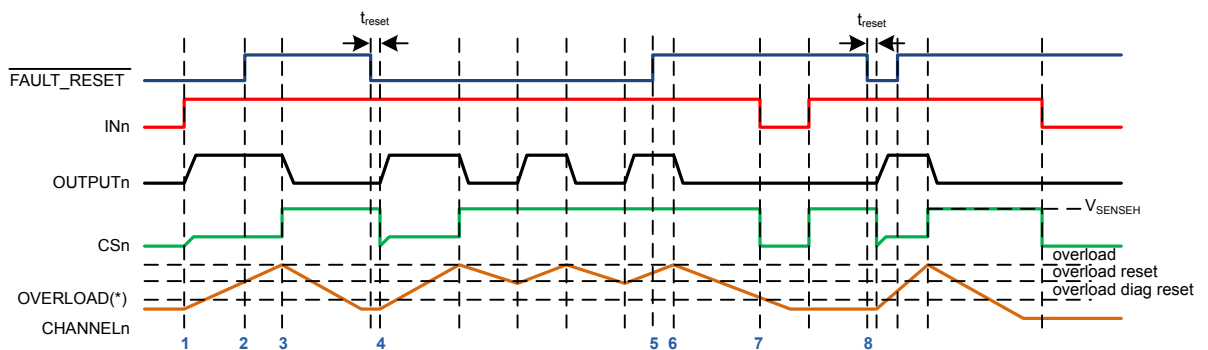
Figure 10. Delay response time between rising edge of output current and rising edge of current sense



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Figure 11. Output voltage drop limitation


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Figure 12. Device behavior in overload condition


- 1: OUTPUTn and CSn controlled by INn
- 2: FAULT_RESET from '0' to '1' → no action on CSn pin
- 3: overload latch-off. INn high → CSn high
- 4: FAULT_RESET low AND Temp channeln < overload_reset → overload latch reset after t_reset
- 4 to 5: FAULT_RESET low AND INn high → thermal cycling, CSn high
- 5: FAULT_RESET high → latch-off reset disabled
- 6 to 7: overload event and FAULT_RESET high → latch-off, no thermal cycling
- 7 to 8: overload diagnostic disabled/enabled by the input
- 8: overload latch-off reset by FAULT_RESET

(*) OVERLOAD = thermal shutdown OR power limitation

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Table 11. Truth table

Conditions	Fault reset standby	Input	Output	Sense
Standby	L	L	L	0
Normal operation	X	L	L	0
	X	H	H	Nominal
Overload	X	L	L	0
	X	H	H	> Nominal
Overtemperature/short to ground	X	L	L	0
	L	H	Cycling	V _{SENSEH}
	H	H	Latched	V _{SENSEH}
Undervoltage	X	X	L	0
Short to V _{BAT}	L	L	H	0
	H	L	H	V _{SENSEH}
	X	H	H	< Nominal
Openload off-state (with pull-up)	L	L	H	0
	H	L	H	V _{SENSEH}
	X	H	H	0
Negative output voltage clamp	X	L	Negative	0

Table 12. Electrical transient requirements (part 1)

ISO 7637-2: 2004 (E) Test pulse	Test levels ⁽¹⁾		Number of pulses or test times	Burst cycle/pulse repetition time		Delays and impedance
	III	IV				
1	-450 V	-600 V	5000 pulses	0.5 s	5 s	1 ms, 50 Ω
2a	37 V	50 V	5000 pulses	0.2 s	5 s	50 μs, 2 Ω
3a	- 150 V	- 200 V	1 h	90 ms	100 ms	0.1 μs, 50 Ω
3b	+ 150 V	+ 200 V	1 h	90 ms	100 ms	0.1 μs, 50 Ω
4	- 12 V	- 16 V	1 pulse			100 ms, 0.01 Ω
5b ⁽²⁾	+ 123 V	+ 174 V	1 pulse			350 ms, 1 Ω

1. The above test levels must be considered referred to V_{CC} = 24.5 V except for pulse 5b.
2. Valid in case of external load dump clamp: 58 V maximum referred to ground.

Table 13. Electrical transient requirements (part 2)

ISO 7637-2: 2004 (E) Test pulse	Test level results ⁽¹⁾	
	III	IV
1	C	C
2a	C	C
3a	C	C
3b ⁽²⁾	E	E
3b ⁽³⁾	C	C
4	C	C
5b ⁽⁴⁾	C	C

1. In order to guarantee the ISO transient classes a minimum 10 kΩ protection resistors are needed on logic pins.
2. Without capacitor between V_{CC} and GND.
3. With 10 nF between V_{CC} and GND.
4. External load dump clamp, 58 V maximum, referred to ground.

Table 14. Electrical transient requirements (part 3)

Class	Contents
C	All functions of the device are performed as designed after exposure to disturbance.
E	One or more functions of the device are not performed as designed after exposure to disturbance and cannot be returned to proper operation without replacing the device.

2.4 Electrical characteristics (curves)

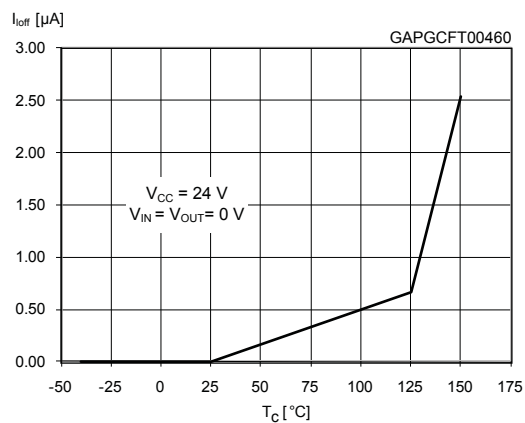
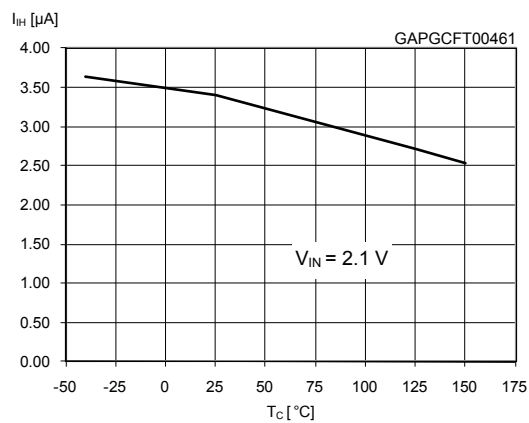
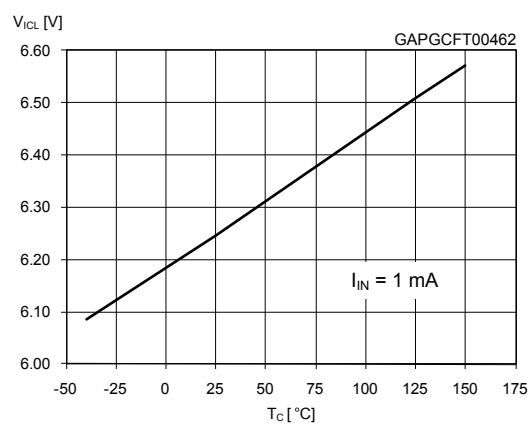
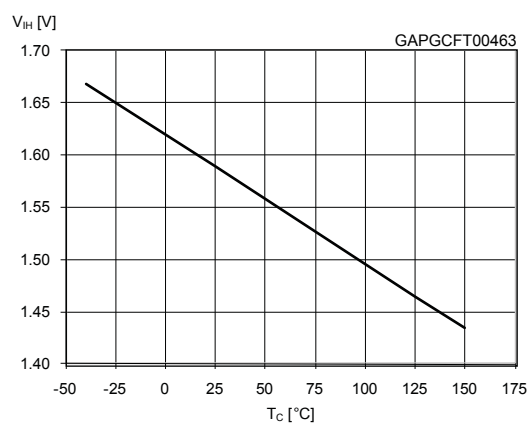
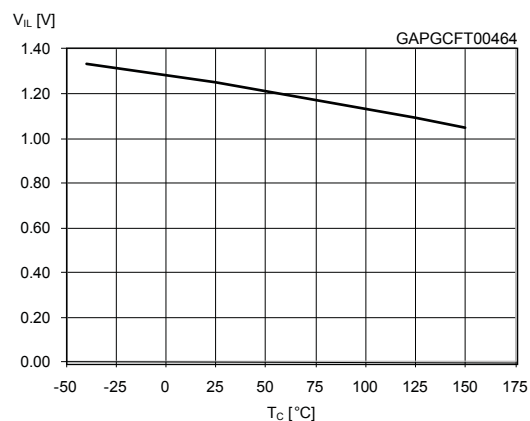
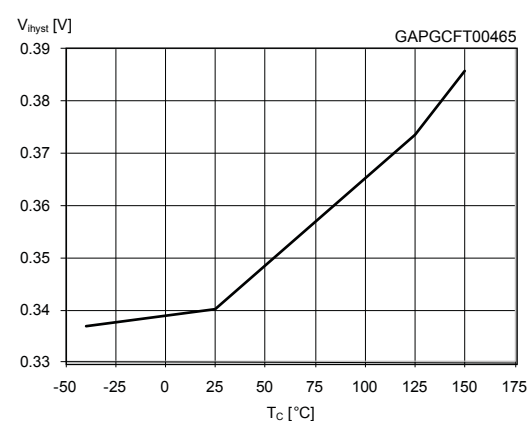
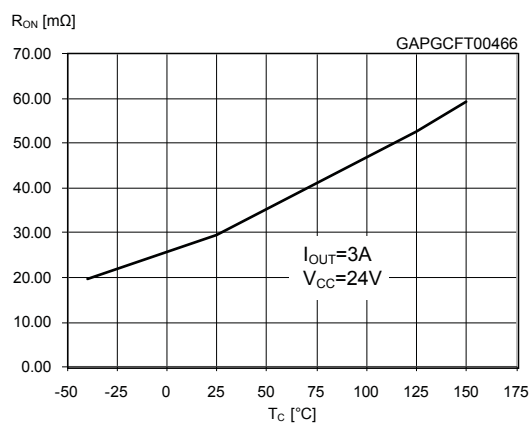
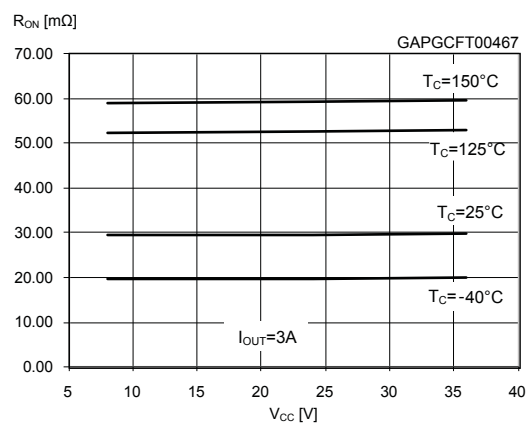
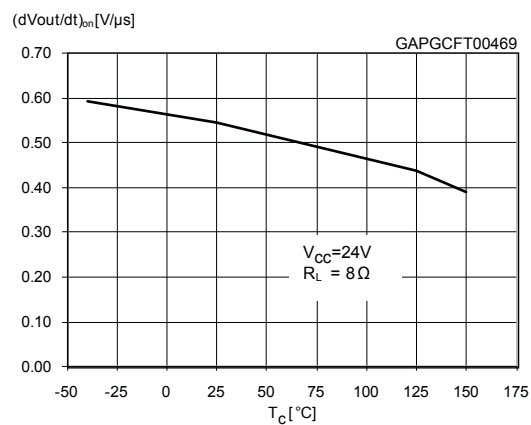
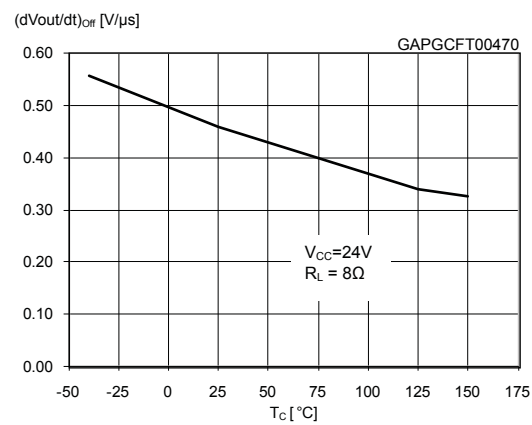
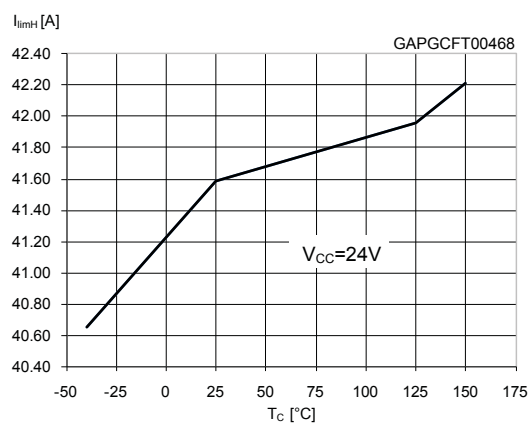
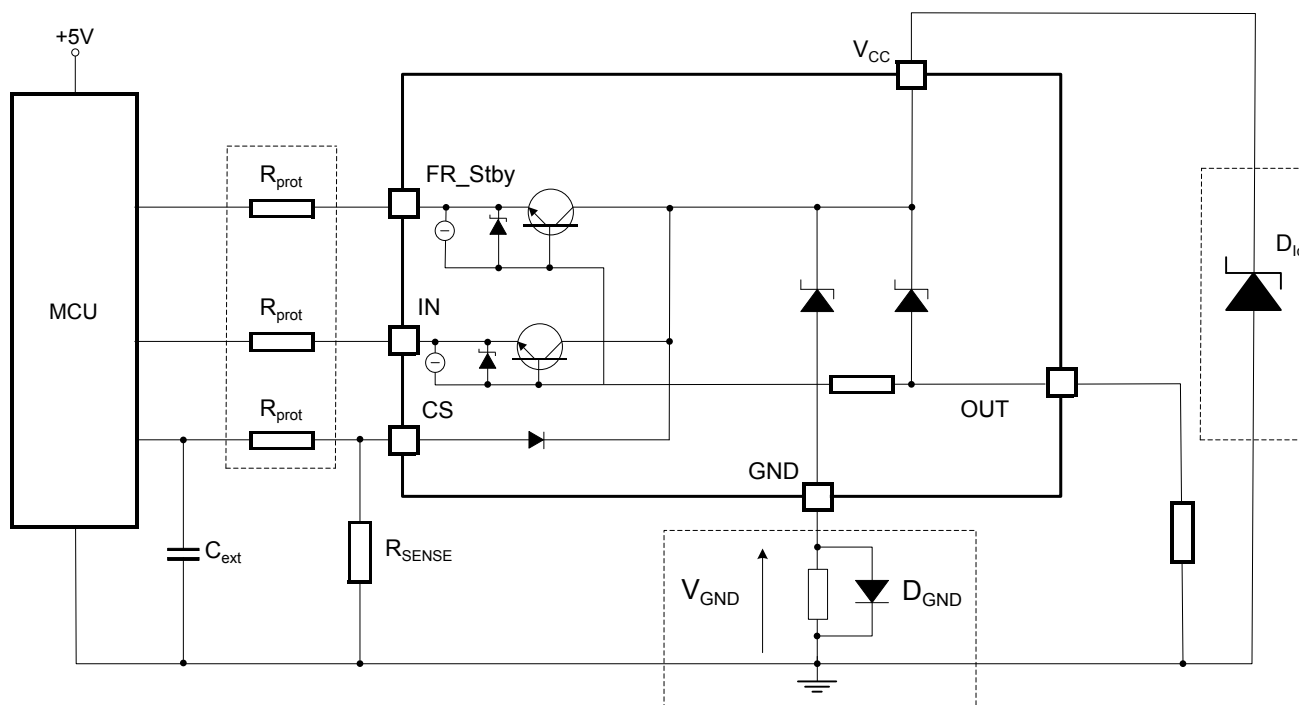
Figure 13. Off-state output current

Figure 14. High level input current

Figure 15. Input clamp voltage

Figure 16. High level input voltage

Figure 17. Low level input voltage

Figure 18. Input hysteresis voltage


Figure 19. On-state resistance vs T_c

Figure 20. On-state resistance vs V_{CC}

Figure 21. Turn-on voltage slope

Figure 22. Turn-off voltage slope

Figure 23. I_{LIMH} vs T_c


3 Application information

Figure 24. Application schematic



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3.1 GND protection network against reverse battery

3.1.1 Solution 1: resistor in the ground line (R_{GND} only)

This solution can be used with any load type.

The following is an indication on how to dimension the R_{GND} resistor.

1. $R_{GND} \leq 600 \text{ mV} / (I_{S(on)max.})$
2. $R_{GND} \geq (-V_{CC}) / (-I_{GND})$

Where $-I_{GND}$ is the DC reverse ground pin current and can be found in the absolute maximum rating section of the device datasheet.

Power dissipation in R_{GND} (when $V_{CC} < 0 \text{ V}$: during reverse battery situations) is:

$$P_D = (-V_{CC})^2 / R_{GND}$$

This resistor can be shared among several different HSDs. Please note that the value of this resistor should be calculated with formula (1) where $I_{S(on)max.}$ becomes the sum of the maximum on-state currents of the different devices.

Please note that if the microprocessor ground is not shared by the device ground then the R_{GND} produces a shift ($I_{S(on)max.} \cdot R_{GND}$) in the input thresholds and the status output values. This shift varies depending on how many devices are ON in the case of several high side drivers sharing the same R_{GND} .

If the calculated power dissipation leads to a large resistor or several devices have to share the same resistor, then ST suggests solution 2 is used (see below).

3.1.2 Solution 2: diode (D_{GND}) in the ground line

A resistor ($R_{GND} = 4.7 \text{ k}\Omega$) should be inserted in parallel to D_{GND} if the device drives an inductive load.

This small signal diode can be safely shared amongst several different HSDs. Also in this case, the presence of the ground network produces a shift ($\approx 600 \text{ mV}$) in the input threshold and in the status output values, if the microprocessor ground is not common to the device ground. This shift does not vary if more than one HSD shares the same diode/resistor network.

3.2 Load dump protection

D_{ld} is necessary (voltage transient suppressor) if the load dump peak voltage exceeds the V_{CC} maximum DC rating. The same applies if the device is subject to transients on the V_{CC} line that are greater than the ones shown in the ISO 7637-2 2004 (E) [Table 12](#), [Table 13](#) and [Table 14](#).

3.3 MCU I/Os protection

If a ground protection network is used and negative transient is present on the V_{CC} line, the control pins are pulled negative. ST suggests that a resistor (R_{prot}) has to be inserted in line to prevent the microcontroller I/Os pins from latching-up.

The value of these resistors is a compromise between the leakage current of the microcontroller and the current required by the HSD I/Os (Input levels compatibility) with the latch-up limit of microcontroller I/Os.

Equation: R_{prot} range calculation

$$-V_{CCpeak}/I_{latchup} \leq R_{prot} \leq (V_{OH\mu C} - V_{IH} - V_{GND}) / I_{IHmax}$$

Calculation example:

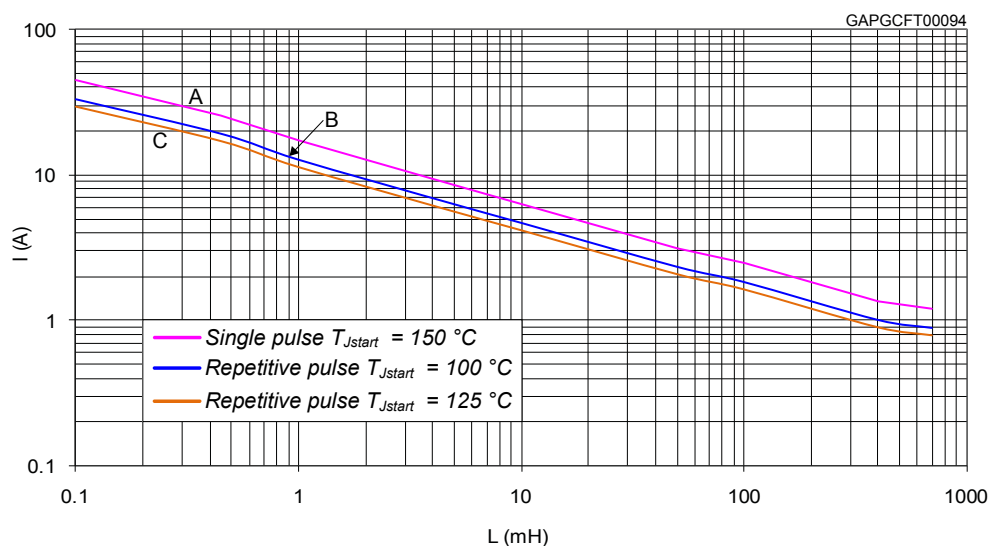
For $V_{CCpeak} = -600 \text{ V}$ and $I_{latchup} \geq 20 \text{ mA}$; $V_{OH\mu C} \geq 4.5 \text{ V}$

$$30 \text{ k}\Omega \leq R_{prot} \leq 180 \text{ k}\Omega.$$

Recommended value: $R_{prot} = 60 \text{ k}\Omega$.

4 Maximum demagnetization energy ($V_{CC} = 24\text{ V}$)

Figure 25. Maximum turn off current versus inductance

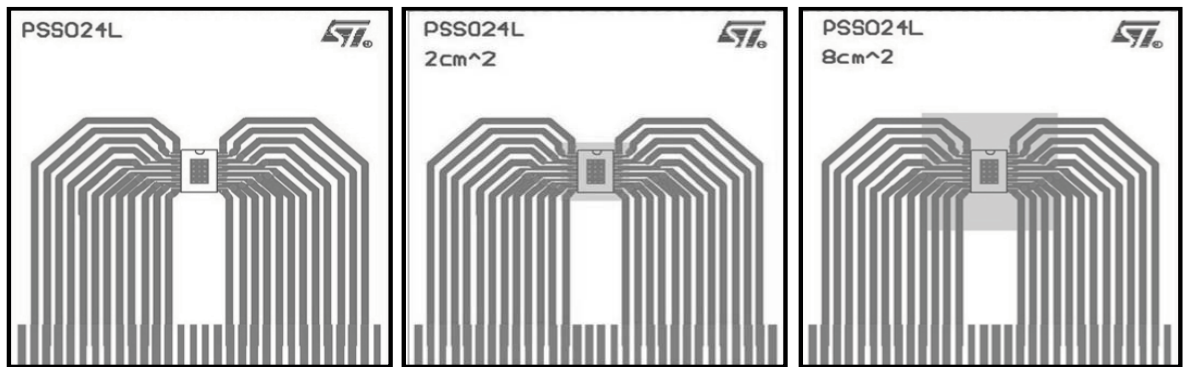


Note: Values are generated with $R_L = 0\text{ }\Omega$. In case of repetitive pulses, T_{Jstart} (at the beginning of each demagnetization) of every pulse must not exceed the temperature specified above for curves A and B.

5 Package and PCB thermal data

5.1 PowerSSO-24 thermal data

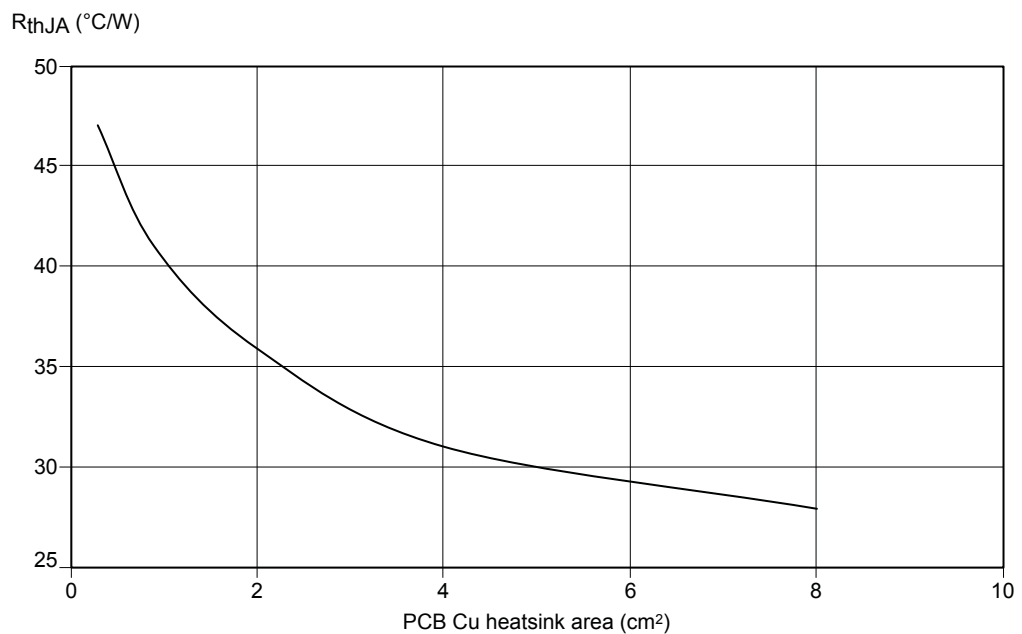
Figure 26. PowerSSO-24 PCB



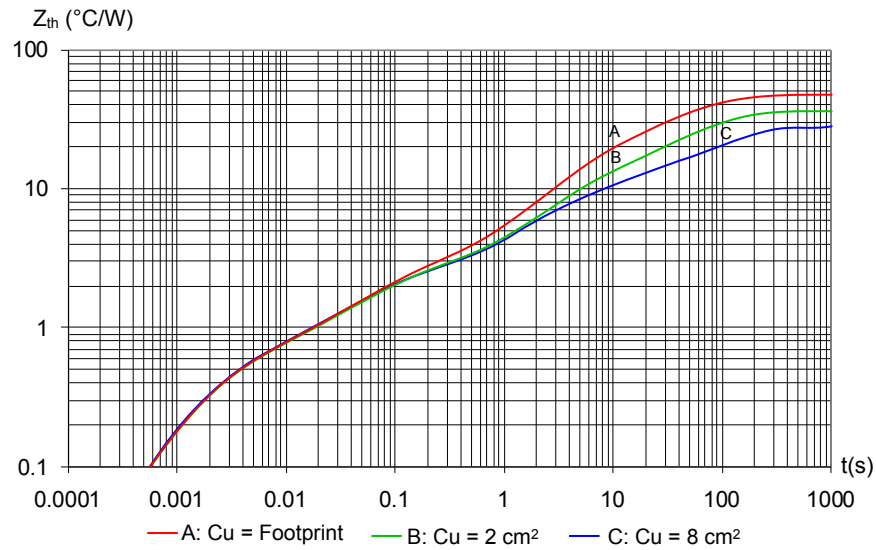
GAPGCFT00418

Layout condition of R_{th} and Z_{th} measurements (PCB: double layer, thermal vias, FR4 area = 77 mm x 86 mm, PCB thickness = 1.6 mm, Cu thickness = 70 μ m (front and back side), copper areas: from minimum pad lay-out to 8 cm²).

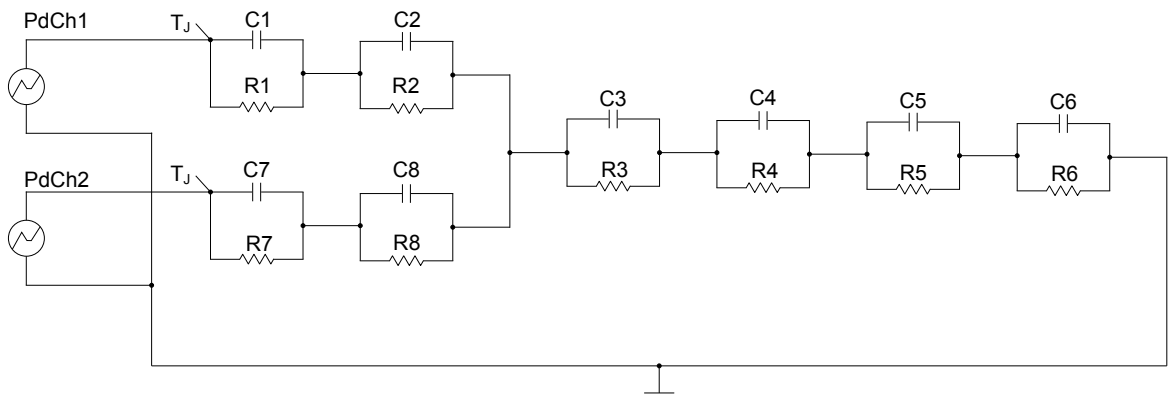
Figure 27. R_{thJA} vs PCB copper area in open box free air condition (one channel ON)



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Figure 28. PowerSSO-24 thermal impedance junction ambient single pulse (one channel ON)


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Figure 29. Thermal fitting model of a double channel HSD in PowerSSO-24


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The fitting model is a simplified thermal tool and is valid for transient evolutions where the embedded protections (power limitation or thermal cycling during thermal shutdown) are not triggered.

Equation: pulse calculation formula

$$Z_{th\delta} = R_{th} \cdot \delta + Z_{thp} (1 - \delta)$$

where $\delta = t_p/T$

Table 15. Thermal parameters

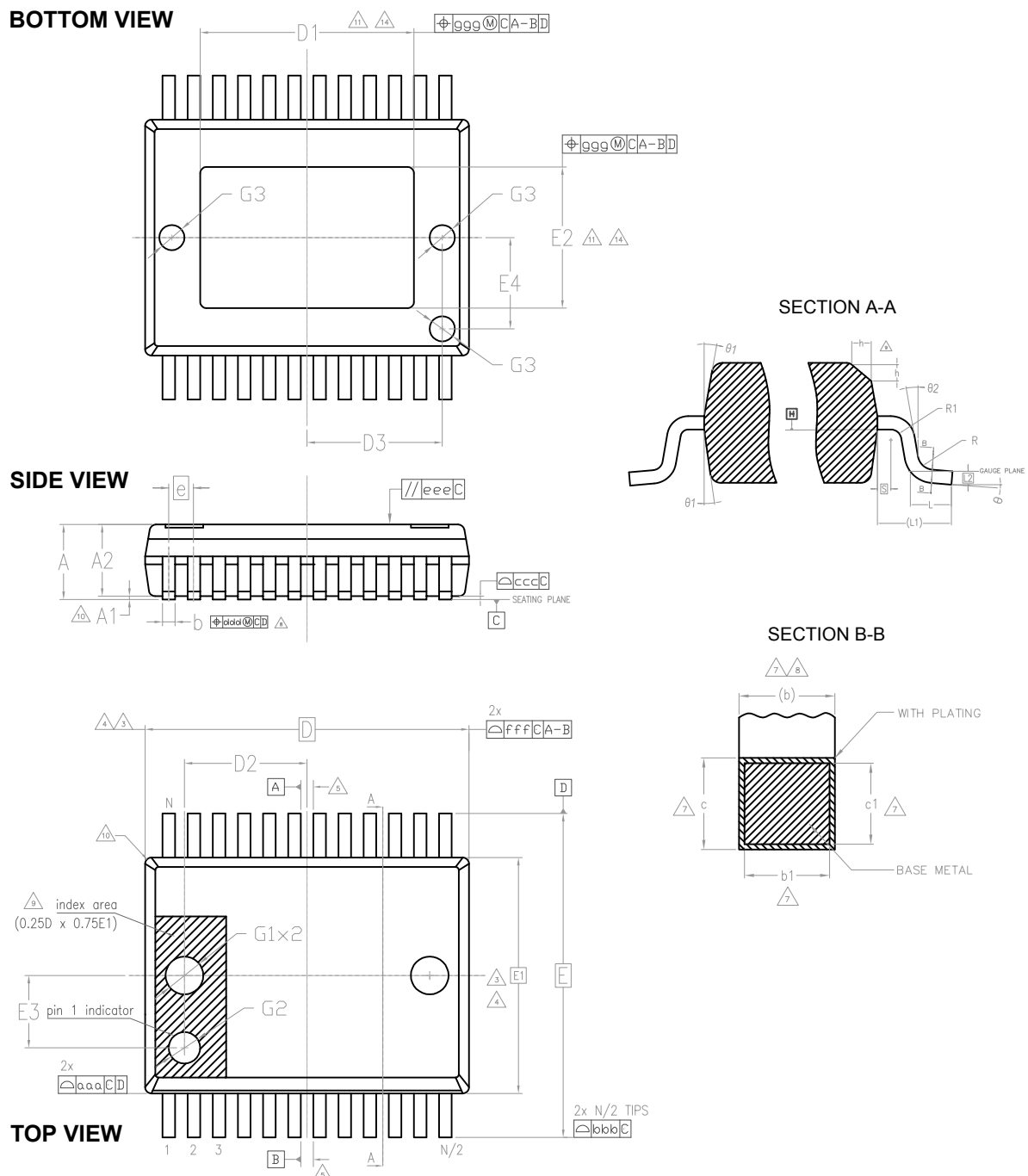
Area/island (cm ²)	Footprint	2	8
R1 (°C/W)	0.5		
R2 (°C/W)	0.75		
R3 (°C/W)	1		
R4 (°C/W)	7.7		
R5 (°C/W)	9	9	8
R6 (°C/W)	28	17	10
R7 (°C/W)	0.5		
R8 (°C/W)	0.75		
C1 (W.s/°C)	0.005		
C2 (W.s/°C)	0.05		
C3 (W.s/°C)	0.1		
C4 (W.s/°C)	0.5		
C5 (W.s/°C)	1	4	9
C6 (W.s/°C)	2.2	5	17
C7 (W.s/°C)	0.005		
C8 (W.s/°C)	0.05		

6 Package information

In order to meet environmental requirements, ST offers these devices in different grades of **ECOPACK** packages, depending on their level of environmental compliance. ECOPACK specifications, grade definitions and product status are available at: www.st.com. ECOPACK is an ST trademark.

6.1 PowerSSO-24 package information

Figure 30. PowerSSO-24 package dimensions



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Table 16. PowerSSO-24 mechanical data

Symbol	Millimeters		
	Min.	Typ.	Max.
θ	0°		8°
$\theta 1$	5°		10°
$\theta 2$	0°		
A			2.45
A1	0.00		0.10
A2	2.15		2.35
b	0.33		0.51
b1	0.28	0.40	0.48
c	0.23		0.32
c1	0.20	0.20	0.30
D	10.30 BSC		
D1	6.50		7.10
D2		3.65	
D3		4.30	
e	0.80 BSC		
E	10.30 BSC		
E1	7.50 BSC		
E2	4.10		4.70
E3		2.30	
E4		2.90	
G1		1.20	
G2		1.00	
G3		0.80	
h	0.30		0.40
L	0.55	0.70	0.85
L1	1.40 REF		
L2	0.25 BSC		
N	24		
R	0.30		
R1	0.20		
S	0.25		

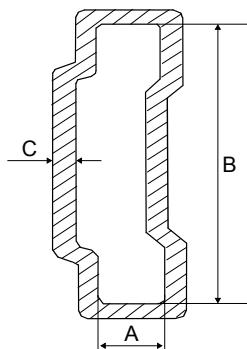
Table 17. PowerSSO-24 tolerance of form and position

Symbol	Millimeters
aaa	0.20
bbb	0.20
ccc	0.10
ddd	0.20

Symbol	Millimeters
eee	0.10
fff	0.20
ggg	0.15

6.2 PowerSSO-24 packing information

Figure 31. PowerSSO-24 tube shipment (no suffix)

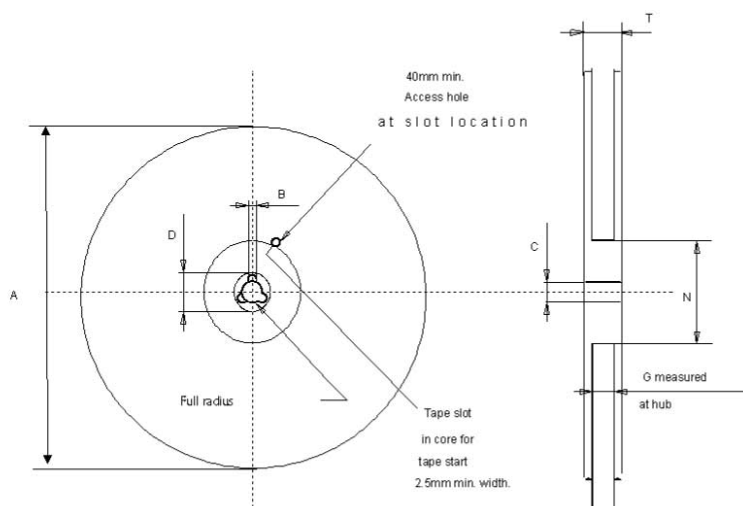


Base Q.ty	49
Bulk Q.ty	1225
Tube length (± 0.5)	532
A	3.5
B	13.8
C (± 0.1)	0.6

All dimensions are in mm.

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Figure 32. PowerSSO-24 tape and reel shipment (suffix "TR")



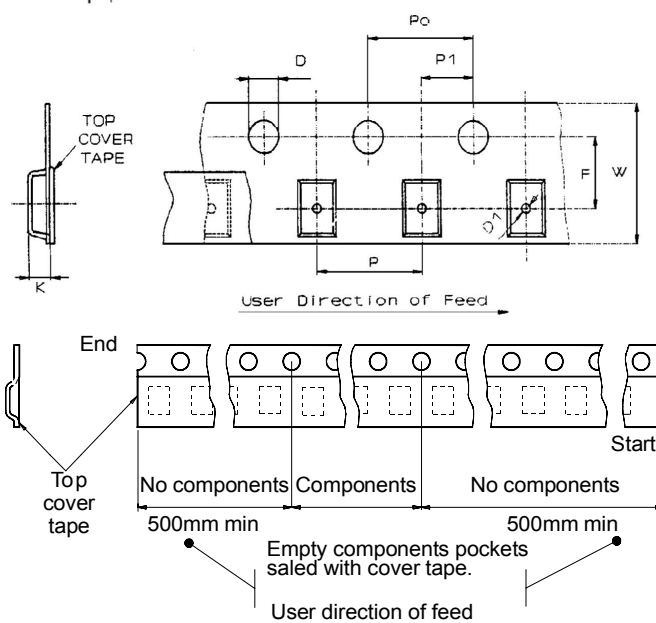
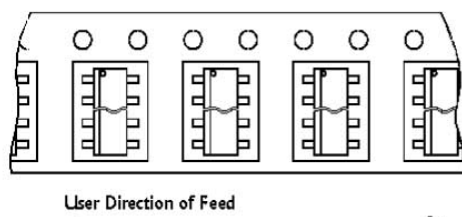
Reel dimensions

Base Q.ty	1000
Bulk Q.ty	1000
A (max)	330
B (min)	1.5
C (± 0.2)	13
F	20.2
G (+2/-0)	24.4
N (min)	100
T (max)	30.4

Tape dimension
 According to Electronic Industries Association (EIA) Standard 481 rev. A, Feb 1986

Tape width	W	24
Tape hole spacing	P0 (± 0.1)	4
Component spacing	P	12
Hole diameter	D (± 0.05)	1.55
Hole diameter	D1 (min)	1.5
Hole position	F (± 0.1)	11.5
Compartment depth	K(max)	2.85
Hole spacing	P1 (± 0.1)	2

All dimensions are in mm.



GAPGCF00421

Revision history

Table 18. Document revision history

Date	Revision	Changes
15-Jun-2012	1	Initial release.
19-Jun-2012	2	Changed document state form "Target specification" to "Production data".
18-Sep-2013	3	Updated Disclaimer.
17-May-2022	4	Updated PowerSSO-24 cover image. Updated Table 5. Power section. Updated notes in Table 12. Electrical transient requirements (part 1) Updated Figure 24. Application schematic. Moved Section 3.4: Maximum demagnetization energy ($V_{CC} = 24\text{ V}$) to Section 4 Maximum demagnetization energy ($V_{CC} = 24\text{ V}$) Updated Section 6.1 PowerSSO-24 package information. Minor text changes.

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